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CHAPTER V : TRANSIENT ELECTROMAGNETICS

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Japanese Contributions  
to IAEA INTOR Workshop, Phase Two A, Part 2  
Chapter V : Transient Electromagnetics

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This report corresponds to Chapter V of Japanese contribution report to IAEA INTOR Workshop, Phase Two A, Part 2. Simulation results are shown for feedback control of plasma position, electromagnetic forces at disruptions, penetration of electric and magnetic fields, and benchmark tests for transient electromagnetics. Design guide lines for feedback control system and database assessments are also reported.

Keywords: Design Guide Lines, Tokamak  
Plasma Position Control, Transient Electromagnetics  
INTOR, Feedback Control, Disruptions, Benchmark Test

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IAER INTOR ワークショップ，フェーズⅡA，パート2 報告書

第V章：電磁過渡現象解析

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## 1. Introduction

The vertically elongated plasmas are adopted in the design of the next generation machines such as Fusion Experimental Reactor at JAERI and INTOR, because of their expected high beta limit and economical use of the space inside the D-shaped toroidal field coils. Since the decay index  $n$ -value of the external equilibrium field is negative, the vertically elongated plasma is positionally unstable against the vertical movement. The growth rate of the vertical instability has to be suppressed with the conductive components surrounding the plasma so that the active feedback control with poloidal coils can stabilize the vertical instability. For this purpose, the highly conductive shells will be installed in the blanket modules in the next generation machines. However, the highly conductive shells are feared to induce too large electromagnetic forces at plasma disruption and to reduce the tritium breeding ratio in the blanket.

The location of active control coils will provide significant impacts on the design and assembling/disassembling of future tokamaks. The active coils inside radiation shield/coil vacuum chamber would make over-all tokamak machine and its maintenance procedure unreliable, though they mitigate the requirement for conductive shell design. On the other hand, when the active coils are located outside the shield/coil chamber with low toroidal one turn resistance, the shielding effect of passive elements against the control magnetic field would be a problem from the view point of feedback control property. Toroidal one turn resistance of radiation shield/coil vacuum chamber affects this shielding effect, since it surrounds plasma perfectly.

The discussions on these problems are described in the following sections.

## 2. Plasma Stabilization

### 2.1 Summary of Plasma Stabilization Experimental Experience

A dee-shaped plasma may have attractive properties for future tokamak reactors. Control of the plasma shape and position is one of most important techniques in these devices, because vertical elongation make the plasma subject to axisymmetric positional instability. The positional instability of elongated plasmas has been studied by many authors and has been observed in several laboratories.

This section reviews the recent progresses of the vertical stability experiment on Doublet III. Plasmas with surface elongations of up to 1.8 have been stably produced in the upper lobe of Doublet III tokamak with the use of both passive and active controls. The growth rates of vertical instability have been measured for plasmas with various elongation, triangularity, other plasma parameters. The measured growth rates are compared qualitatively with the predictions of a simplified model which is based on the relationship between the decay index of the externally applied field and the passive stabilization effect due to eddy currents, and also compared quantitatively with those obtained from numerical calculations based on a more realistic model which utilizes a linear perturbation treatment of the plasma equilibrium.

The following conclusions are obtained (see Reference [1]).

- a) Stable plasmas with elongations over the entire range from 1.0 to 1.8 (aspect ratio  $\approx 3.4$ ) are produced in the upper half of Doublet III with passive and active control. The maximum value 1.8 lies well above elongations obtained in other experiments. This is due to the fact that the field-shaping coils, which are connected in parallel, are located close to the plasma surface. They play a large part in stabilizing positional instability.
- b) Active feedback control fails to stabilize the positional instability of plasmas with elongations of more than 1.8. The upper limit of the elongation is determined by the strength of the passive stabilization effect ( $n_s$ ) even with active feedback control. The unstable plasma oscillates in the vertical direction, and simultaneous changes occur in the plasma shape and elongation.
- c) A dee-shaped plasma with a larger triangular deformation is more positionally stable than a plasma with a smaller triangular deformation because the triangular deformation of the cross-section tends to reduce the value of the decay index relative to that required for an ellipse of the same elongation.
- d) The broad profile of the plasma current is beneficial in reducing the vertical instability of an elongated plasma.



- e) The maximum elongation is decreased from  $\sim 1.8$  to  $\sim 1.25$  as the plasma minor radius is reduced from  $\sim 41$  cm to  $\sim 33$  cm. A low aspect ratio is beneficial for increasing the elongation and reducing vertical instability.

## 2.2 Vertical Position Control

### 2.2.1 Plasma model

The analyze transient electromagnetic effects, many different models are being used, including various models for the plasma, models for passive elements, and models for active components. This section provides a brief description of the various plasma models.

The following plasma models are being considered.

1. Rigid circular filament of constant radius and current, with zero mass and only vertical displacements allowed.
2. Same plasma model as 1, but for evaluating the mutual inductance between the plasma and other conductive circuits during plasma movement, the plasma is represented by a pair of filaments. This is called a dipole mode.
3. Plasma column with rigid vertical displacement either with or without skin currents in accordance with ideal MHD theory.
4. Plasma column with non-rigid vertical displacement either with or without current redistribution in accordance with ideal MHD theory.
5. Two-dimensional MHD equilibrium plasma flux diffusion and electromagnetics model.

The difference of analysis results on plasma stabilization effect is not large between model 1 and 2, as shown in Fig. 2.2.1. However, model 3 provides higher stabilizing effect comparing with model 2, since plasma magnetic surfaces shift outward direction and plasma current effectively approaches to outboard shell structures. So this improvement depends on shell configuration and plasma  $\beta_p$  etc. In the case of '83 JAERI FER, model 3 provides approximately 50% lower growth rate of vertical position instability as shown in Fig. 2.2.2.

In order to examine the difference between third and fourth models, the stability of the axisymmetric modes of a tokamak plasma is analyzed for the Solov'ev equilibrium [2] by using the linear ideal MHD code ERATO-J [3,4]. The comparison with the results by using the rigid model ( $n$ -index theory) [5] and the rigid displacement model [6] is made. In the rigid displacement model the constant vertical or horizontal displacement is assumed in the equation of motion. Figures 2.2.3 shows the stability boundary in the  $e$ - $\delta$  plane for different values of the aspect ratio,  $A$ , where  $e$  and  $\delta$  denote the ellipticity and triangularity of the plasma cross-section, respectively. The position of the conducting shell

is infinitely far from the plasma surface. The inner region of the solid lines are the stable region obtained by using the full MHD equation. The dashed line and the dashed solid line correspond to the lines for  $n_i=0$  and  $n_i=1.5$ , respectively. The  $n$ -index,  $n_i$ , is obtained at the magnetic axis of the Solov'ev equilibrium. The circle denotes the stability boundary against the vertical shift of the plasma ( $n_i=0$ ) by using the formula [5].

$$e < e_c = 1 + \frac{1}{A^2} \cdot \left( \frac{3}{4} \cdot \ln 8A - \frac{17}{16} \right) \quad (2.2.1)$$

The results by using the rigid model agree with the ones by using full MHD equation for  $\delta \sim 0$ . The stability boundaries by using the rigid displacement model are shown by the dashed lines in Fig. 2.2.4. The rigid displacement model gives the opposite dependency of the triangularity on the stability boundary. The comparison of three models shows that the analysis of the axisymmetric instability requires the full MHD calculations for large  $\delta$ .

The analysis by using the full MHD equation shows that the maximum value of the ellipticity,  $e_c$ , is about 1.2 when the conducting shell is placed at infinity. The stabilizing effect of the conducting shell increases  $e_c$  up to  $e_c=1.8$  for  $a_w/a_p \sim 2.0$  [4], where  $a_w$  and  $a_p$  denote the minor radius of the conducting shell and the plasma.

## 2.2.2 Parametric studies on vertical position control

### Fundamental conditions

To denote the positions and the directions, we use the cylindrical coordinate ( $R, \phi, Z$ ). Configurations of the plasma, the control coils and the conductive components are assumed to be symmetric to the  $Z=0$  plane. In what follows, we shall study under the following conditions.

(a) In the equilibrium state, the axisymmetric plasma is maintained by the external equilibrium field which is mainly characterized by the vertical field at the plasma center,  $B_{z0}$  and the decay index  $n$ -value. The  $n$ -value is defined by [7]

$$n = - \frac{1}{I_p B_{z0}} \int_p \frac{\partial B_R^e}{\partial Z} j_p R ds, \quad (2.2.2)$$

where,  $I_p$  is the toroidal plasma current,  $B_R^e$  is the equilibrium radial field and  $j_p$  is the distributed toroidal plasma current density. The integral is taken over the plasma cross section. It is assumed that the plasma moves rigidly in the vertical direction without the plasma current redistribution and that the plasma current and the external equilibrium field do not change during the plasma motion. In these assumptions, the  $Z$ -directional force exerted on the plasma column by the external equilibrium field is given by  $2\pi n I_p B_{z0} Z_p$  where  $Z_p$  is the plasma vertical position from the equilibrium position.

(b) The active control radial field is generated by one pair of the poloidal coils to simplify the analysis. The top and bottom coils are connected in a series and current directions are inversed to each other.

(c) The conductive structures surrounding the plasma are not axisymmetric in general but are located cyclically in the toroidal direction. These are approximated by the assemblies of thin conductors and the eddy currents in the conductors are expanded to the eddy current eigenmodes which are mutually decoupled and have the time constants,  $\tau_i$ 's as shown in Ref. [8,9]. Each eddy current mode is normalized to be a circuit with a self-inductance  $\tau_i$  and a resistance of one unit.

(d) In the equilibrium state, the mutual inductances between the plasma and the control coils and between the plasma and the eddy current modes are zero because of the symmetry for the  $Z=0$  plane. The  $Z$ -derivatives of the mutual inductances between the plasma and the others are calculated by the dipole plasma current approximation given in Ref. [10]. The dipole currents are located at  $R=R_p$  and  $Z=\pm\pi a_p/4$  and the  $Z$ -derivatives of the mutual inductances are given by dividing the mutual inductances between the dipole currents and others by  $\pi a_p/2$ , here  $a_p$  is the minor radius of the plasma.

(e) The plasma displacement from the equilibrium state is small and the coupling between the plasma motions in the radial direction and in the vertical direction is neglected and the equations of the plasma motion in the vertical direction is linearized around the equilibrium state.

### Basic equations

The interactions among the plasma vertical motion, the control coil and the eddy current modes in the conductive components are shown by the following evolution equations. The first one is the equation of the plasma motion given by,

$$M_p \ddot{Z}_p = -2\pi R_p I_p \left( -\frac{nB_{z0}}{R_p} Z_p + v_c I_c + \sum v_i I_i + B_d \right). \quad (2.2.3)$$

Here,  $M_p$  and  $R_p$  are the plasma mass and the major radius of the plasma column, respectively.  $I_c$ ,  $I_i$  and  $B_d$  are the control coil current, the current of the  $i$ -th eddy current mode and the disturbance radial field, respectively.  $v_c$  and  $v_i$  are the radial field of the unit current of the control coil and the average radial field of the unit current of the  $i$ -th eddy current mode, respectively. These are given by,

$$v_c = -\frac{M'_p}{2\pi R_p} \text{ and } v_i = -\frac{M'_p}{2\pi R_p} \quad (2.2.4)$$

where  $M'_{pc}$  and  $M'_{pi}$  are the Z-derivatives of mutual inductances between the plasma and the control coil and between the plasma and the i-th eddy current mode, respectively. A dot (·) indicates differentiation with respect to time. The summation is taken over the eddy current modes taken into consideration. The inertia term of Eq. (2.2.3) is neglected in what follows under the condition that the poloidal Alfvén time scale motion is stabilized by the stabilizing effect of the conductive components.

The second equation is the circuit equation of the control coil.

$$L_c \dot{I}_c + \sum_i M_{ci} \dot{I}_i + I_p M'_{pc} \dot{Z}_p + R_c I_c = V_c, \quad (2.2.5)$$

where  $L_c$  and  $R_c$  are the self-inductance and resistance of the control coil, respectively and  $V_c$  is the voltage applied to the control coil by the power supply.  $M_{ci}$  is the mutual inductance between the control coil and the i-th eddy current mode.

The others are the circuit equations of the eddy current modes.

$$M_{ci} \dot{I}_c + \tau_i \dot{I}_i + I_p M'_{pi} \dot{Z}_p + I_i = 0, \quad (i=1, \dots, N_{\text{mode}}). \quad (2.2.6)$$

Here,  $N_{\text{mode}}$  is the number of the eddy current modes taken into consideration.

By Laplace transformations of Eqs. (2.2.3), (2.2.5) and (2.2.6), the block diagram of the control object composed of the plasma, the control coil and the eddy current modes is obtained as shown in Fig. 2.2.5. Here,

$$N(s) = \sum_i \frac{n_i s \tau_i}{1 + s \tau_i}, \quad \text{where } n_i = - \frac{M'_{pi} I_p}{2\pi B_{z0} \tau_i}, \quad (2.2.7)$$

$$M(s) = \sum_i \frac{\mu_i s \tau_i}{1 + s \tau_i}, \quad \text{where } \mu_i = \frac{M_{ci} M'_{pi}}{M'_{pc} \tau_i}, \quad (2.2.8)$$

and

$$K(s) = \sum_i \frac{\kappa_i s \tau_i}{1 + s \tau_i}, \quad \text{where } \kappa_i = \frac{M_{ci}^2}{L_c \tau_i}, \quad (2.2.9)$$

These transfer functions represent the properties of the eddy current in the conductive components. The N-function,  $N(s)$ , represents the stabilizing effect of the plasma vertical motion by the eddy current. The

shielding function,  $M(s)$ , represents the shielding effect to the magnetic field of the control coil. When the plasma is perfectly surrounded by the conductive components and the control coil is located outside the components,  $M(\infty)=1$ . The coupling function,  $K(s)$ , represents the magnetic coupling effect between the control coil and the eddy current.

The overall transfer function of the control object from the coil voltage  $V_c$  to the plasma displacement  $Z_p$  is given by,

$$F(s) = \frac{\frac{R_p \tau_c V_c}{B_{zo} L_c} \cdot \frac{1-M(s)}{1+s\tau_c \{1-K(s)\}}}{n+N(s)+N_c(s)}, \quad (2.2.10)$$

where

$$N_c(s) = \frac{n_c s \tau_c \{1-M(s)\}^2}{1+s\tau_c \{1-K(s)\}}, \quad (2.2.11)$$

$$n_c = -\frac{M'_{pc} \frac{2I_p}{2\pi B_{zo} L_c}}{\tau_c} \quad \text{and} \quad \tau_c = \frac{L_c}{R_c}. \quad (2.2.12)$$

The growth time  $\gamma_g$  of the plasma vertical motion without the feedback control is the root of the equation given by,

$$n + N(\gamma_g) + N_c(\gamma_g) = 0. \quad (2.2.13)$$

When the decay index  $n$ -value is negative, Eq. (2.2.13) has a real positive root  $\gamma_g$  and the plasma moves vertically in the time scale of  $1/\gamma_g$ .

We consider a feedback control system as shown in Fig. 2.2.5. The position detector detects the plasma vertical position with the first order time delay,  $T_{\text{delay}}$ . The PID controller generates the proportional, integral and derivative signal of the difference between the reference plasma position,  $Z_{\text{ref}}$  and the detected plasma position,  $Z_{\text{det}}$ . Here,  $G$ ,  $T_I$ ,  $T_D$  and  $T_A$  is the gain, the integral time, the derivative time and the first order time delay of the derivative, respectively. The voltage limiter limits the maximum and minimum voltages of the power supply to  $\pm V_L$ . The power supply is approximated by the thyristor with the dead time,  $T_{\text{dead}}$ .

The transfer function of the open-loop of this feedback control system is given by,

$$A(s) = G \left( \frac{1}{sT_I} + 1 \frac{sT_D}{1+sT_A} \right) \cdot \frac{1}{1+sT_{\text{delay}}} \cdot e^{-sT_{\text{dead}}} F(s). \quad (2.2.14)$$

Since the transfer function  $F(s)$  has one real positive pole at  $s=\gamma$  when the  $n$ -value is negative, the stability criterion of the feedback loop is represented by Nyquist criterion that the plot of the transfer function of the open-loop,  $A(s)$  in the complex plane encircles counter-clockwise the point  $(-1, j0)$  when the  $s$  vaies from  $-j\infty$  to  $+j\infty$ , here  $j$  is an imaginery unit. Equivalently, the stability criterion is represented that the phase is greater than  $-180$  degrees when the gain is equal to 1 in the gain-phase diagram of  $A(s)$ .

Based on these formulations, the stabilizing properties of passive elements are studied for various types of shell models.

In order to see the machine dependence of control properties (such as difference of the decay index  $n$  value, shell design, etc.), we will show the parametric studies on two types of tokamak reactors, i.e. JAERI FER and INTOR.

#### 1. Design studies on passive shells of JAERI FER '84

Main plasma parameters of the JAERI FER '84 are summerized in Table 2.2.1.

The following items should be taken into account in designing shell structures for tokamak fusion reactors.

- 1) The conductive shells have high enough stabilizing effect on plasma vertical position instability without making active coil design and its installation difficult.
- 2) They should not induce excessive electro-magnetic forces in the blanket structures at plasma disruption.
- 3) Their structures should be consistent with remote maintenance procedures.
- 4) They should not excessively reduce the tritium breeding ratio.

We first studied whether the usual reactor component, such as first wall, blanket, shield and coil vacuum chamber, can stabilize the vertical position instability of the JAERI FER plasma without installing special conductive shell structures around the plasma, since this system should be the most preferable based on the above mentioned items except the item on active coil design and installation.

Figure 2.2.6 shows  $N(S)$ -function (defined by eq. (2.2.7)) of this passive element system. Since the  $n$ -index of the FER plasma is  $-1.62$ , the growth time of plasma vertical movement is approximately 4 msec. Figure 2.2.7 shows the gain-phase diagrams of this feedback control system as a parameter of active coil location. The location (a) is inside the shield and coil vacuum chamber, the location (b) is between the coil vacuum chamber and the toroidal field coils. Only the system with active coils at location (a) can stabilize the plasma vertical position instability. The design and installation of active coils used at location (a) (inside the shield/coil vacuum chamber) should not be practical considering associated engineering problems such as remote handling, insulation,

support of feeder lines, space arrangement, etc. So we gave up this kind of system and decided to install high conductive shells for stabilization of plasma positional instability.

It is well known that the toroidal one-turn conductive ring inside the blanket is the most preferable for a shell effect alone. However, this kind of structure will uselessly complicate the remote handling process and make it unreliable, since the segments of conductive ring must be connected and/or disconnected with each other by remote handling machine in the torus chamber. And the design of the remote machine itself would be difficult, because the blanket structure would not be so rigid to support the machine, and the port areas available for machine access are considered to be limited by the design requirements such as neutron/ $\gamma$ -ray shield around the ducts and the space for drawing coolant pipes and supporting the units. Therefore, we installed the conductive shell into each blanket module.

In this type of structure, large electro-magnetic stress and deformation may be induced at plasma disruptions especially in the inboard structure containing the conductive shell, since their support could not be sufficient because of restricted accessibility behind the inboard blanket modules. Further more, shell structures result in the reduction of tritium breeding ratio due to the increase in neutron absorption and the decrease in volume fraction of breeding material.

Considering these problems, we designed various types of the conductive shells, and performed parametric studies on their stabilizing characteristics and on the feedback control of plasma vertical position.

#### Parametric study on shell effects

Figures 2.2.8 (a) and (b) show our model of the conductive shell installed in the blanket module for this parametric study. In designing these configurations, the followings are considered.

- (a) Each conductive shell should be separately installed in each blanket module, since it is difficult to connect and/or disconnect the conductive shells with each other by remote handling within the limited space and accessibility.
- (b) The material of the conductive shell behind the first wall (front shell) serves also as neutron multiplier, since this region is the most effective zone for the tritium breeding.
- (c) The copper is used as the shell material at the side wall of the blanket module, since the conductivity of this part affects the overall shell effects significantly.

Five design choices are selected for the parametric study:

- (1) number of conductive shell segments along the torus, (2) with or without conductive shell in the inboard blanket (inboard shell), (3) thickness of outboard front shell, (4) thickness of outboard side shell, and (5) with or without conductive shell at the rear of tritium breeding

zone (end shell). The stabilizing effect of conductive shells represented by N-function is numerically calculated for ten shell models shown in Table 2.2.2. The parameters used to compare the stabilizing effect are summarized in Table 2.2.3.

#### (1) Number of conductive shell segments along the torus

The number of blanket modules is one of the most important design parameters with respect to assembling and disassembling the torus structure. Here, two series of survey are carried out: (a) with only outboard shell (model 10-30-80) and (b) with both inboard and outboard shell (model 20-40-90). Figures 2.2.9 and 2.2.10 show the N-functions of these two series with parameter of the number of the shell segments. As shown in these figures, the curves of N-function increase with the number of shell segments decreasing for both cases. In Figure 2.2.11, plasma instability growth times are plotted for the number of shell segments. The growth time is approximately in proportion to the reverse of the number of segments.

#### (2) With or without inboard conductive shell

The effect of inboard conductive shell can be understood by comparing two series of survey described above. Figure 2.2.11 shows that the inboard shell increases the growth time of positional instability by approximately 10 msec for every number of torus segmentation.

#### (3) Thickness of outboard front shell

Neutronics indicates that the tritium breeding ratio is sensitive to material and thickness of the outboard front shell. Here, we surveyed the effect of Pb shell thickness for the system of 28 shell segments. Two series are studied: model 30 and 31 for outboard side shell of 15 mm in thickness, and model 32 and 33 for that of 22.5 mm. The N-functions of these models are shown in Fig. 2.2.12. The sensitivities of the front shell thickness,  $t_{fs}$ , on the growth time,  $\tau_g$ , are summarized as

$$\frac{\partial \tau_g}{\partial t_{fs}} = \begin{cases} 0.07 \text{ msec/mm } P_b & \text{for 15 mm thick Cu side shell} \\ 0.11 \text{ msec/mm } P_b & \text{for 22.5 mm thick Cu side shell} \end{cases}$$

#### (4) Thickness of outboard side shell

Copper is used as a material of side shell, because of its high electrical conductivity. The tritium breeding ratio is also sensitive to the side shell thickness, since copper absorbs neutrons and reduce the area of breeding zone. Summarizing the results in Fig. 2.2.12, the results are represented as

$$\frac{\partial \tau_g}{\partial t_{ss}} = \begin{cases} 0.55 \text{ msec/mm} & \text{for 60 mm thick } P_b \text{ front shell} \\ 0.76 \text{ msec/mm} & \text{for 100 mm thick } P_b \text{ front shell} \end{cases}$$

where  $t_{ss}$  is the thickness of the conductive side shell.



(5) With or without outboard end shell

The effect of the existence of outboard end shell can be evaluated by the comparison of the N-functions for model 30 and 50. As shown in Fig. 2.2.13, the N-function is almost the same for these two models. The difference of  $\tau_g$  between these models is a few milliseconds.

Evaluation of shell effect and design of the shell structure

Based on the results of parametric analyses and other design requirements, we designed two candidate shell structures including shield and vacuum chamber for superconducting coils (SC coils): (a) reference shell structure and (b) alternative shell structure. Both structures are designed so that the conductive shells (excluding non conductive components such as shield and vacuum chamber for SC coils) have the stabilizing effect of Approximately 30 msec in growth time of plasma positional instability, according to the previous year's design study. Both shell structures are shown in Figs. 2.2.14 and 2.2.15, respectively.

(1) Number of segments

Since the number of toroidal coils is fourteen, the minimum number of the blanket modules should be twenty eight from the view point of assembling and disassembling by remote handling. Consequently, the number of the conductive shell segmentation can not be less than 28, unless the conductive shells are connected with each other at some points in the toroidal direction. However, the connections between the adjacent conductive shells are prohibited by the specifications of FY '83 FER design, since the remote handling in the torus with narrow access ports would not be reliable. Therefore, we chose the shell segmentation of 28 to maximize the stabilizing effect within our design restrictions.

(2) Inboard shell

The inboard conductive shell is effective to stabilize the plasma positional instability. However, the electro-magnetic forces at plasma disruption would induce large stresses and deformations in the inboard blanket structures, when the conductive shells are set up in the inboard blankets. Figs. 2.2.16 and 2.2.17 show deformations of the blanket structures designed in the previous year, i.e. FY '82 with and without the conductive shells in the inboard blanket modules, respectively. The maximum deformations are 27.24 mm for the blanket module with inboard conductive shell (case 1), and 9.66 mm for that without inboard shell (case 2), respectively. The maximum electro-magnetic force of  $\sim 60 \text{ kgf/cm}^2$  in case 1 inboard module is six times as large as that of  $10 \text{ kgf/cm}^2$  in case 2. The maximum stress intensities of case 1 inboard module are  $62.6 \text{ kgf/mm}^2$ ,  $113.3 \text{ kgf/mm}^2$  and  $115.1 \text{ kgf/mm}^2$  in the membrane, bending and membrane plus bending stresses, respectively. On the other hand, those of case 2 are  $19.3 \text{ kgf/mm}^2$ ,  $16.2 \text{ kgf/mm}^2$  and  $30.6 \text{ kgf/mm}^2$ . From these results, we gave up to install the inboard shell in spite of its desirable stabilizing effect.

### (3) Front shell

The material of the front shell are designed to serve as both conductive shell and neutron multiplier. The typical neutron multipliers, i.e. Pb and Be are chosen for this purpose. Since the electrical conductivity of Be is higher than that of Pb, Be is primarily selected as the front shell material. However, the parametric studies on neutronics indicate that the tritium breeding ratio is higher with Pb front shell than with Be one. So the Be front shell is locally replaced by Pb shell at the portion being relatively less effective for plasma stabilization (i.e. around the mid-plane) in the reference shell structure. In the alternative design, to reduce the electro-magnetic forces at plasma disruption, this portion of Pb is designed to serve only as the neutron multiplier by cutting the Pb shell into small blocks and coating their surface with ceramics, for example. In this case, the Pb shell is electrically replaced by the stainless steel blanket wall.

The thickness of Be/Pb shell is set to be 60 mm from the following reasons: (a) the Be/Pb shell of 60 mm thick is considered to be reasonable, based on the parametric study in section 2, (b) the sensitivity of Be/Pb thickness on the tritium breeding ratio is relatively weak at the thickness of about 60 ~ 100 mm, since the thick Pb/Be shell needs to be cooled, which results in decrease in the tritium breeding ratio due to the neutron absorption by the coolant pipe and (c) difficulty of fabrication increases with the Pb/Be shell thickness.

### (4) Side shell

In both reference and alternative shell designs, the thickness of Cu side shell is set to be 30 mm thicker than that used in parametric study to compensate the deterioration of shell property caused by the removal of inboard conductive shell.

### (5) End shell

In the alternative model, the end shell of 30 mm thick is placed to moderate the shell effect reduction due to the removal of Pb shell effect.

The N-functions of these two shell designs are shown in Figures 2.2.18 and 2.2.19. The growth times of plasma positional instability are 44 msec and 32 msec for the reference and alternative models, respectively. The increases in the growth time due to the shield and the vacuum chamber for SC coils are approximately 10 msec for reference model and 5 msec for alternative model. Figures 2.2.20 and 2.2.21 show the shielding and coupling functions with the control coil-A and -B (see the following paragraphs) for both shell models.

### Active control

In this paragraph, we show the results of parametric studies and simulations on the feedback control of plasma positional instability for two shell structures selected in previous paragraph.

### (1) Position of feedback control coil

One pair of poloidal coils is considered to generate the radial magnetic field for simplicity of the analysis. The top and bottom coils are connected in series so that the current directions are inverse to each other. These coils are located outside the toroidal field coils to avoid the remote assembling and disassembling of control coils. Two locations are considered as the candidate. One is at 3.5 m and  $\pm 6.15$  m in the radial and vertical directions, respectively (control coil-A). The other is at 5.7 m and  $\pm 6.5$  m (control coil-B). Figures 2.2.22 (a) and (b) show the magnetic field configurations produced in plasma region by control coil-A and -B, respectively. The control coil-A produces fairly good radial field. However, the mutual inductance of  $6.2 \times 10^{-7}$  H/m between the coil-A and plasma vertical movement is lower than that of  $1.8 \times 10^{-6}$  H/m between the control coil-B and plasma vertical movement. So we chose the control coil-B as a reference.

### (2) Gain-phase diagram of feedback control system

Figures 2.2.23 (a) and (b) show the gain-phase diagrams of open loop transfer functions for the P and PID feedback control system composed of the reference shell structure and the control coil-B. In these diagrams, both  $T_{\text{dead}}$  and  $T_{\text{delay}}$  are assumed to be 1 msec. And  $T_A$  is given by  $T_A = T_D/10$  in Fig. 2.2.23 (b). There are stable regions where the phase is larger than  $-180$  degree in both diagrams. However, the PID controller increases the phase margin as compared with the P controller. This is essentially due to the derivative action.

### (3) Parametric studies on the feedback control

Figures 2.2.24 (a) ~ (d) show the typical simulation results of the plasma vertical position control. The disturbance field is assumed to rise as

$$B_d = B_{do} (1 - e^{-t/\tau_{Bd}}), \quad B_{do} = 10 \text{ Gauss}, \quad \tau_{Bd} = 1 \text{ msec.}$$

The maximum voltage of the control coils is limited to be  $\pm 500$  V in these simulations. The PID feedback control (essentially PD control) reduces the maximum plasma displacement,  $Z_{\text{pMax}}$ , and the power supply capacity (Max. voltage x Max. current) of the control coils,  $P_{\text{PSC}}$ , in comparison with the P feedback control. From Figures 2.2.24(a) and (b), the  $Z_{\text{pMax}}$  and the  $P_{\text{PSC}}$  can be estimated to be approximately  $1.7 \sim 1.9$  mm and  $50 \sim 70$  MVA, respectively for the control system with the reference shell structure and the control coil-B. This power of  $50 \sim 70$  MVA is relatively small compared with that for equilibrium field and ohmic heating field poloidal coils. However, if we would superpose the active control field component on the equilibrium field component, the plasma vertical position control will increase the requirement for the overall poloidal power supply capacity to the intolerable level, since the maximum coil voltage comes from the requirement by the positional control, and the maximum coil current comes from that by plasma equilibrium. In our FER design, the maximum coil voltage for positional control is approximately 10 times as high as that for plasma start up, and the maximum coil current

for plasma equilibrium is approximately 10 to 100 times as large as that for positional control. Consequently, we install the active control coils separately from the equilibrium field coils. In case of the control coil-A, the  $P_{psc}$  is approximately 1.5 times as large as that in case of the control coil-B for the reference shell structure, though the  $Z_{pMax}$  is almost the same. On the other hand, the  $Z_{pMax}$  in case of the alternative shell structure is approximately 1.5 times as large as that in case of the reference one, though the  $P_{psc}$  is almost the same.

Table 2.2.4 summarizes the sensitivity studies of P feedback control for the parameters of uncertainty of position detector,  $Z_{qnt}$ , rise time constant of disturbance field,  $\tau_{Bd}$ , delay time of position detector,  $T_{delay}$ , dead time of the thyrister,  $T_{dead}$ , and coil voltage limiter,  $V_L$ . In these studies, the quantitizer is put into the block diagram to estimate the effect of accuracy of position detection. The quantitizer generates a signal of  $k \cdot Z_{qnt}$  for the value from  $(k - \frac{1}{2})Z_{qnt}$ . Here,  $k$  is integer, and the infinitely accurate detector is represented by  $Z_{qnt}=0$ . The sensitivity of these parameters on the  $Z_{pMax}$  and the power supply capacity is fairly weak except  $\tau_{Bd}$  and  $V_L$ . The voltage limiter makes the sensitivity of these parameters weak furthermore. From this table and Fig. 2.2.24 (a), it can be noticed that the power supply capacity is approximately proportional to the maximum voltage, though the excessively low voltage limitation makes the control quality poor as shown in Fig. 2.2.25.

### Summary

We designed two candidate shell structures based on the sensitivity analyses of the conductive shell effects on plasma vertical position instability. And we performed the simulation studies on the feedback control of plasma vertical movement for the disturbance radial field of 10 Gauss with a rise time of 1 msec. The main conclusions of this work are as follows.

- (1) The growth time of plasma vertical position instability, is roughly proportional to the inverse of the number of the conductive shell segments in the toroidal direction. From the constraint on remote maintenance, the conductive shell is installed in each blanket of twenty eight modules, minimizing the number of the shell segments consistent with remote maintenance.
- (2) Though the conductive shells in the inboard blanket modules are effective to suppress the growth rate of plasma vertical position instability, they would enlarge the electro-magnetic forces induced at plasma disruption in the inboard blanket structures. Because of the restriction on accessibility behind the inboard blanket modules, the supports of the modules would not be so rigid that the inboard blanket structures could withstand the induced forces. So it is desirable to design the shell structure so that the plasma vertical position instability can be suppressed only by the conductive shells in the outboard blanket modules.

- (3) The thickness of Cu side shell and Be/Pb front shell are set to be 30 mm and 60 mm in our FER design, respectively, so that the enough stabilizing effect can be obtained only by the conductive shells in the outboard blanket modules. The Pb/Be front shells serve also as the neutron multiplier. If the Pb/Be front shells are not desired, we can replace them by Cu plane of approximately 4.5 mm thick, for example.
- (4) The PID controller (essentially PD controller) increases the stable region in gain-phase diagram and improves the quality of the feedback control system, as compared with the P controller.
- (5) The power supply capacity does not strongly depend on the shell structure but on the control coil position. On the other hand, the maximum plasma displacement depends mainly on the shell structure.
- (6) The maximum plasma displacement and the power supply capacity do not largely depend on the parameters such as the thyristor dead time, the position detector delay time and the accuracy of plasma position detection. Since the voltage limitation of the control coils does not affect so much the maximum coil current, the power supply capacity is roughly proportional to the maximum coil voltage, though the maximum plasma displacement does not largely depend on it.

## II. INTOR

In the INTOR, it is tried to obtain stable high- $\beta$  plasma with a non-circular cross-section of comparatively large elongation and also both of suppression and control of the vertical position instability are necessary. We had already reported a shell structure producing the shell effect through which short instability growth time could be suppressed by much the same as a few tens of milliseconds, and there was shown a brief description about the feedback control of its instability [11].

Here, parametric calculations using the INTOR plasma parameters as shown on Table 2.2.5 are carried out for both structures of Figures, 2.2.26 and 2.2.27.

Fig. 2.2.26 shows three kinds of the INTOR shell structures, which have already set as one of obligations on the INTOR Workshop, January 1984, and also the numerical values with their parameters of structures have been determined and shown in the Appendix 2 of January session summary.

Fig. 2.2.27 shows the case that the shell structure shown in Fig. 2.2.26(c) is temporarily installed in the INTOR reactor shown in Fig. 2.2.28. This reactor is divided into 24 sectors along the toroidal direction. In Fig. 2.2.27 is shown the thin plate structure of the reactor (a half of the upper part of one sector). The overall structure is shown in Fig. 2.2.27(c). Its structure of the shell-I (corresponding to Fig. 2.2.26(c), which plays an important role in producing shell effect, and the structure of the shell-(I+II) are especially magnified as shown in Figs. 2.2.27, (a) and (b), respectively.

Although the thickness of each part of the shell-I and shell-II is shown in copper, this does not mean that copper is only a material available for the purpose. It was selected merely as an example. Both dimensions of the shell-I and shell-II are shown in reference [11].

Each thickness of the shields and the blanket is shown as follows:

Shield-1 (Inboard Side)	40 cm (SUS)
Shield-2 (Inboard Side)	40 cm (SUS)
Shield-I (Outboard Side)	30 cm (SUS)
Shield-II (Outboard Side)	40 cm (SUS)
Shield-III (Outboard Side)	30 cm (SUS)
Blanket	5 cm (SUS)

The candidate positions of control coils and their cross-section of the simplified reactor structure are shown in Figure 2.2.28. The positions and dimensions of the above mentioned radiation shields and blanket are determined referring to this structure.

$n_s$  values obtained from three cases in Figs. 2.2.26 (a), (b) and (c) had been already given in Reference [11] and found to be about 0.73, 0.64 and 1.52, respectively. In this case, the dipole model was used for representation of the INTOR plasma column, and the radial position of dipole was selected the INTOR plasma major radius, 5.3m. Field index,  $n$  with the INTOR high- $\beta$  plasma is about -1.3 and therefore useful shell effect can be not obtained from both of (a) and (b) in Fig. 2.2.26, if based on the dipole model. Only the shell structure, (c) is found to give a useful shell effect.

Figure 2.2.29 shows  $n_s$  values obtained from the filamentary model and there, both of (a) (called Plate type) and (c) (called Rectangular type) are given together on selecting the radial position as a variable. Solid lines indicate the sum,  $\sum n_i$  (more in  $n_i$ -value than  $(\times 10^{-3})$ ) and dotted lines are given for the maximum  $n_s$ . It is found from the figure that  $n_s$  value obtain from the filamentary model is larger than one from the dipole model but more  $n_s$  value in  $\sum n_i$  (the maximum  $n_s$ ) than 1.0 can be not obtained before more major radius than 5.4m (5.5m) is selected. Judging from these results, the above shell of the plate type, (c) in Fig. 2.2.26 may be used for the INTOR high- $\beta$  pump limiter plasma.

In Figure 2.2.30 is shown the time dependence of the displacement of the plasma column  $z_p$ , along the axis (in case of being not controlled), when the external disturbance,  $B_d = B_0 [1 - \exp(-t/\tau_d)]$ , is imposed with  $B_0$  being fixed at 10 Gauss and  $\tau_d$  being varied from 0.1 ms to 1 ms, in the case of Fig. 2.2.27. It is found from the value of  $z_p$  after 1 ms that the time evolutions is  $z_p$  from different  $\tau_d$  show almost the same growth time, about 40 ms. It is also shown that  $z_p$  can be kept less than or nearly equal to 2 cm due to the shell effect at the time of several ms after the external disturbance was imposed.

Figure 2.2.31 is obtained from the case that the decay index is -1.3 and that the torus is toroidally divided into 24 sectors. In Fig. 2.2.31 is shown the time dependence of  $z_p$  when an external disturbance, of which  $\tau_d = 1.0$  ms and  $B_d = 10$  Gauss, is imposed, where the number of toroidally divided sectors is given as a parameter. The plasma column growth time,  $\gamma_g$  is approximately 50 ms, 40 ms and 30 ms, when the number of sectors is 14, 24 and 34, respectively.

In Figure 2.2.32 is shown the plasma column growth time,  $\gamma_g^{-1}$ , expected in the reactor structure given as a function of the thickness of a pair of plates placed along the toroidal direction at the upper and lower part of shell-I. In this case the number of sectors is set to be 24.

The shell-I gives the shell effect caused only by the structure shown in Fig. 2.2.26 (a), and the shell-(I+II) gives the shell effect caused only by the structure shown in Fig. 2.2.26 (b). Comparison of these two cases shows that the structure of the shell-II can be expected to give more shell effect than in case of only the shell-I. The upper two cases in Fig. 2.2.32 correspond to the case shown in Fig. 2.2.27 (c). The case that the structure of the shell-II is removed from the structure shown in Fig. 2.2.27 (c), has more shell effect than in Fig. 2.2.27 (b) but less shell effect than in Fig. 2.2.27 (c). These show that substantial shell effect due to the shell-II can be expected, too.

The open loop transfer function,  $A(s)$  is examined first before the vertical position control is studied. In this discussion, the reactor structure with 24 sectors is selected, and the thickness of the shell structure is set to be equal to the value shown in Fig. 2.2.27 (a) and (b).

The Bode diagram are shown in Figure 2.2.33 for the case of PI control. The delay time at the detector,  $T_{\text{delay}}$ , and the dead time for the power supply,  $T_{\text{dead}}$ , are taken as parameters. It can be seen that PI control can give a stable system if  $T_{\text{delay}} = T_{\text{dead}} = 0$ , but that it does not give a stable system if  $T_{\text{delay}} \neq 0$ , as the phase is shifted less than  $-180^\circ$  for  $\omega \geq 10^3$  (1/sec.).

The Bode diagram is shown in Figure 2.2.34 for the case of PID control. It can be seen that the phase is shifted less than  $-180^\circ$  for  $\omega \approx 10^3$  (1/sec) if  $T_{\text{delay}} = 5$  ms, but that stable control is possible if  $T_{\text{delay}} = 4$  ms, as more phase margin than  $30^\circ$  is given.

The typical control results for the case  $T_{\text{dead}} = T_{\text{delay}} = 2$  ms are given in Figure 2.2.34. The PID control is started at  $t=1$  ms. The value of  $z_p$  reaches the maximum at about 3 ms, and then it fluctuates slightly to be attenuated and stabilized. The coil voltage is started to increase rapidly at 3 ms and reaches the maximum value of 2.0 KV at about 4.7 ms, and then it fluctuates to be attenuated. The coil current is slightly flowed prior to the time at 3 ms, increase to reach the first peak at about 6 ms, gets the maximum of approximately 0.14 MA at about 20 ms, and then is stabilized. In this case, the power

supply of approximately 300MVA is required for suppression and control of the vertical instability.

In Figure 2.2.35, the time dependence of  $z_p$  is shown, where  $T_{\text{delay}}$  is fixed at 2 ms and  $T_{\text{dead}}$  is varied as a parameter. Three cases in  $T_{\text{dead}}$ , -1.0, 2.0 and 4 ms are shown there. As  $T_{\text{dead}}$  decreases, the maximum value of  $z_p$  decreases and the fluctuation is moderated, attaining more stable control. The current required for the control coils in three cases is not much varied, and the voltage required decreases as  $T_{\text{dead}}$  decreases. The required power supply is approximately 230 MVA when  $T_{\text{dead}} = 1.0$  ms.

The survey of  $T_{\text{delay}}$  is also considered to be required. The installation position of the detector should be fully examined in the future, because it can be seen from Fig. 2.2.34 and Fig. 2.2.35 that the choice of the installation position is a critical factor.

As shown in the above, parametric calculations are carried out for the case that the shell structure described in the previous report (11) is actually installed in the simplified reactor structure. One of the aims in these analyses is to know if expected shell effects can be realized in the reactor structure. As a result, it is found that such a reactor structure as in Fig. 2.2.27 (c) can have the instability growth time, about 40 ms as much as in shell-(I+II) of previous report, but only the shell-(I+II) in Fig. 2.2.27 (b) can have about 20 ms in the growth time. This difference is thought to come from that, in this report, the shell effect is evaluated taking various modes of the eddy current into consideration but, in the previous [11], it is evaluated in consideration of only one mode with the maximum time constant. Also it is confirmed that the shell effect increases with increasing toroidal bar thickness of the shell-I, and, therefore, its thickness and material should be selected on trade-off with the breeding ratio of tritium in blankets. Dependence of the shell effect upon the number of toroidally divided sectors shows that three cases, 14, 24 and 34 sectors result in about 47, 36 and 28 ms, respectively.

In the above mentioned reactor, some control characteristics are studied under the disturbance field :  $B_0 = 10$  Gauss and  $\tau_d = 1$  msec, whose field is found to produce the maximum axial displacement, about 2 cm. Although the capacity of power supply required for suppression and control of the vertical instability is dependent upon the dead time of the power supply, the delay time of position detector, etc., it is found to be about 230 MVA if typically selected on  $T_{\text{delay}} = 2$  ms and  $T_{\text{dead}} = 1.0$  ms.

### 2.3 Radial Position Control

#### Formations

The kinetic equation of plasma radial movement is simply expressed as eq. 2.3.1 by using the Shafranov formula [12].



$$M_p \ddot{R}_p = \frac{\mu_0 I_p^2}{2} \left( \ln \frac{8R_p}{A_p} + \beta_p + \frac{\ell_i - 3}{2} \right) + 2\pi R_p I_p B_z \quad (2.3.1)$$

where,  $M_p$  : plasma mass  
 $R_p$  : plasma major radius  
 $A_p$  : plasma minor radius  
 $I_p$  : plasma current  
 $\beta_p$  : poloidal beta  
 $\ell_i$  : normalized internal inductance  
 $B_z$  : vertical field  
 $\mu_0$  : permeability

Circuit equations of plasma current, active coils and eddy currents are represented as follows.

$$(L_p \dot{I}_p) + \sum_i (M_{pi} \dot{I}_i) + \sum_k (M_{pk} \dot{I}_k) + \eta_p I_p = 0 \quad (2.3.2)$$

$$L_i \dot{I}_i + (M_{pi} \dot{I}_p) + \sum_j M_{ij} \dot{I}_j + \sum_k M_{ik} \dot{I}_k + \eta_i I_i = V_i \quad (2.3.3)$$

$$\tau_k \dot{I}_k + (M_{pk} \dot{I}_p) + \sum_i M_{ik} \dot{I}_i + I_k = 0 \quad (2.3.4)$$

where,  $L_p$  : plasma self inductance  
 $\eta_p$  : plasma resistance  
 $M_{pi}$  : mutual inductance between plasma and i th coil  
 $M_{pk}$  : mutual inductance between plasma and k th eddy current mode  
 $L_i$  : self inductance of i-th coil  
 $M_{ij}$  : mutual inductance between i-th coil and j-th coil  
 $M_{ik}$  : mutual inductance between i-th coil and k-th eddy current mode  
 $\eta_i$  : resistance of i-th coil  
 $I_i$  : i-th coil current  
 $I_j$  : k-th coil current  
 $\tau_k$  : time constant of k-th eddy current mode  
 $I_k$  : eddy current of k-th mode

Linearizing eqs. (2.3.1) ~ (2.3.4), the following equations are obtained.

$$2\pi I_p B_v (1 - \frac{1}{\Lambda_0} - n) \delta R_p - \frac{\mu_0 I_p^2}{2a_p} \delta a_p - 2\pi R_p B_v \delta I_p + \frac{\mu_0 I_p^2}{2} \delta \beta_p + \frac{\mu_0 I_p^2}{4} \delta \ell_i + I_p (\sum_i \frac{\partial M_{pi}}{\partial R} \delta I_i + \sum_k \frac{\partial M_{pk}}{\partial R} \delta I_k) = 0 \quad (2.3.5)$$

$$(I_p \frac{\partial L_p}{\partial R_p} + \sum_i I_i \frac{\partial M_{pi}}{\partial R_p} + \sum_k I_k \frac{\partial M_{pk}}{\partial R_p}) \delta R_p + I_p \frac{\partial L_p}{\partial a_p} \delta a_p + L_p \delta \dot{I}_p + I_p \frac{\partial L_p}{\partial \ell_i} \delta \dot{\ell}_i + \sum_i M_{pi} \delta \dot{I}_i + \sum_k M_{pk} \delta \dot{I}_k + \eta_p \delta \dot{I}_p = 0 \quad (2.3.6)$$

$$I_p \frac{\partial M_{pi}}{\partial R_p} \delta R_p + M_{pi} \delta \dot{I}_p + L_i \delta \dot{I}_i + \sum_k M_{ik} \delta \dot{I}_k + \eta_i \delta \dot{I}_i = \delta V_i \quad (2.3.7)$$

$$I_p \frac{\partial M_{pk}}{\partial R_p} \delta R_p + M_{pk} \delta \dot{I}_p + \sum_i M_{ik} \delta \dot{I}_i + \tau_k \delta \dot{I}_k + \delta \dot{I}_k = 0 \quad (2.3.8)$$

where,  $\Lambda_0 = \ln(8R_p/a_p) + \beta_p + (\ell_i - 3)/2$  and  
 $B_v = -\mu_0 I_p \Lambda_0 / (4\pi R_p)$ .

By Laplace transformations of eqs. (2.3.5) ~ (2.3.8), six transfer functions are obtained as follows [13,14].

$$N(s) = \sum_k \frac{s\tau_k}{1+s\tau_k} n_k, \quad n_k = -\frac{I_p M_{pk}^2}{2\pi B_v \tau_k} \quad (2.3.9)$$

$$M(s) = \sum_k \frac{s\tau_k}{1+s\tau_k} m_k, \quad m_k = \frac{M_{pk} M_{ik}}{M_{pi} \tau_k} \quad (2.3.10)$$

$$K_i(s) = \frac{s\tau_i}{1+s\tau_i} \sum_k \frac{s\tau_k}{1+s\tau_k} k_{ck}, \quad k_{ck} = \frac{M_{ik}^2}{L_i \tau_k} \quad (2.3.11)$$

$$K_p(s) = \frac{s\tau_p}{1+s\tau_p} \sum_k \frac{s\tau_k}{1+s\tau_k} k_{pk}, \quad k_{pk} = \frac{M_{pk}^2}{L_p \tau_k} \quad (2.3.12)$$

$$V(s) = \sum_k \frac{s\tau_k}{1+s\tau_k} v_k, \quad v_k = \frac{M_{pk} M_{ik}}{M_{pi} \tau_k} \quad (2.3.13)$$

$$B(s) = \sum_k \frac{s\tau_k}{1+s\tau_k} b_k, \quad b_k = \frac{I_p M'_{pk} M_{pk}}{2\pi R_p B_v \tau_k} \quad (2.3.14)$$

The  $N(s)$  represents the stabilizing effect of eddy currents on plasma radial movement. The  $M(s)$  represents the shielding effect to radial magnetic field of control coils. The  $K_i(s)$  and  $K_p(s)$  represent the effective reductions of control coil and plasma impedances due to eddy currents, respectively. The  $V(s)$  represents the effective reduction of mutual inductance between plasma and coil. The  $B(s)$  represents the vertical magnetic field reduction due to eddy currents induced by plasma current change.  $M'$  means radial derivative of mutual inductance in eqs. (2.3.9)~(2.3.14).

### Simulation results

Figures 2.3.1 and 2.3.2 show the time evolutions of plasma radial position,  $R_p$ , and plasma current,  $I_p$ , for the JAERI FER with and without high conductive shell structures, respectively. In these simulations, the active feedback control of plasma radial position is assumed to be out of operation. Poloidal beta  $\beta_p$ , and normalized internal inductance,  $l_i$ , decrease with the time scale of 5 msec by 40%, and plasma are assumed to be compressed so as to conserve toroidal magnetic flux in the plasma. The high conductive shells increase the time scale of plasma movement. So the presence of good conductors near the plasma would be effective to lead to long current decay times [15].

Figure 2.2.3 represents the simulation result of the feedback control of plasma radial movement in case of high conductive shell installed. The shell model and the locations and current weight of active coils are summarized in Fig. 2.3.4. Though plasma radial position can be recovered, the maximum coil voltage is quite high (approximately 4400 V) and the power supply capacity of active coil system is required to be approximately 2000 MVA which is impractical.

Figure 2.3.5 shows the simulation result of more practical case. In this case, maximum coil current and power supply capacity are approximately 1000 V, 110 KA and 110 MVA, respectively. Though plasma position can not be recovered, the time scale of plasma movement can be approximately doubled, comparing with uncontrolled case, if the initial gap of 20 m between inboard first wall and plasma surface can be held.

Concluding remarks

- i) The presence of high conductive shells around the plasma would be effective to increase current decay time at plasma disruption.
- ii) The impractically large power supply capacity and high coil voltage are required to recover the plasma radial position at disruption.
- iii) Within the practical power supply capacity and coil voltage, the time scale of plasma radial movement may be increased by approximately factor 2, provided the gap of  $\sim 25$  cm between inboard first wall and plasma surface can be initially held.

## 2.4 Plasma Shape Control

Considering the control of plasma shape, it may be helpful to classify the problems into the following two categories.

- a. The problems related to the operation scenario such as how to evolve plasma configuration and make high beta configuration with suppressing peak power and capacity of poloidal coil power supply system.
- b. The problems related to the unexpected perturbations such as how to fix null points and/or separatrix lines at plasma position movements.

The problems of the category a are dealt with preprogramming and/or feedforward control. One of the problems in category a is how to obtain and keep favorable plasma current profile and plasma shape for high beta plasma. Another problem is how to reduce the absolute value of the decay index  $n$ -value to mitigate the requirements for plasma vertical position control. It would be desirable to adopt limiter configurations at early start up phase, since it is difficult to suppress the increase in the absolute  $n$ -index value at low beta phase if divertor configurations are adopted at that time. Another way to ease plasma vertical position control is to decrease plasma aspect ratio and increase triangularity, since low plasma aspect ratio and high plasma triangularity would reduce the absolute  $n$ -index value as well known.

The problems of the category b are dealt with feedback control. It is important problem in this category to control null points and separatrix lines when plasmas move in the radial and/or vertical directions. If we control plasma position, shape, null points and separatrix lines independently, unavoidable interaction will be induced among the control objects mentioned above. Noninteracting control method is developed in JT-60 [16] to avoid this kind of interaction. It would be also necessary to adopt this method in the next generation tokamak reactors.

Table. 2.2.1 Main plasma parameters

Parameters	Values
Major Radius (m)	5.5
Minor Radius (m)	1.1
Aspect Ratio	5.0
Ellipticity (target)	1.5
Triangularity	0.2
First wall minor radius (m)	1.25
Averaged ion temperature (keV)	10
Averaged ion density (m <sup>-3</sup> )	1.36x10 <sup>20</sup>
Averaged effective charge	1.5
Plasma current (MA)	5.3
Safety factor at plasma (effective)	2.5
Toroidal beta (%)	4.0
Poloidal beta	2.3
Initial loop voltage (V)	50 (0.1s)
Error field (inside the plasma) (G)	50

Table. 2.2.2 Summary of conductive shell models used in parametric studies on shell effects.

MODEL NO.	NO. OF DIVISION	INBOARD		OUTBOARD			GROWTH TIME: $\tau_G$ (ms)
		F/W	S/W	F/W	S/W	E/W	
10	14	SS (10)	SS (20)	Pb (60)	Cu (15)	NONE	32.21
20		Pb (60)	Cu (15)				41.04
30	28	SS (10)	SS (20)	Pb (100)	Cu (22.5)	NONE	17.74
31				Pb (60)			20.62
32				Pb (100)			21.90
33				Pb (100)			26.29
40		Pb (60)	Cu (15)	Pb (60)	Cu (15)	Cu (15)	27.62
50		SS (10)	SS (20)				19.73
80		SS (10)	SS (20)			NONE	9.27
90	42	Pb (60)	Cu (15)				17.07

F/W: FRONT WALL      S/W: SIDE WALL      E/W: END WALL  
 FIGURES IN ( ) IS THICKNESS; mm

Table. 2.2.3 Parameters used in parametric studies

1. Number of Conductive Shell Segments  
14, 28, 42
2. Inboard Conductive Shells  
Existence (Pb, 60mm & Cu, 15mm) / None
3. Thickness of Outboard Front Shell  
Pb (100mm) / Pb (60mm)
4. Thickness of Outboard Side Shell  
Cu (22.5mm) / Cu(15mm)
5. Outboard End Shell  
Existence Cu (15mm) / None

Table 2.2.5 Major Parameters of the INTOR

Major Radius, $R_p$ (m)	5.3
Minor Radius, $a_p$ (m)	1.2
Plasma Current, (MA)	6.4
Divertor Plasma	
Elongation, K	1.6
Shafranof's Field, $B_0$ (T)	- 0.50
Decay Index, n	- 1.30
Pump Limiter	
Elongation, K	1.50
Shafranof's Field, $B_0$	- 0.52
Decay Index, n	- 1.0

Table. 2.2.4 Parametric studies on plasma position control  
(Reference design, Control coil-B)

		MAX. VOLT (V)	MAX. CURRENT (kA)	P/S CAPACITY (MVA)	MAX. DIS- PLACEMENT (mm)
$V_L = 250 \text{ V}$	$Z_{\text{gnt}}$ (mm)	0.0	118.5	72.57	-18.27
		2.0	118.5	74.67	-18.24
		5.0	120.7	84.50	-18.32
	$\tau_{\text{Bd}}$ (msec)	1.0	118.5	72.57	-18.27
		2.5	117.1	62.50	-15.59
		5.0	112.5	49.21	-12.64
	$T_{\text{delay}}$ (msec)	0.5	115.2	70.87	-17.83
		1.0	118.5	72.57	-18.27
		2.0	125.9	74.49	-18.92
	$T_{\text{dead}}$ (msec)	0.5	115.3	67.80	-17.62
		1.0	118.5	72.57	-18.27
		2.0	126.5	83.69	-19.56
$V_L = 250 \text{ V}$	$Z_{\text{gnt}}$ (mm)	0.0	132.4	33.09	-19.08
		2.0	132.0	33.01	-19.05
		5.0	133.1	33.28	-19.17
	$\tau_{\text{Bd}}$ (msec)	1.0	132.4	33.09	-19.08
		2.5	128.3	32.08	-17.47
		5.0	121.4	30.35	-14.62
	$T_{\text{delay}}$ (msec)	0.5	130.4	32.60	-18.90
		1.0	132.4	33.09	-19.08
		2.0	135.8	33.95	-19.35
	$T_{\text{dead}}$ (msec)	0.5	129.2	32.30	-18.66
		1.0	132.4	33.09	-19.08
		2.0	138.6	34.66	-19.90

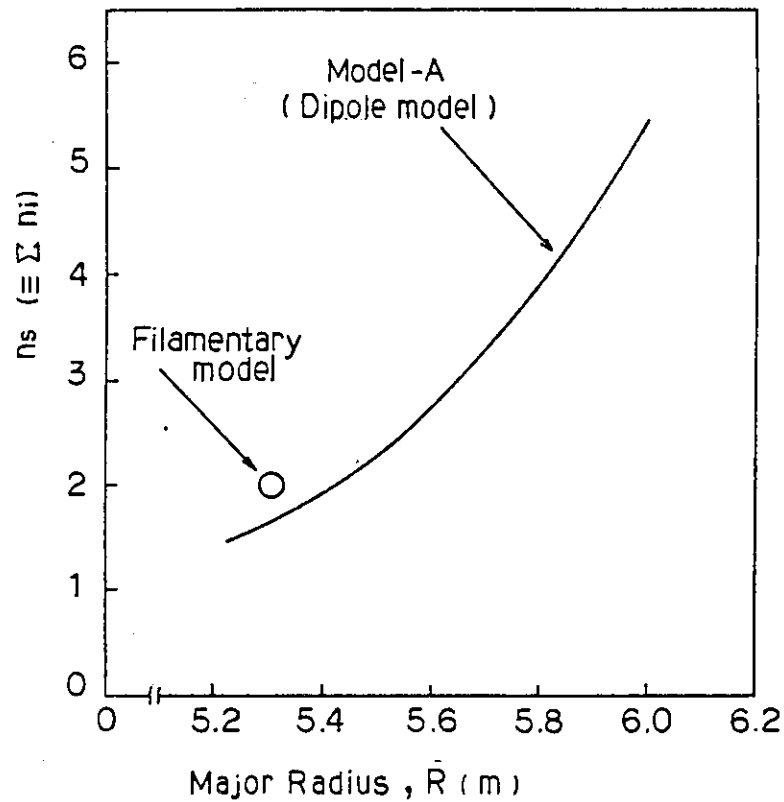


Fig. 2.2.1 Comparison of stabilizing property between filamentary model and dipole current model.

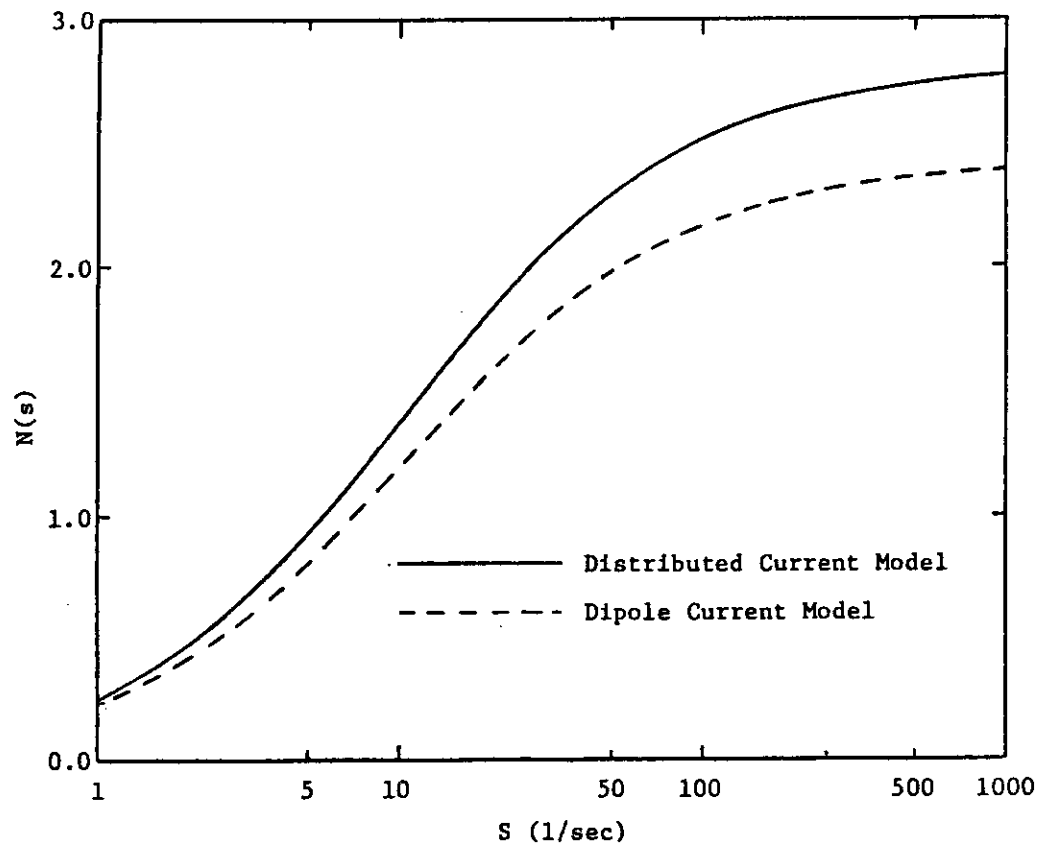


Fig. 2.2.2 Comparison of stabilizing property between distributed current model and dipole current model.



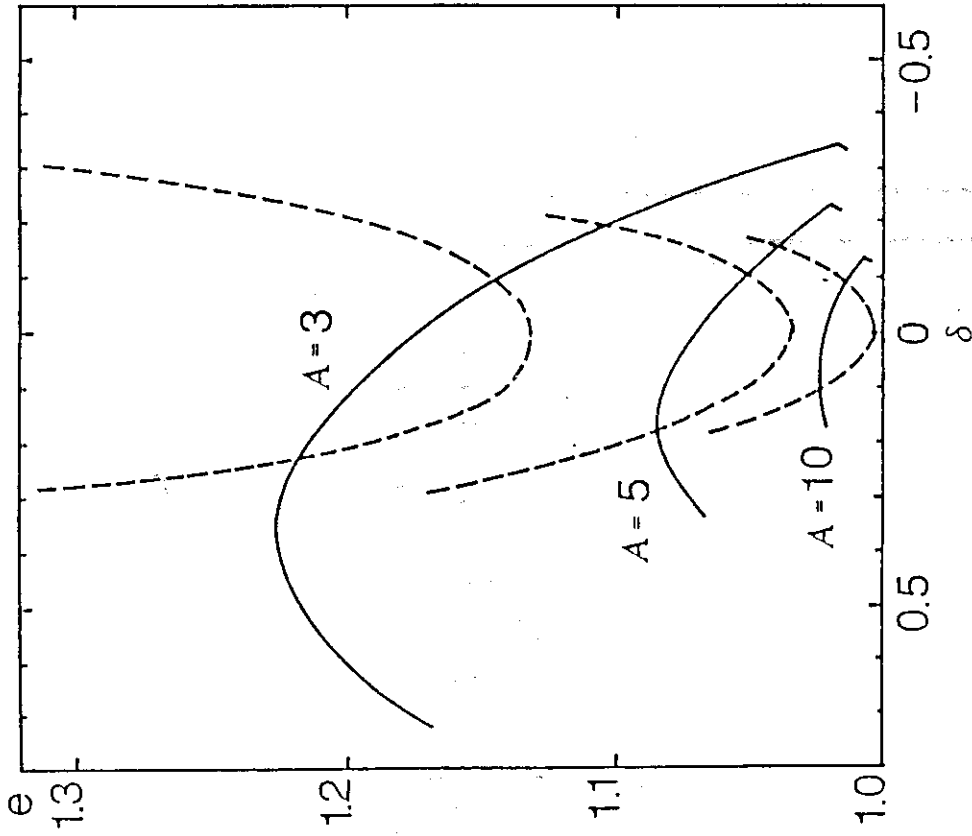


Fig. 2.2.3 The stability diagram of the Solov'ev equilibrium in  $e$ - $\delta$  plane for different values of  $A$ . The safety factor at the magnetic axis,  $q_0$ , is unity. The dashed line and the dashed solid line correspond to ones for  $n_i=0$  and  $n_i=1.5$ , respectively.

Fig. 2.2.4 The stability diagram for the full M2D model (solid lines) and the rigid displacement model (dashed lines).

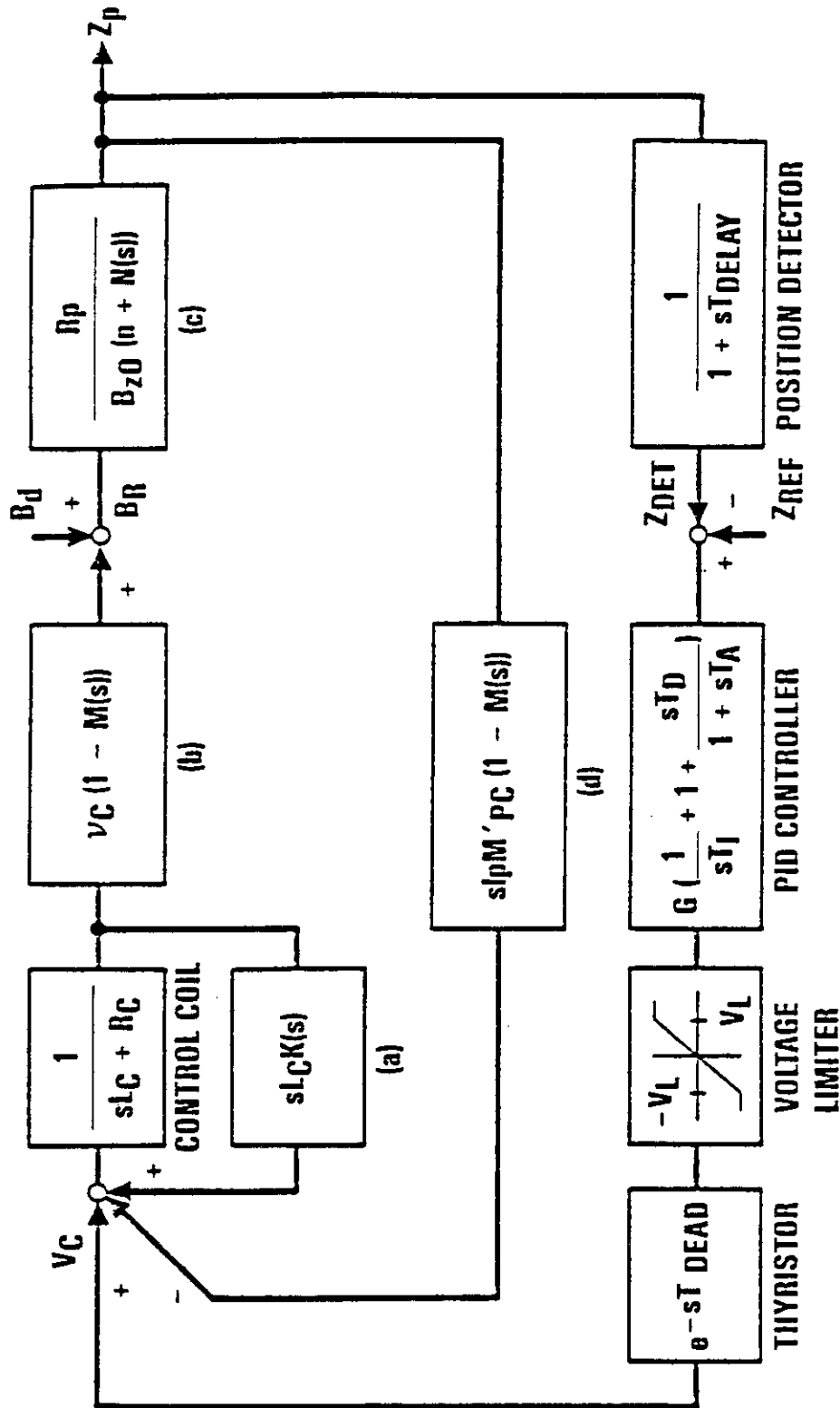


Fig. 2.2.2.5 Block diagram of the feedback control system of the plasma vertical position.

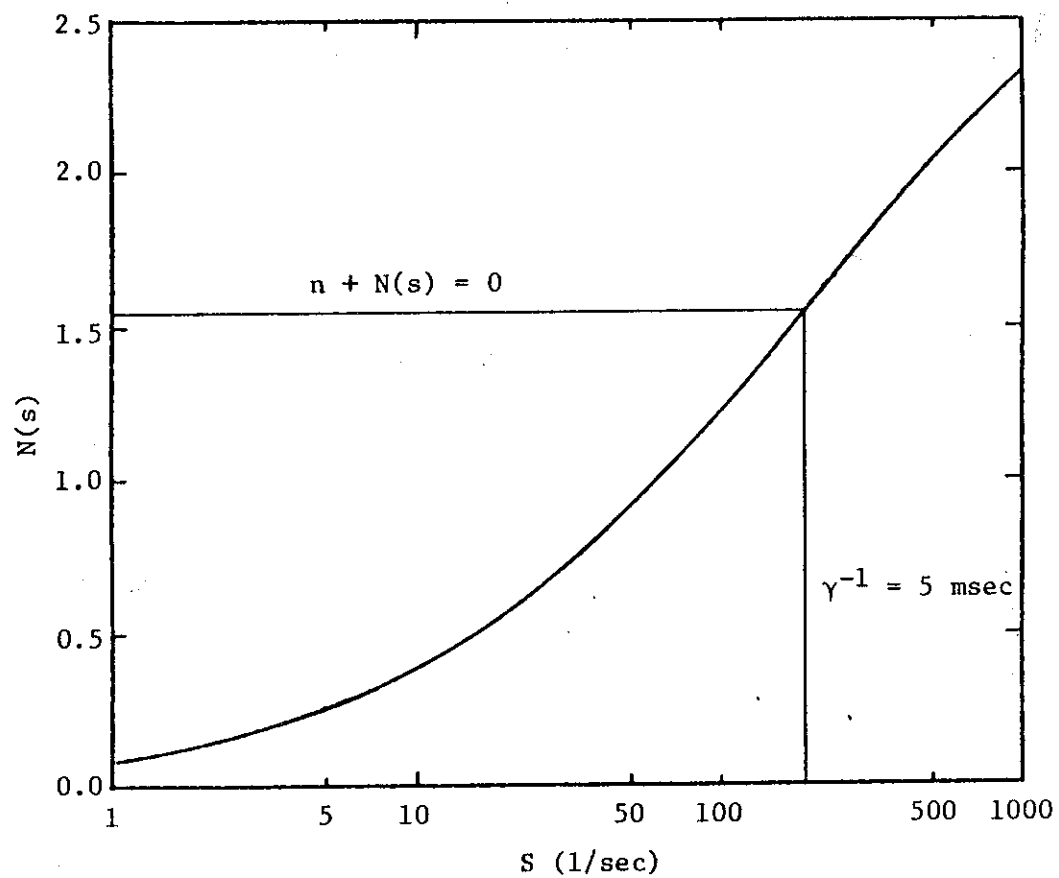


Fig. 2.2.6  $N(s)$ -function of FWBS without high conductive shells

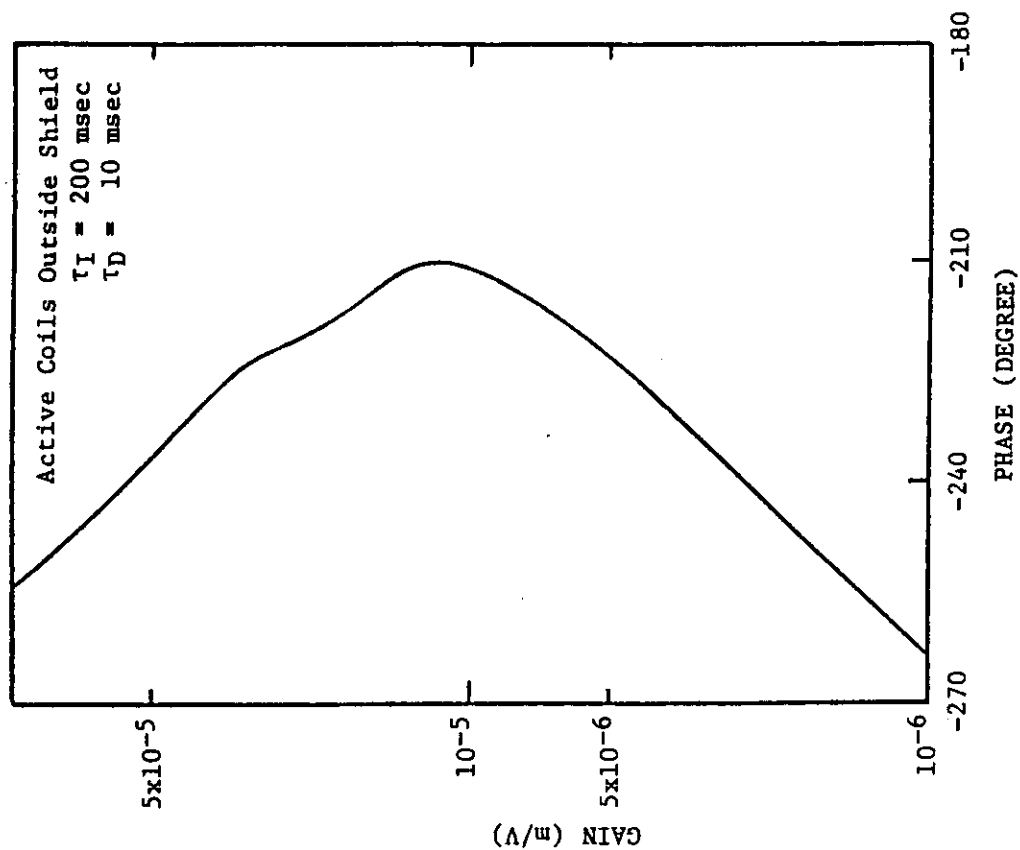


Fig. 2.2.7(b) Gain-phase diagram of open loop transfer function for PID control system without conductive shells. Active coils are located outside shield. Delay time of position detector and dead time of thyristor are assumed to be 1 msec. Integral time constant and derivative time constant are 200 msec and 10 msec, respectively.

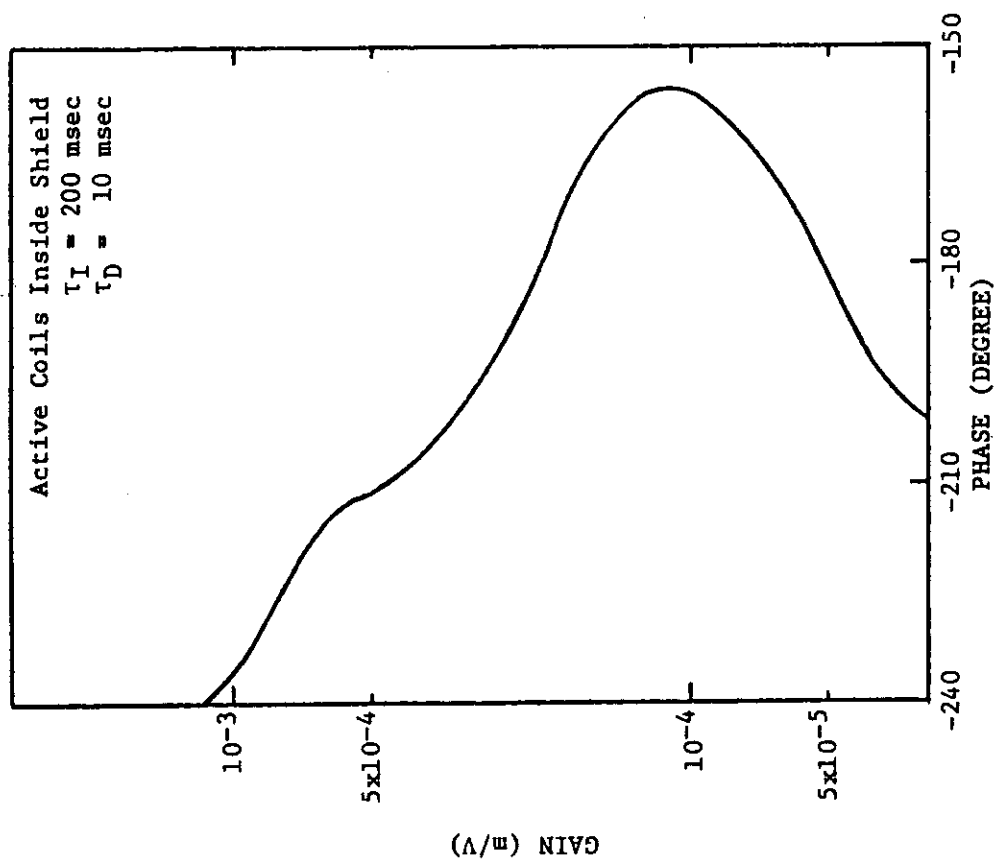


Fig. 2.2.7(a) Gain-phase diagram of open loop transfer function for PID control system without conductive shells. Active coils are located inside shield. Delay time of position detector and dead time of thyristor are assumed to be 1 msec. Integral time constant and derivative time constant are 200 msec and 10 msec, respectively.

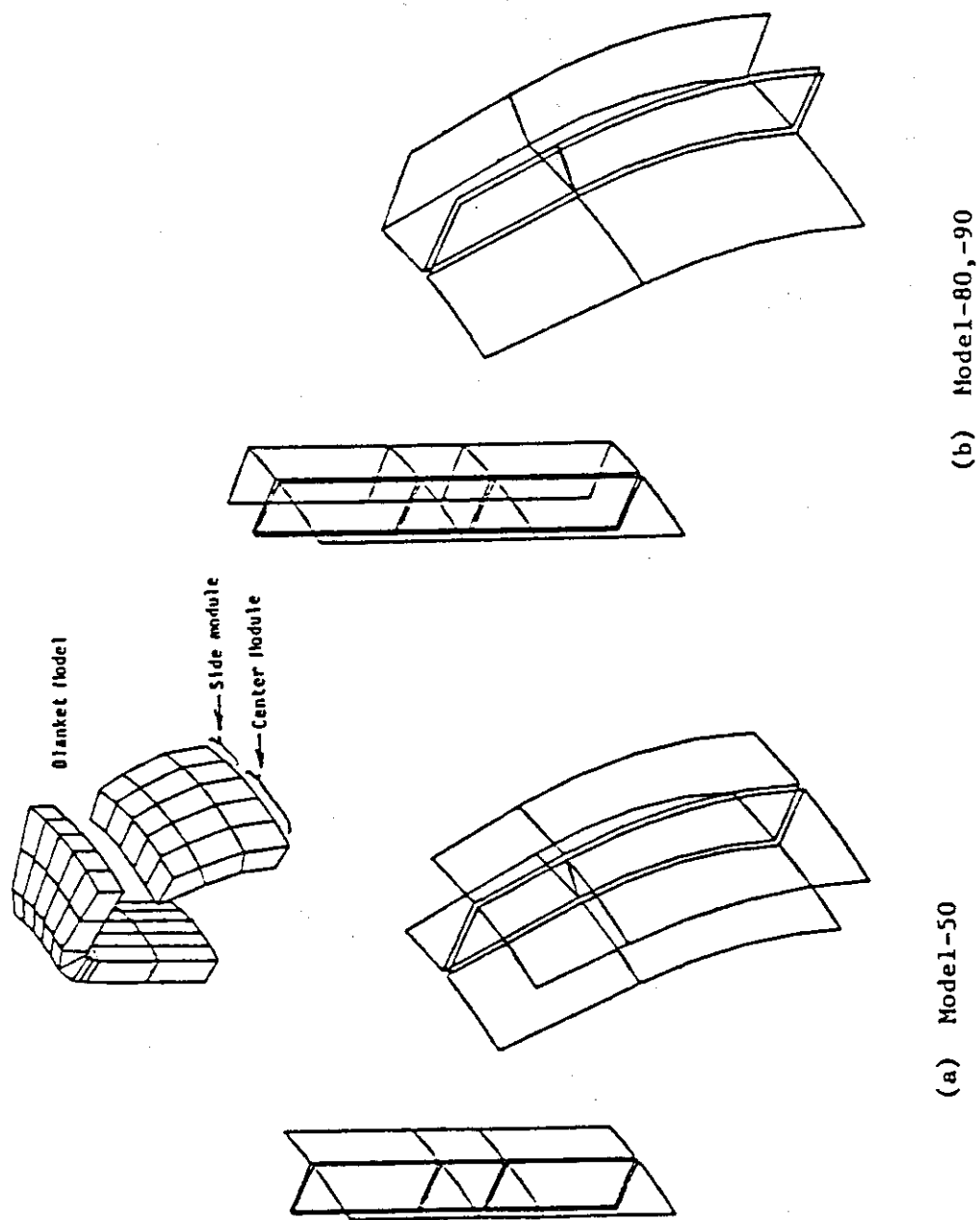


Fig. 2.2.8 Conductive shell structures used in parametric studies on shell effects.

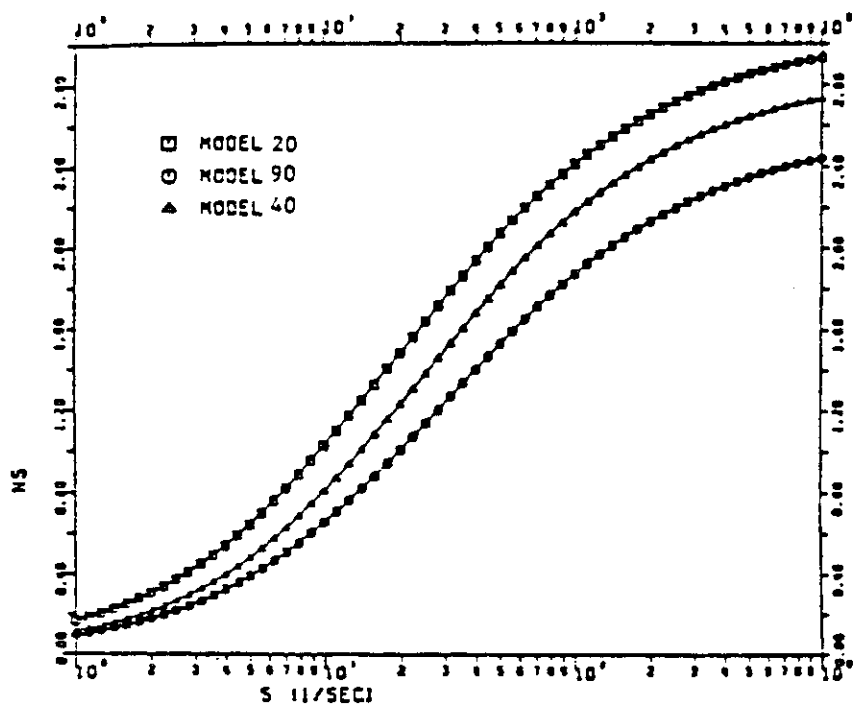


Fig.2.2.9 N-function of conductive shell in both inboard and outboard blanket module with parameter of the number of shell segments. Model 20, 90, 40, : 14, 28, 42 segments

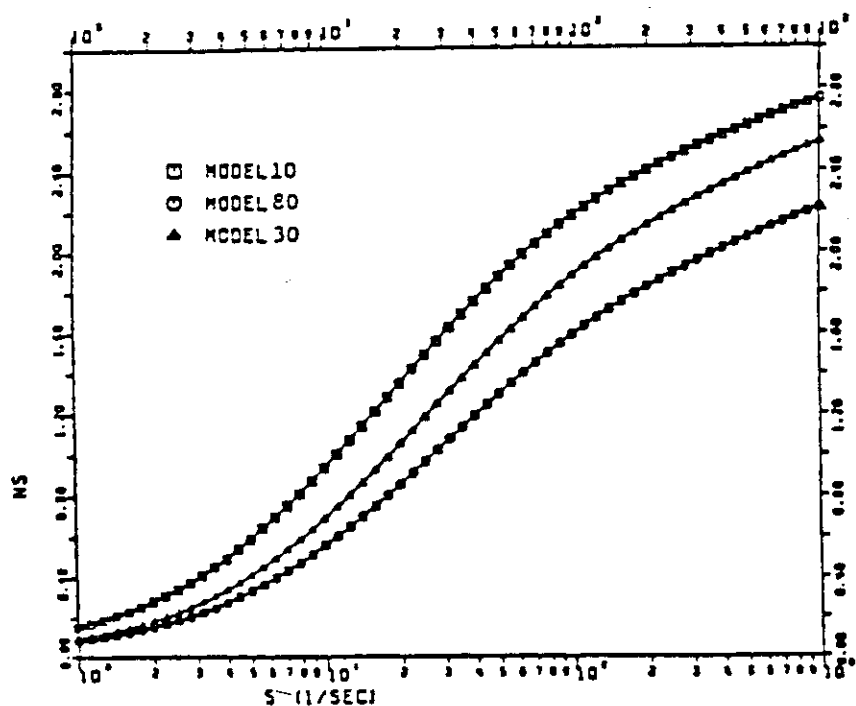


Fig. 2.2.10 N-function of conductive shell in only outboard blanket module with parameter of the number of shell segments. Model 10, 80, 30 : 14, 28, 42 segments

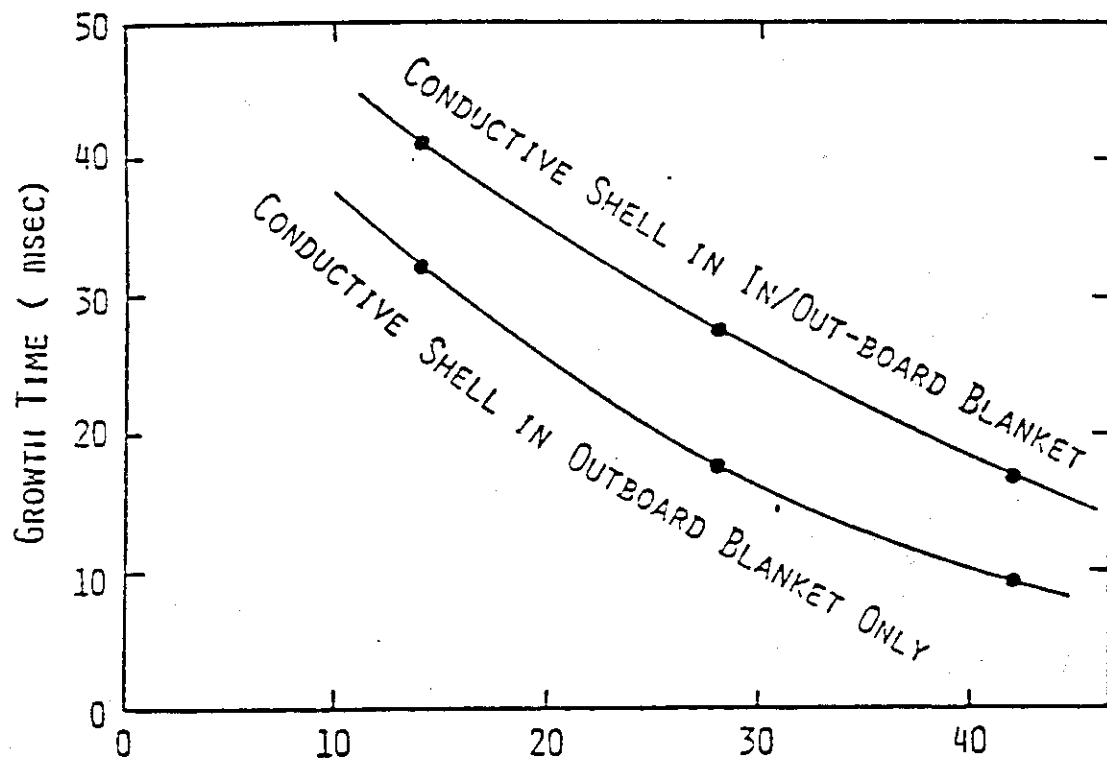


Fig. 2.2.11 Growth time vs. the number of shell segments.

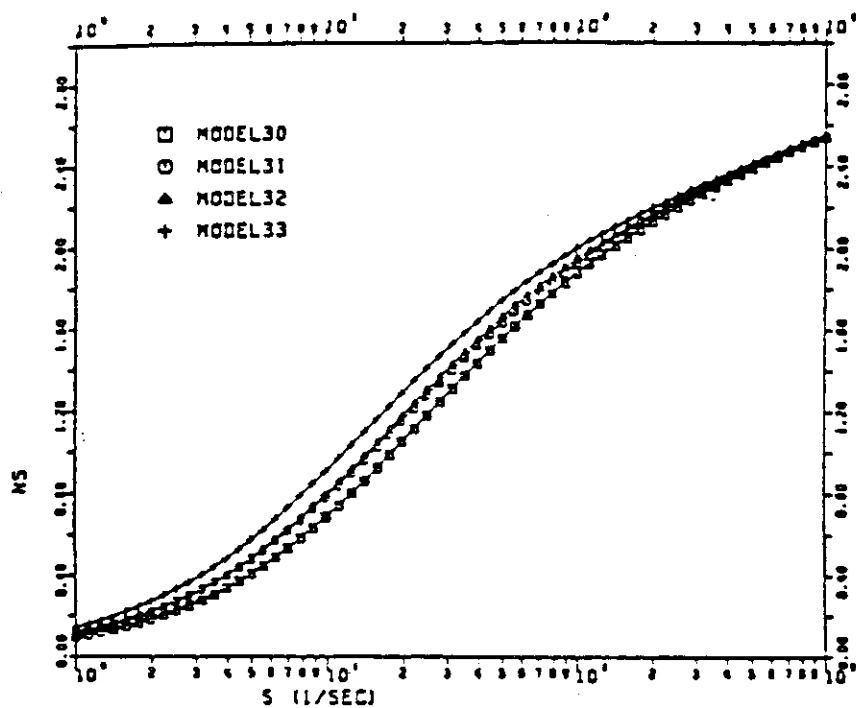


Fig. 2.2.12 N-function of conductive shell with parameter of shell thickness. Model 30, 31, 32, 33 : Pb(60mm), Cu(15mm)/ Pb(100mm), Cu(15mm)/ Pb(60mm), Cu(22.5mm)/ Pb(100mm), Cu(22.5mm)

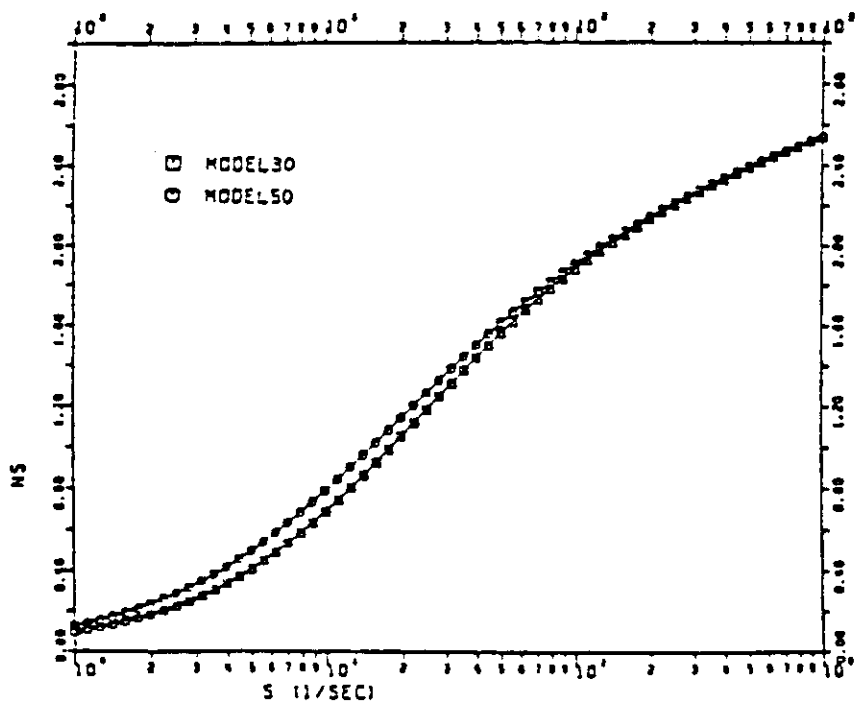


Fig. 2.2.13 Comparison of the N-function between the conductive shell with and without end shell.  
Model 30, 50 : with end shell, without end shell



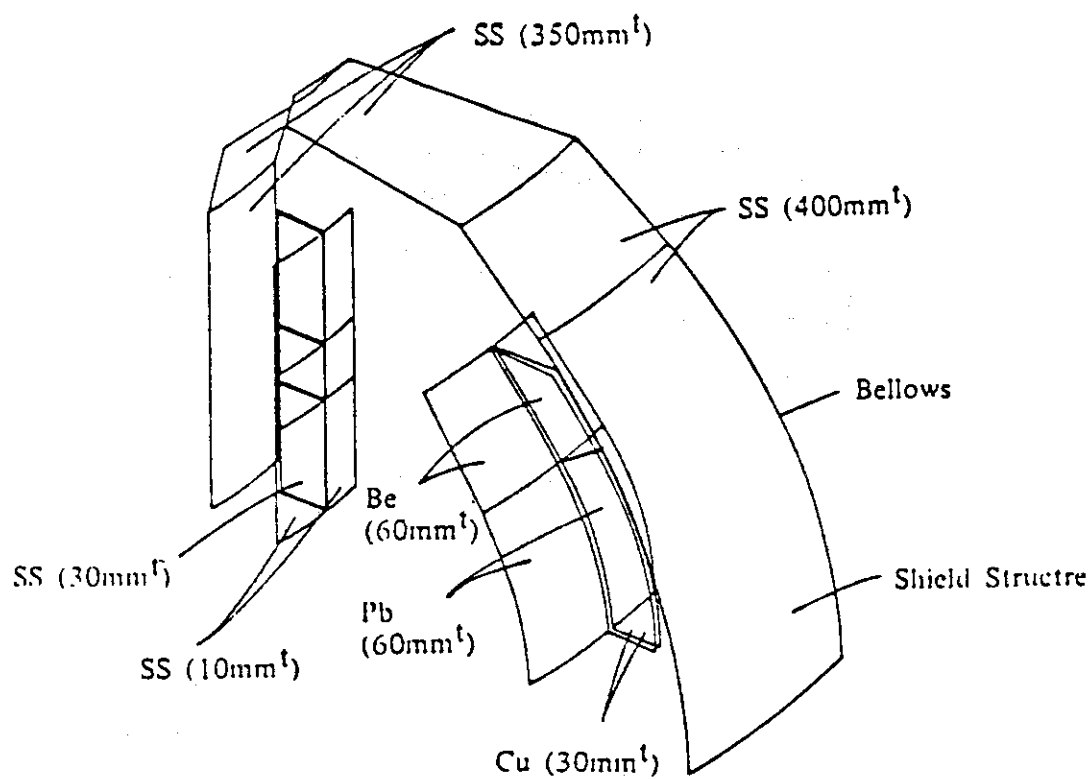


Fig. 2.2.14 Reference shell structure

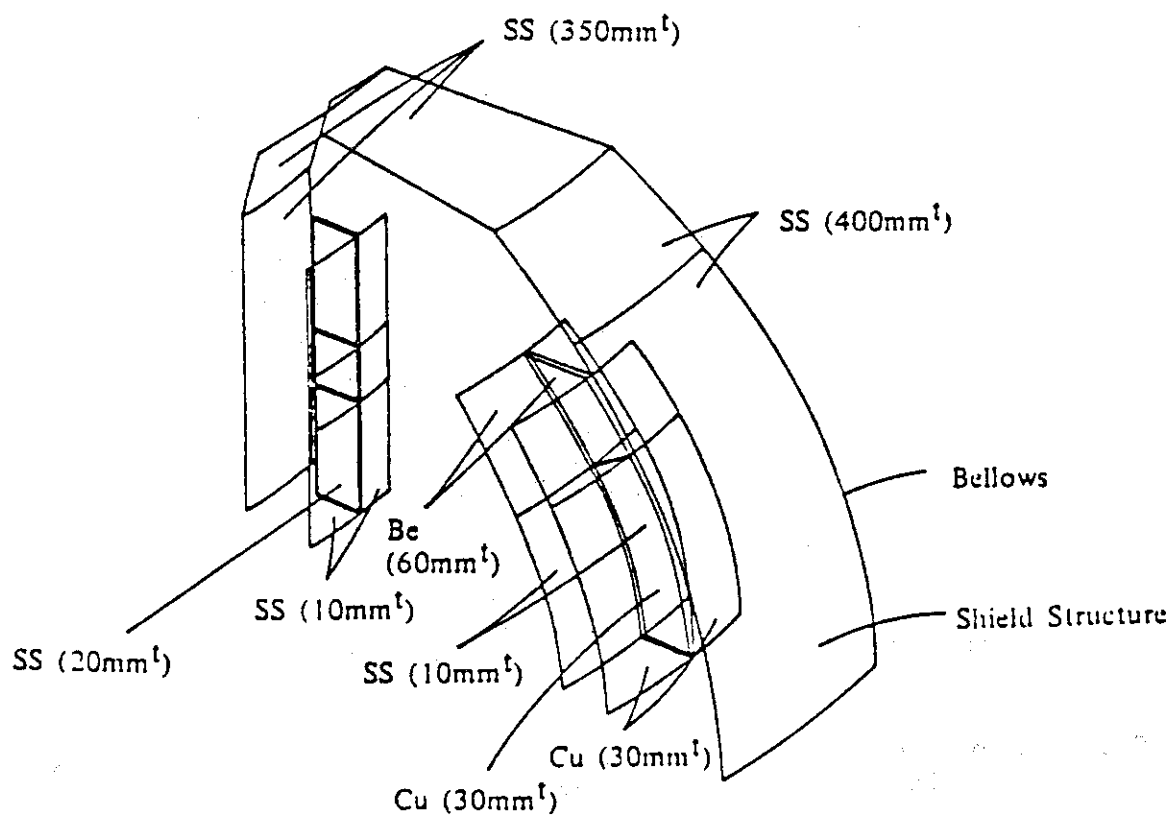


Fig. 2.2.15 Alternative shell structure

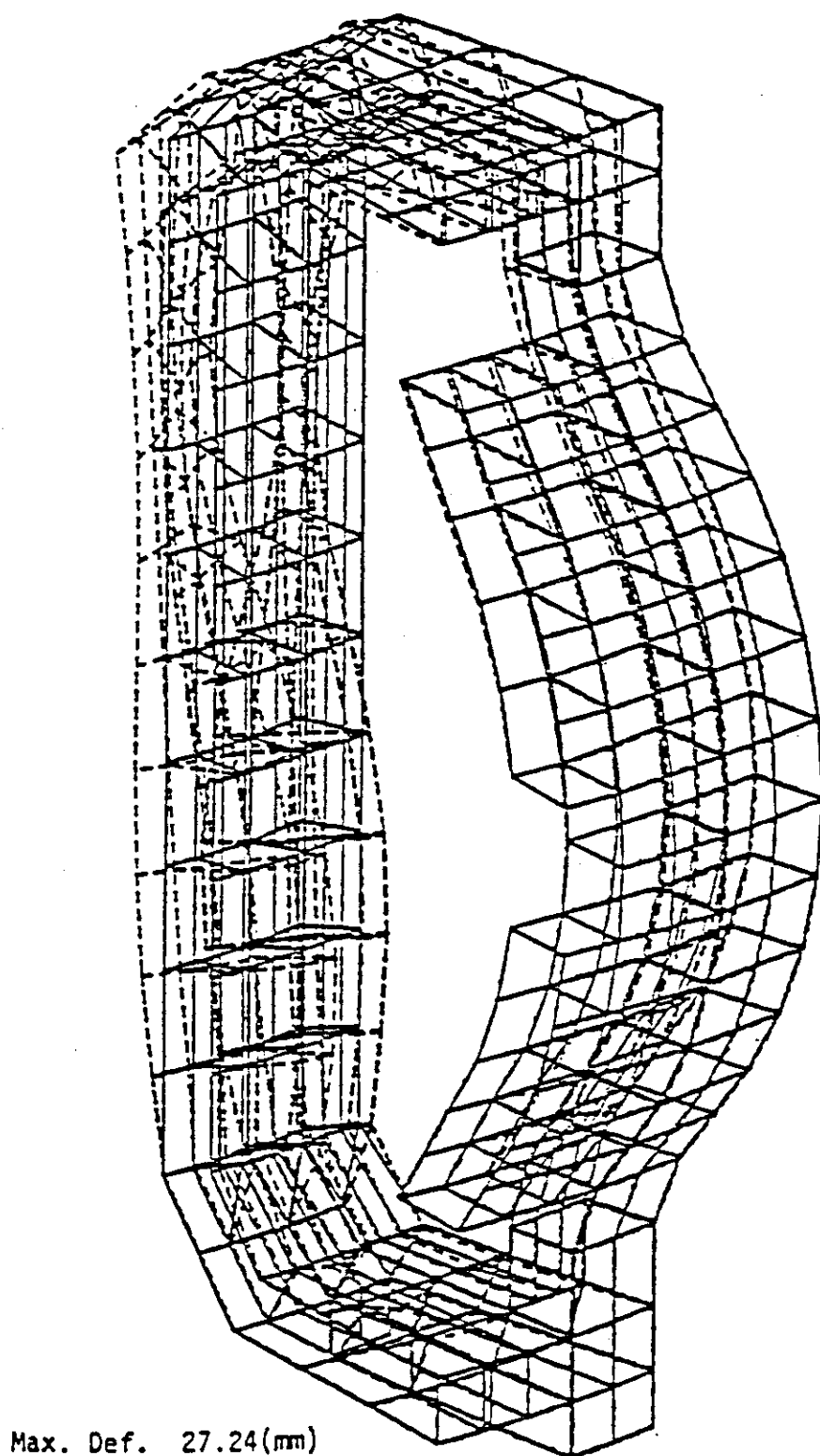


Fig. 2.2.16 Deformation of blanket structure with inboard conductive shell, designed in previous year (FY' 82).

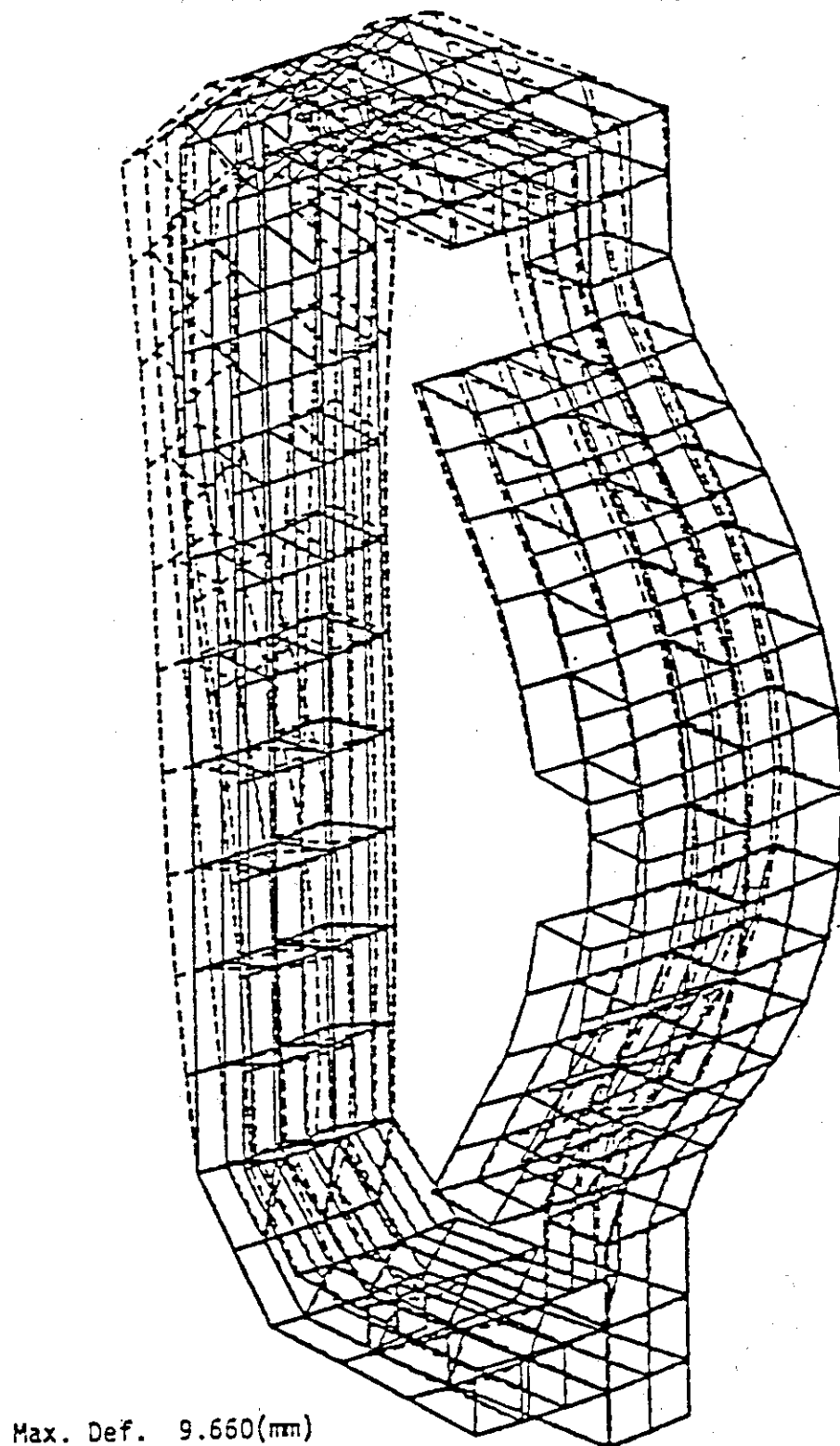


Fig. 2.2.17 Deformation of blanket structure without inboard  
conductive shell, designed in previous year (FY'82).

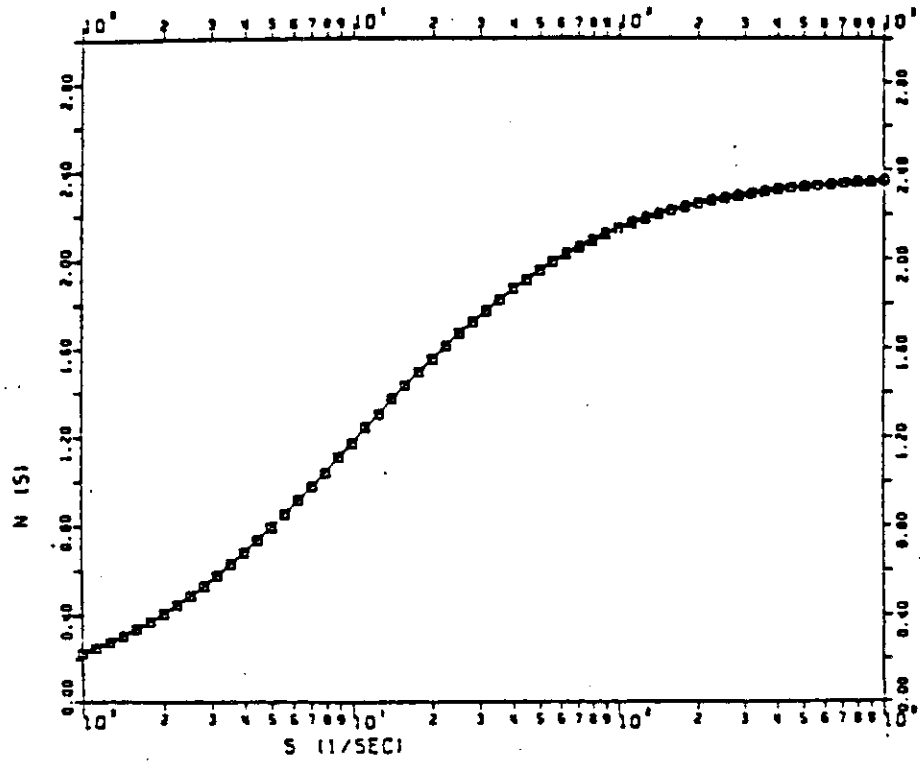


Fig. 2.2.18 N-function of reference shell structure.

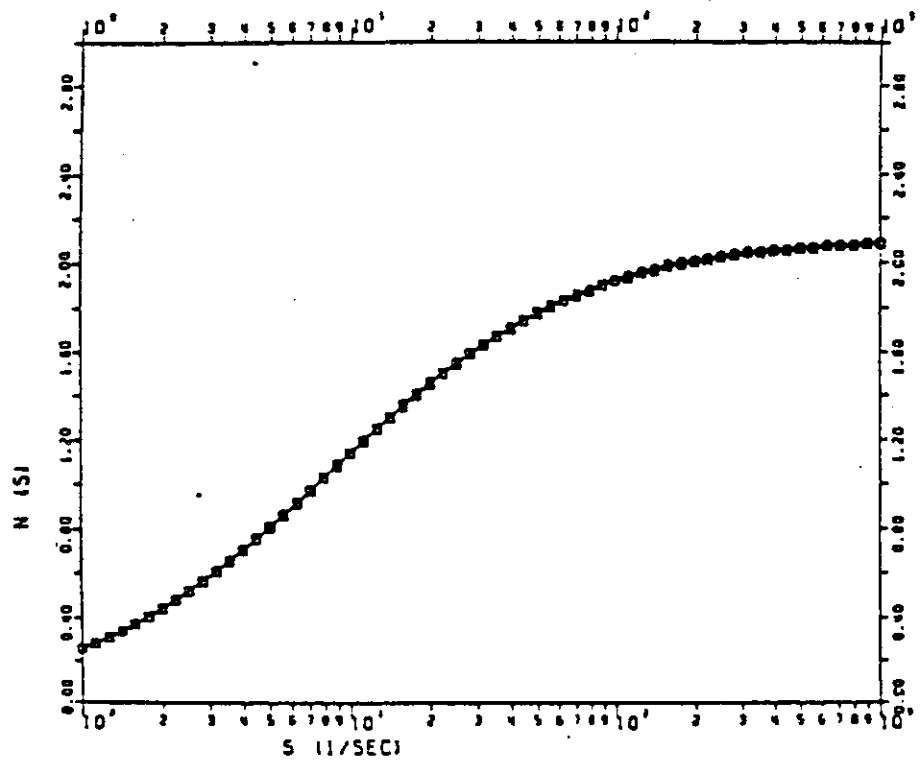


Fig. 2.2.19 N-function of alternative shell structure.

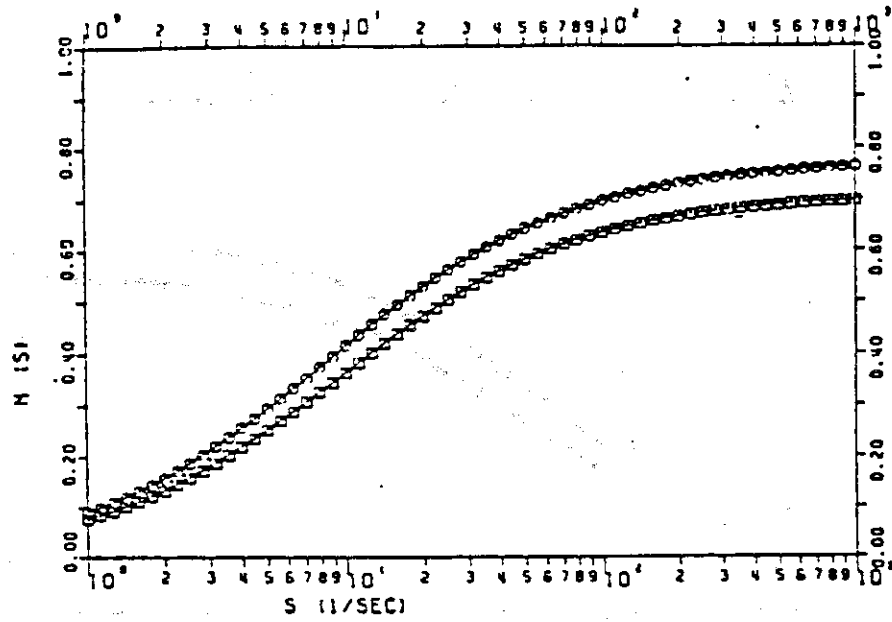


Fig. 2.2.20(a) Shielding function of reference shell structure.

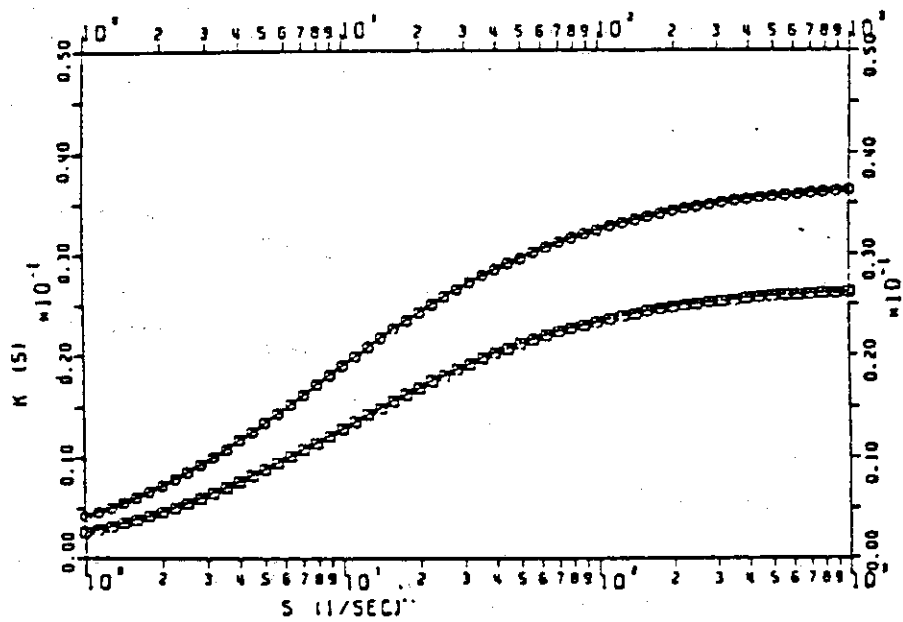


Fig. 2.2.20(b) Coupling function of reference shell structure.

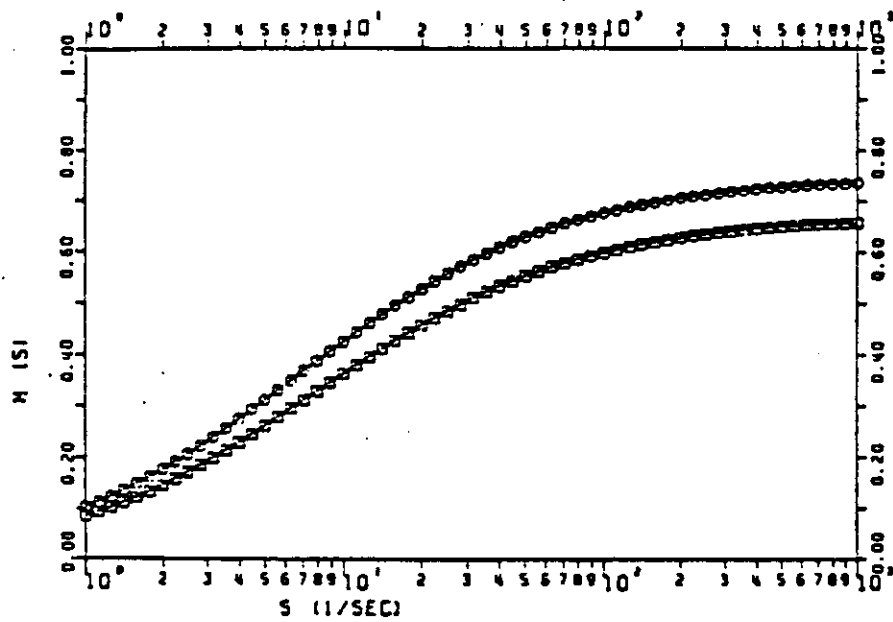


Fig. 2.2.21(a) Shielding function of alternative shell structure.

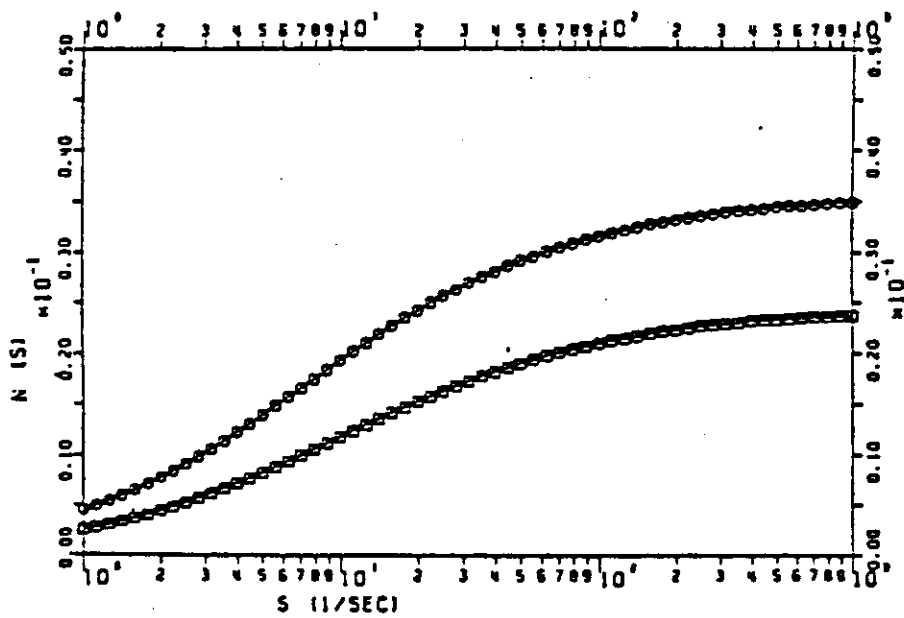


Fig. 2.2.21(b) Coupling function of alternative shell structure.

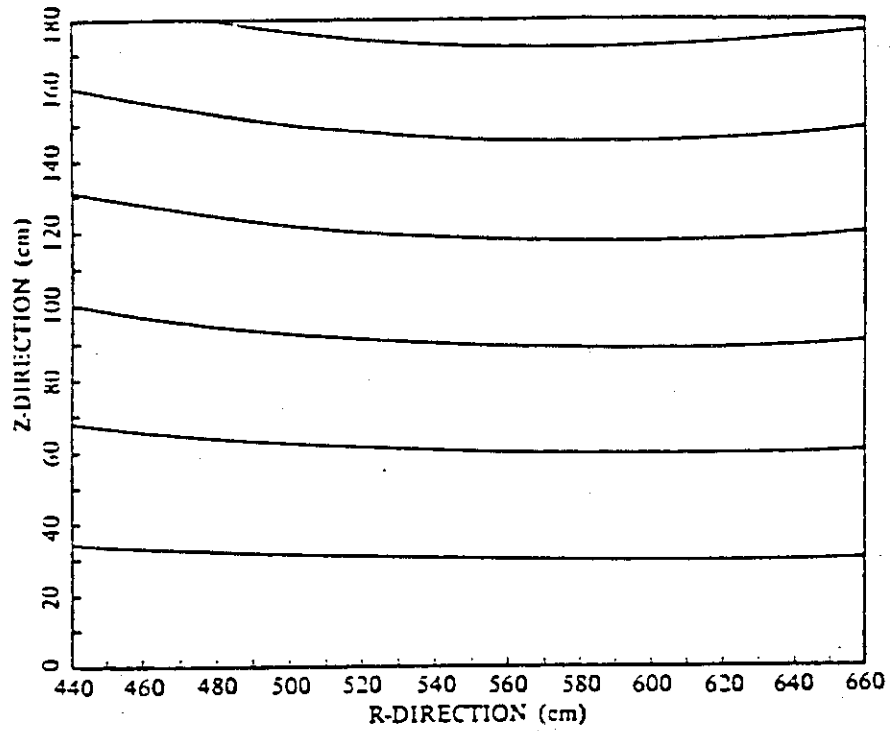


Fig. 2.2.22(a) Magnetic field configuration by control coil-A located at 3.5 m and  $\pm 6.15$  m in the radial and vertical direction.

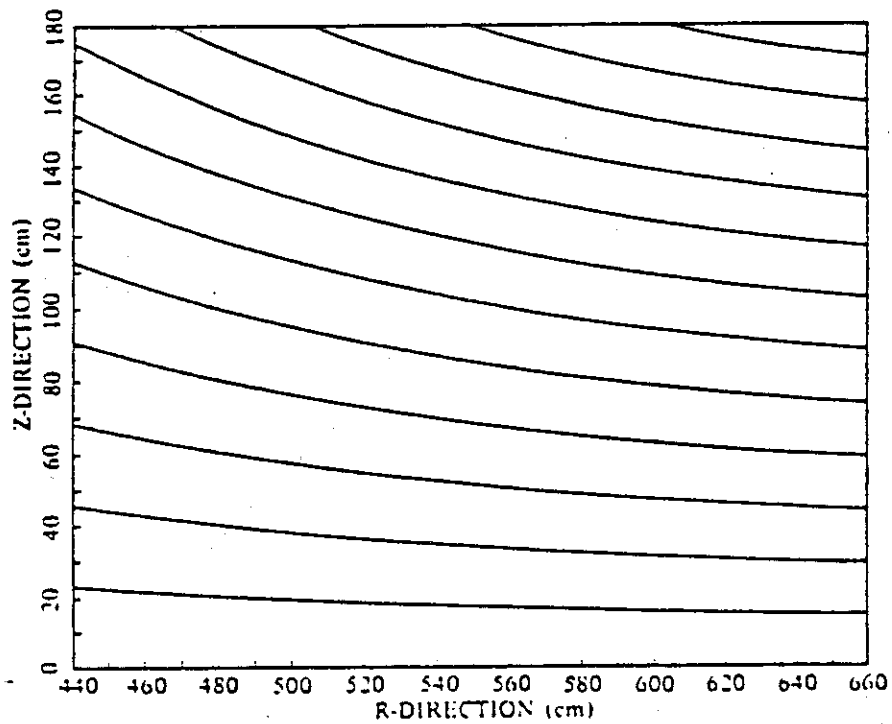


Fig. 2.2.22(b) Magnetic field configuration by control coil-B located at 5.7 m and  $\pm 6.5$  m in the radial and vertical direction.

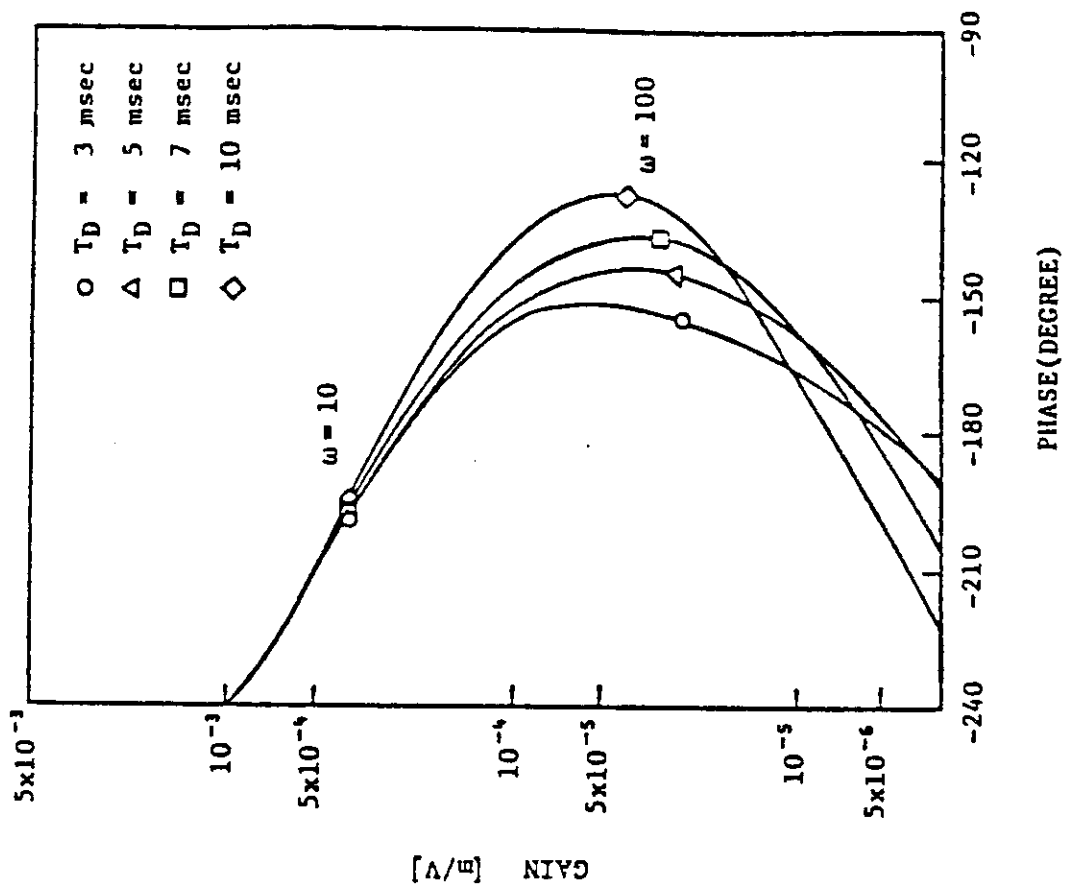


Fig. 2.2.23(b) Gain-phase diagram of open loop transfer function for PID feedback control system with the alternative shell structure and control coil-B.

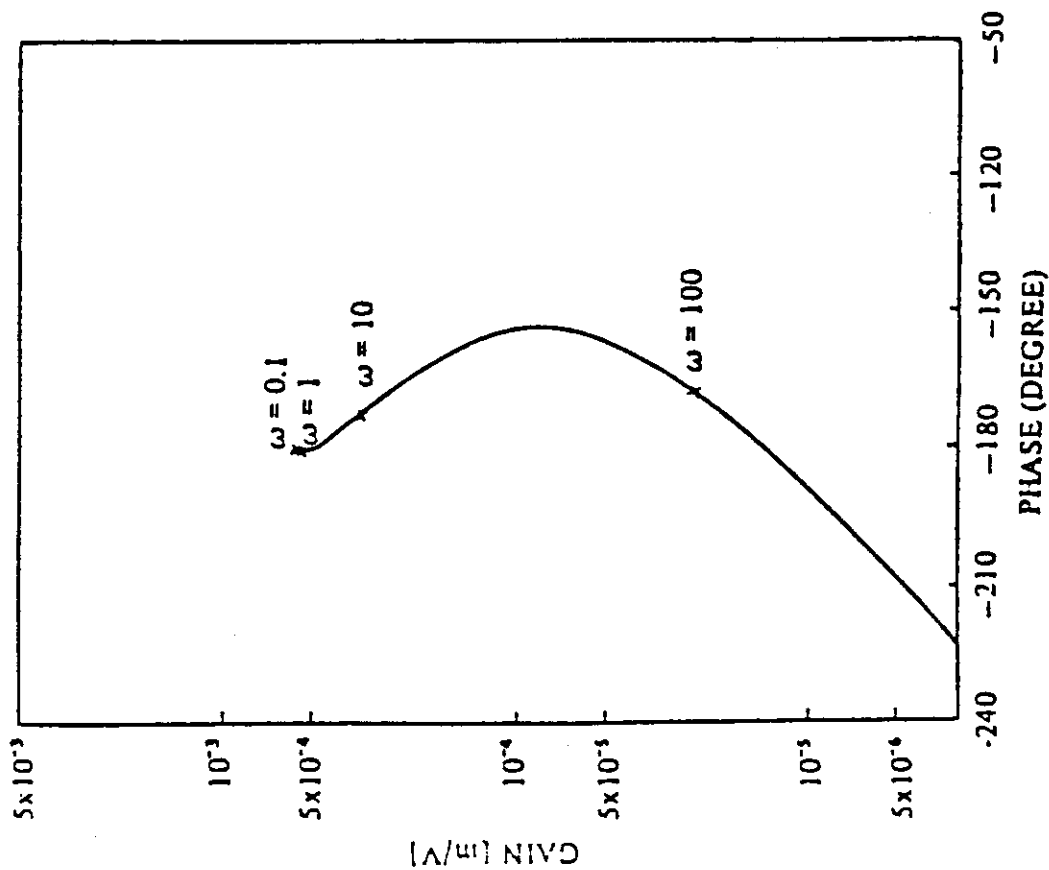


Fig. 2.2.23(a) Gain-phase diagram of open loop transfer function for P feedback control system with the reference shell structure and control coil-B.



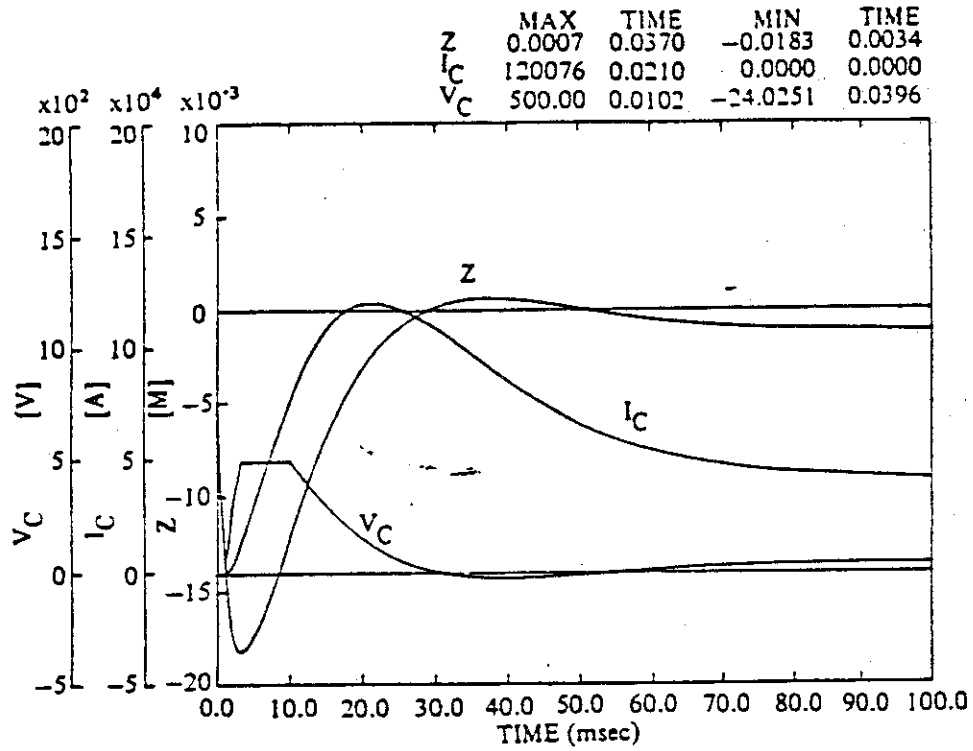


Fig. 2.2.24(a) Simulation result of plasma position control by P action with reference shell and control coil-B.  $G = 35000$ ,  $V_L = \pm 500V$ , Power Supply Capacity : 60 MVA.

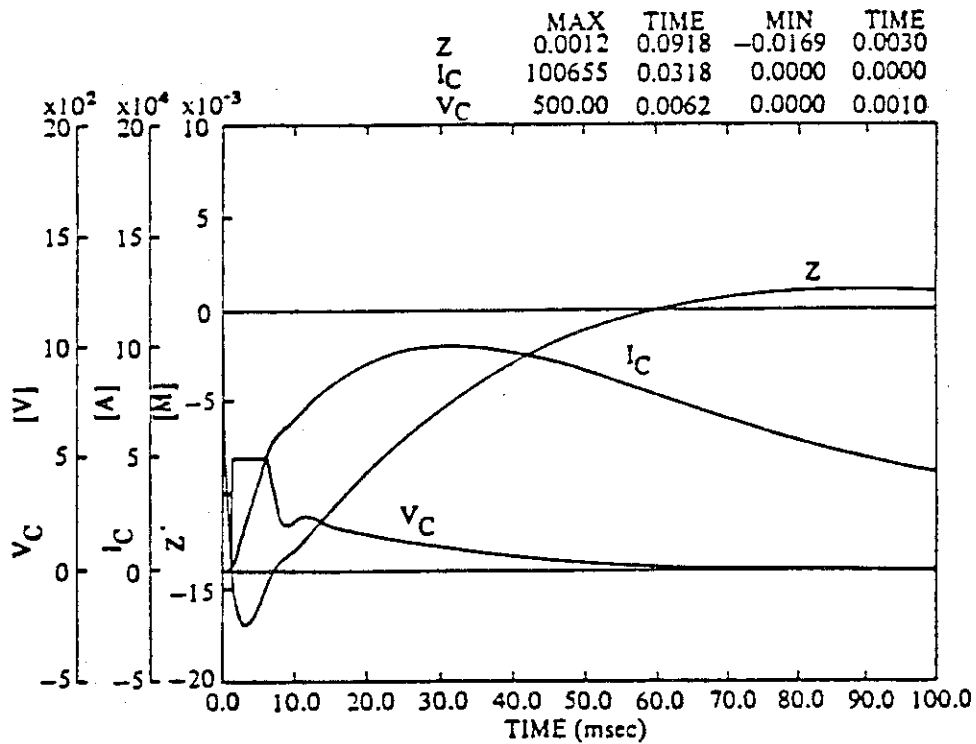


Fig. 2.2.24(b) Simulation result of plasma position control by PID action with reference shell and control coil-B.  $G = 25000$ ,  $T_I = 200$  msec,  $T_D = 10$  msec,  $V_L = \pm 500$  V, Power Supply capacity : 50 MVA.

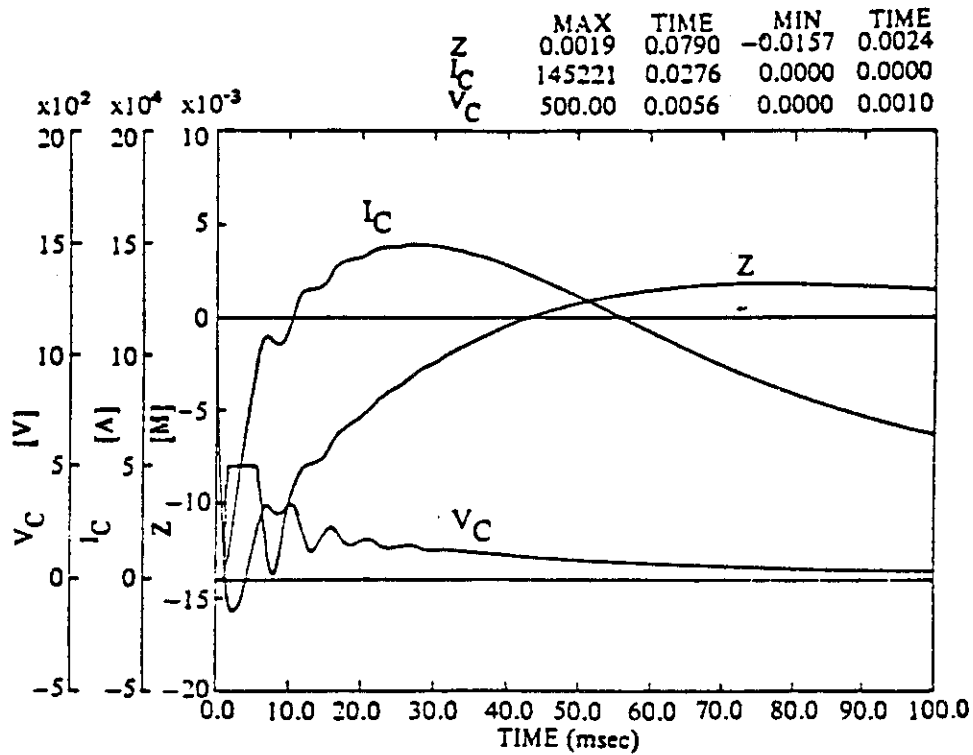


Fig. 2.2.24(c) Simulation result of plasma position control by PID action with reference shell and control coil-A.  $G = 30000$ ,  $T_I = 50$  msec,  $T_D = 10$  msec,  $V_L = \pm 500$  V, Power Supply Capacity : 75 MVA

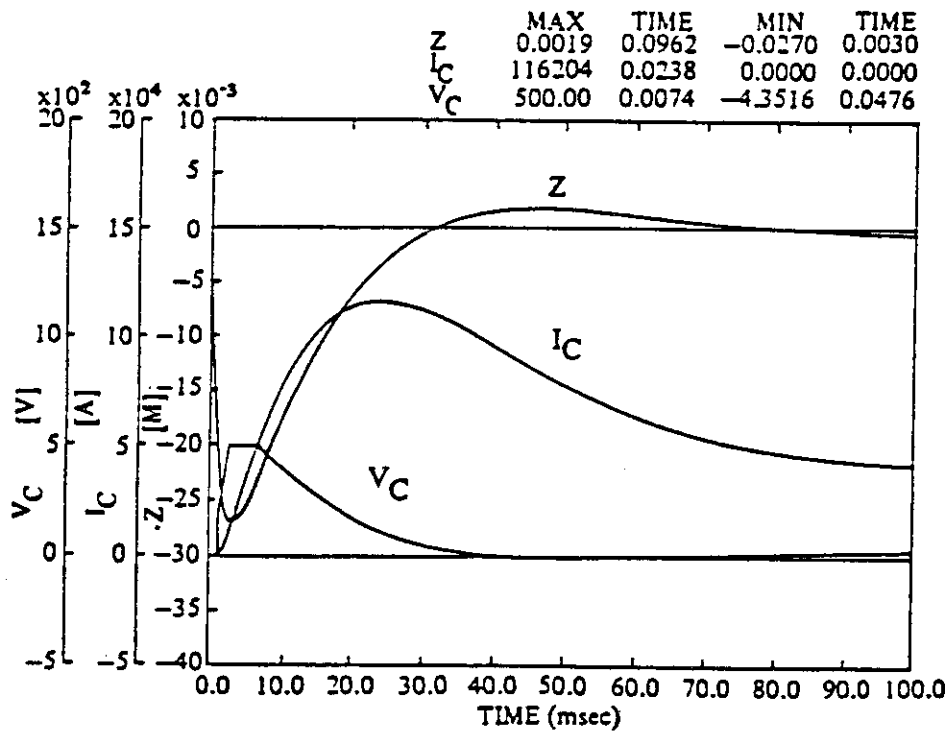


Fig. 2.2.24(d) Simulation result of plasma position control by PID action with alternative shell and control coil-B.  $G = 20000$ ,  $T_I = 200$  msec,  $T_D = 10$  msec,  $V_L = \pm 500$  V, Power Supply Capacity : 58 MVA.

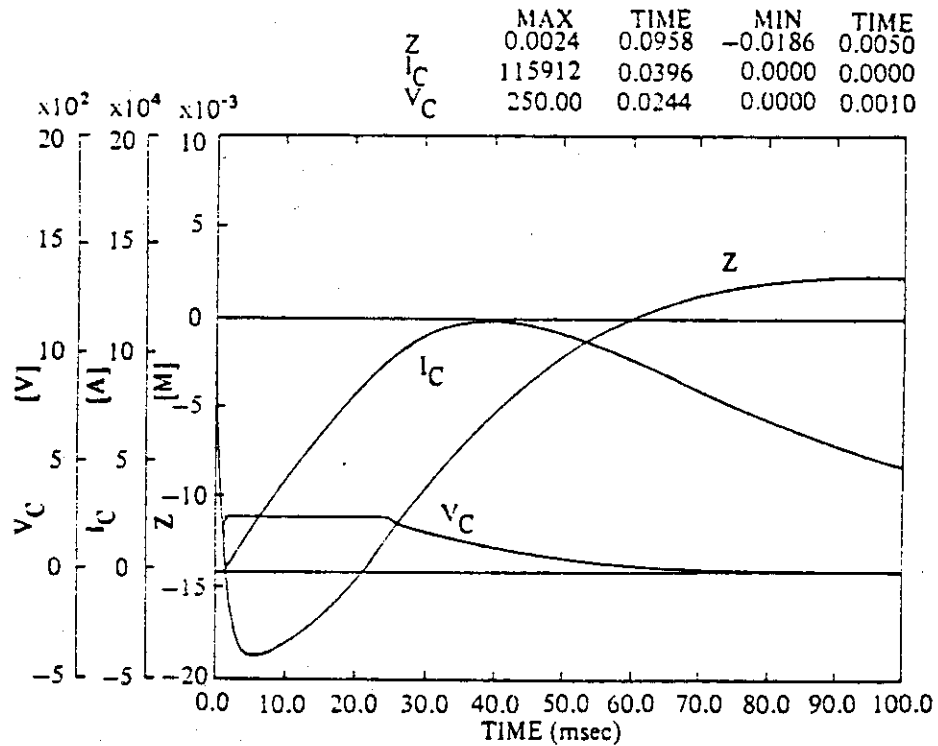


Fig. 2.2.25 Simulation result of plasma position control by PID action with reference shell and control coil-B.  $G = 25000$ ,  $T_I = 200\text{msec}$ ,  $T_D = 10\text{msec}$ ,  $V_L = \pm 250\text{V}$ , Power Supply Capacity : 29 MVA.

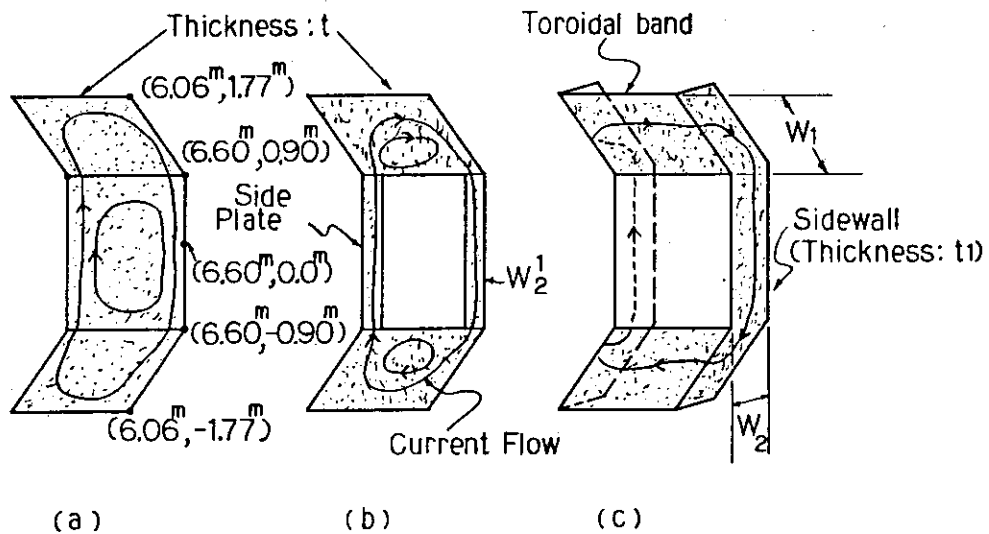


Fig. 2.2.26 Three kinds of shell structures

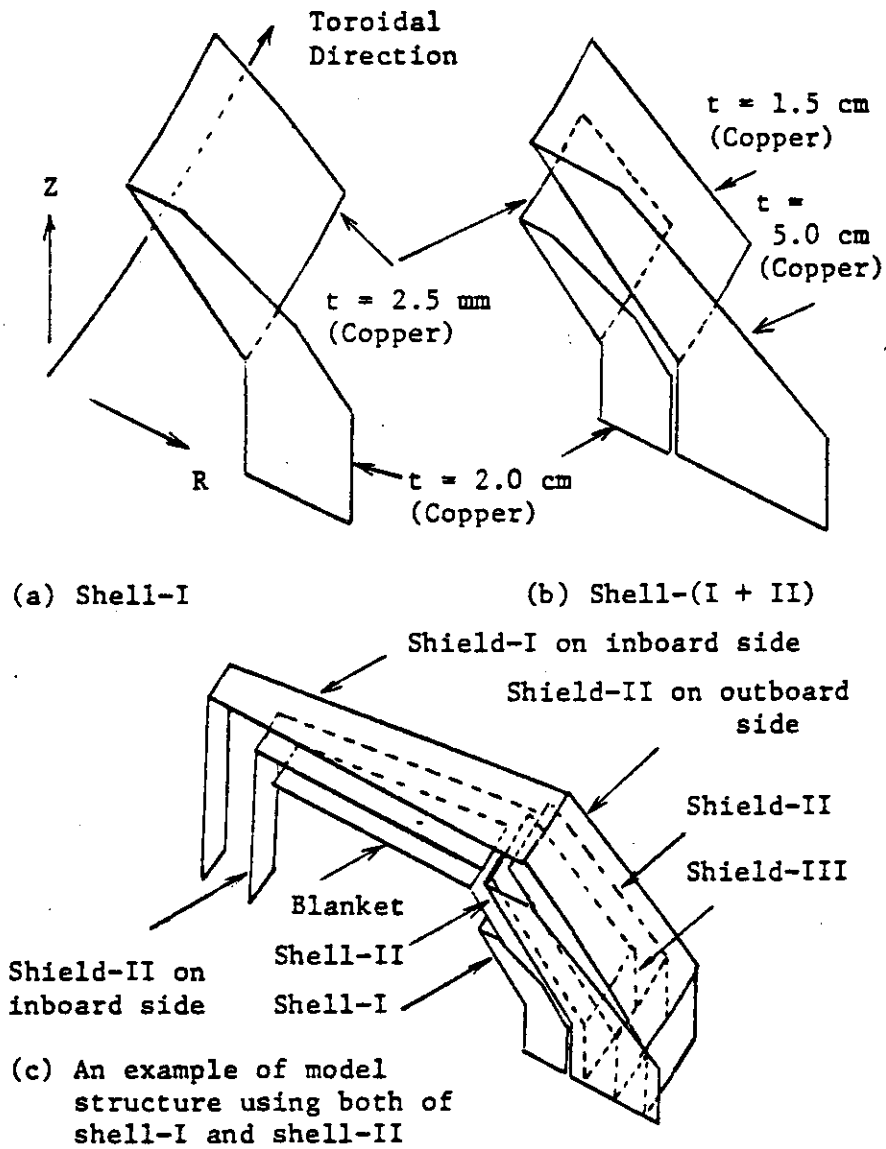


Fig. 2.2.27 A simplified model of the INTOR

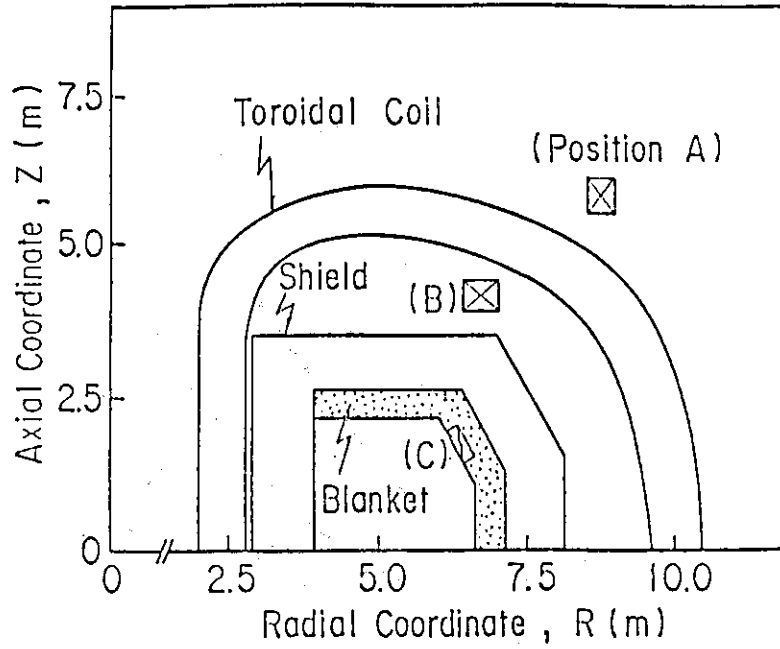


Fig. 2.2.28 Three candidates for the coil location of vertical position control

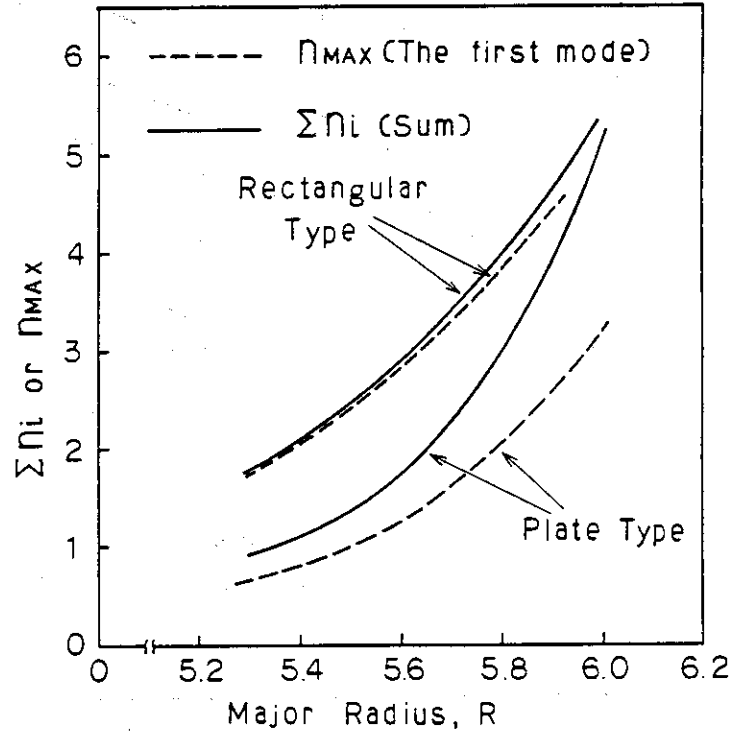


Fig. 2.2.29  $\Sigma n_i$  and  $n_{MAX}$  as a function of the radius position (Major Radius:  $R$ ) of the filament of the filamentary model on both cases of (a) and (c) in Fig. 2.2.26

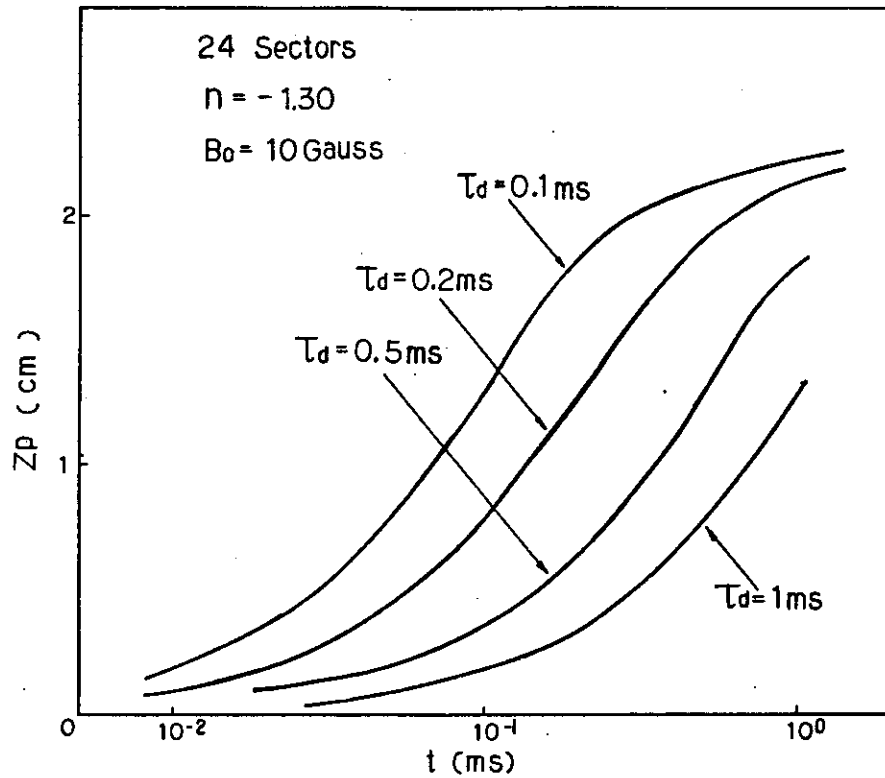


Fig. 2.2.30  $Z_p$  as a function of  $\tau_d$  with the disturbance field

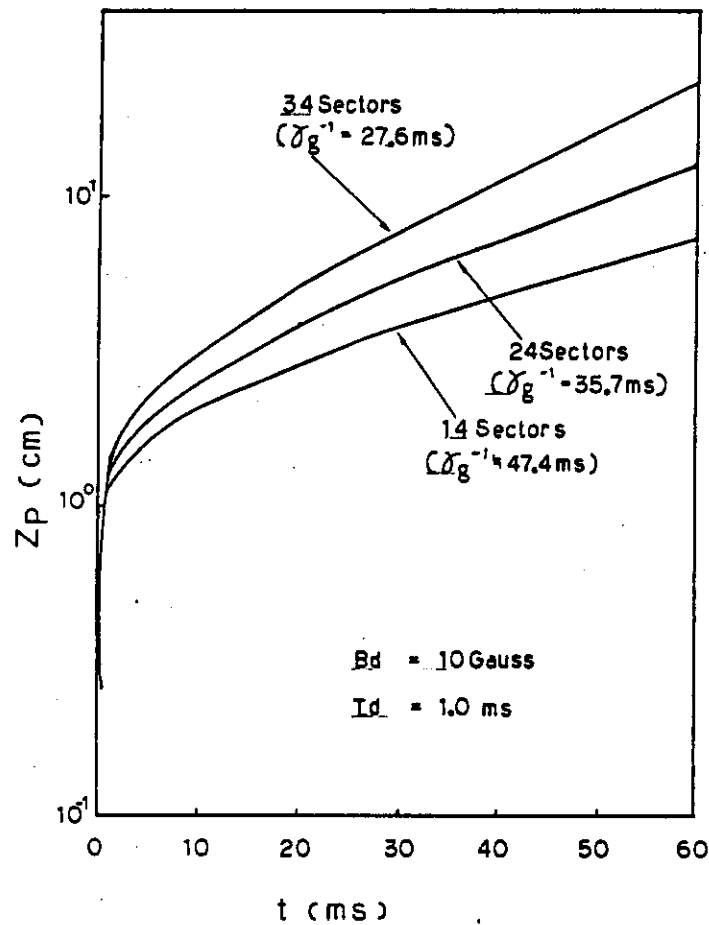


Fig. 2.2.31 Shell effect as a function of the number of toroidally divided sectors

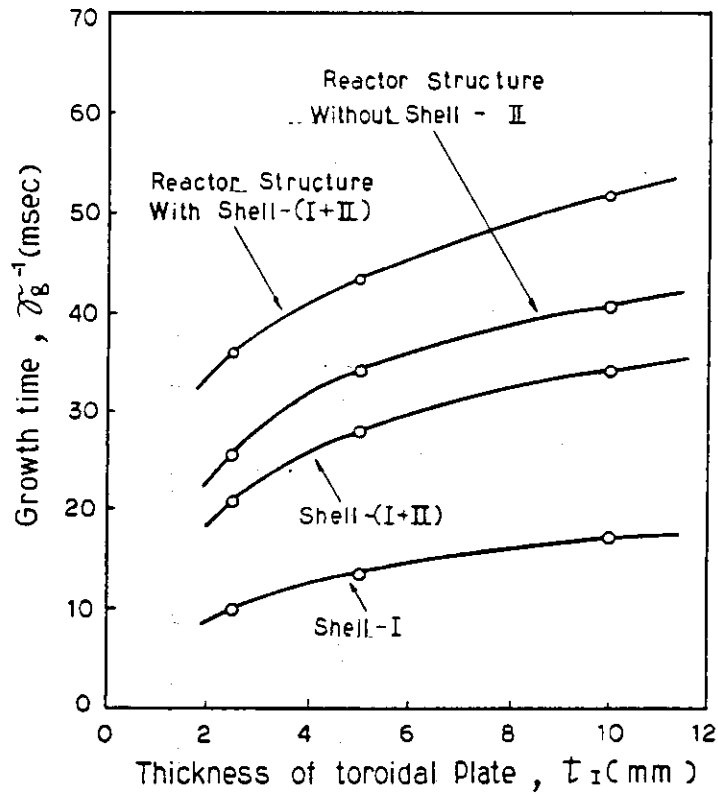


Fig. 2.2.32 Instability growth time as a function of the thickness of toroidal plate with the shell-I

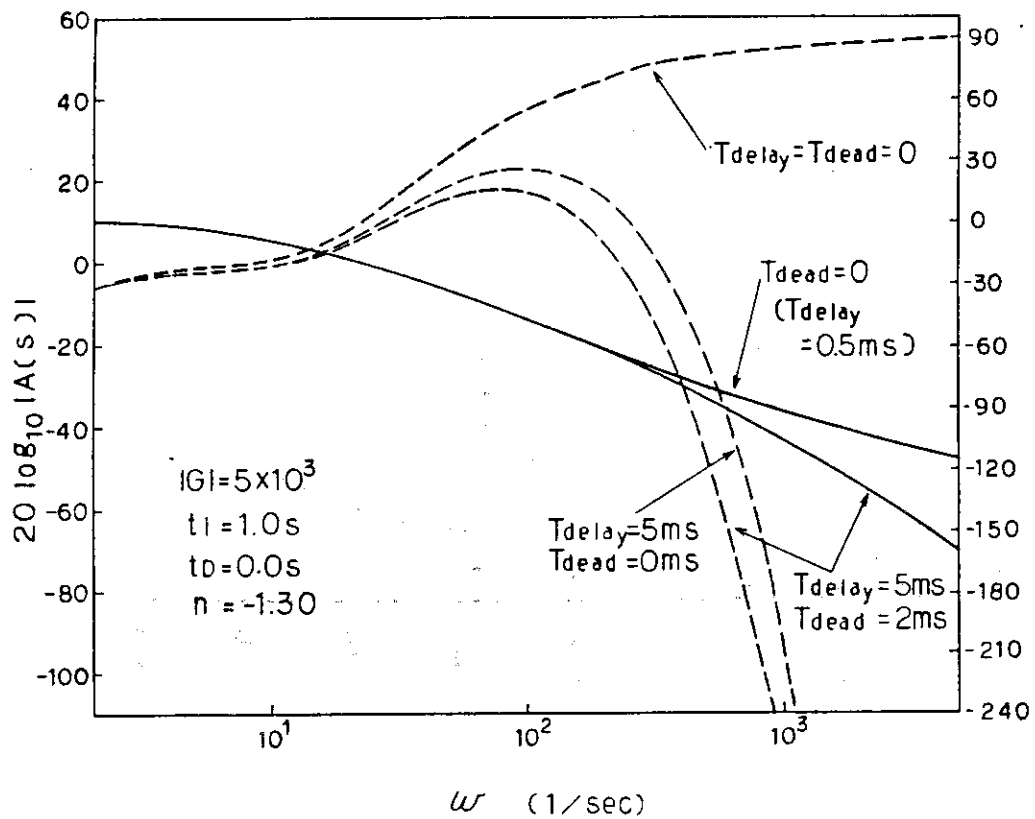


Fig. 2.2.33 Bode diagram with the PI controller

# INTOR PARAMETRIC STUDY

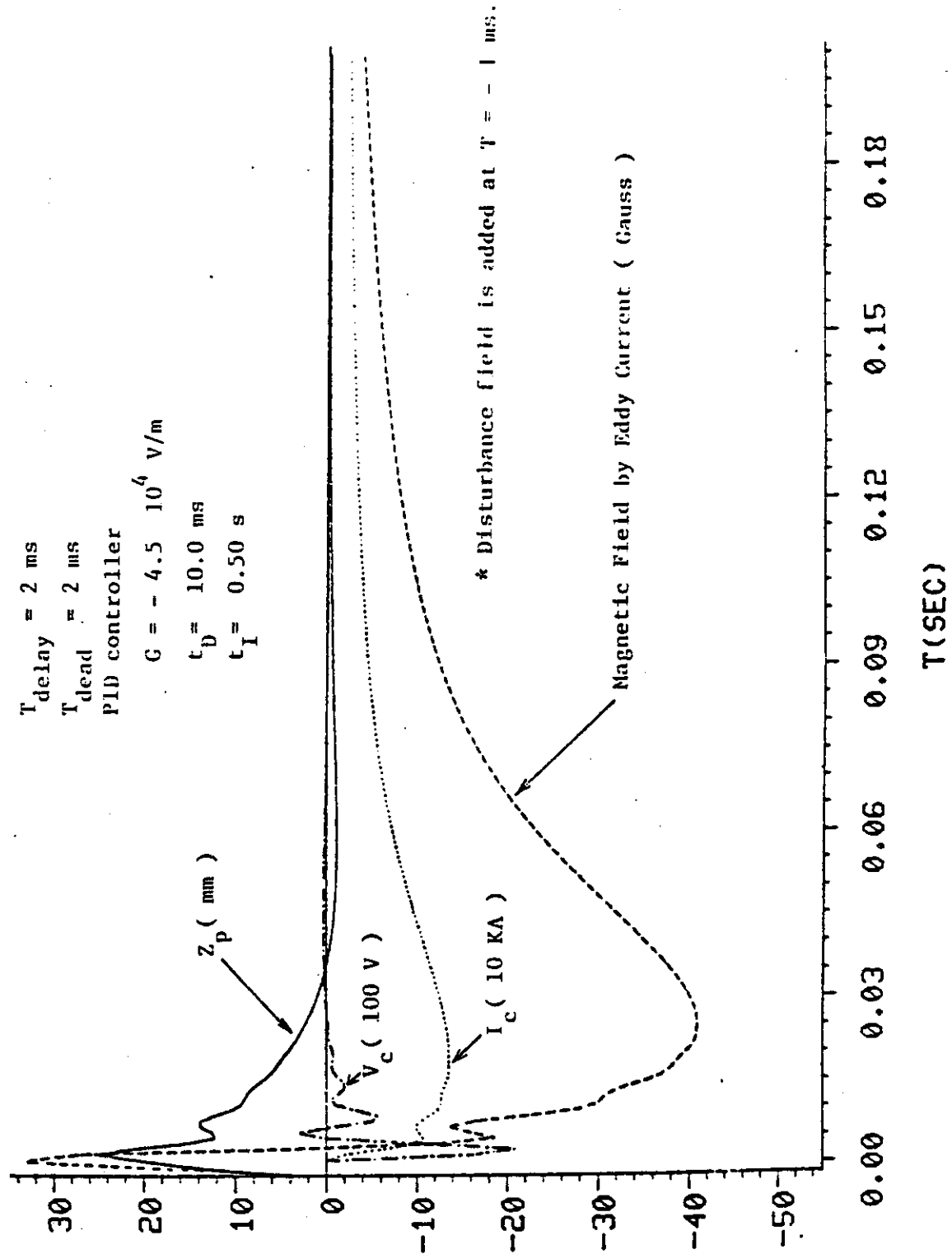


Fig. 2.2.34 A typical example of the vertical position control



# INTOR PARAMETRIC STUDY

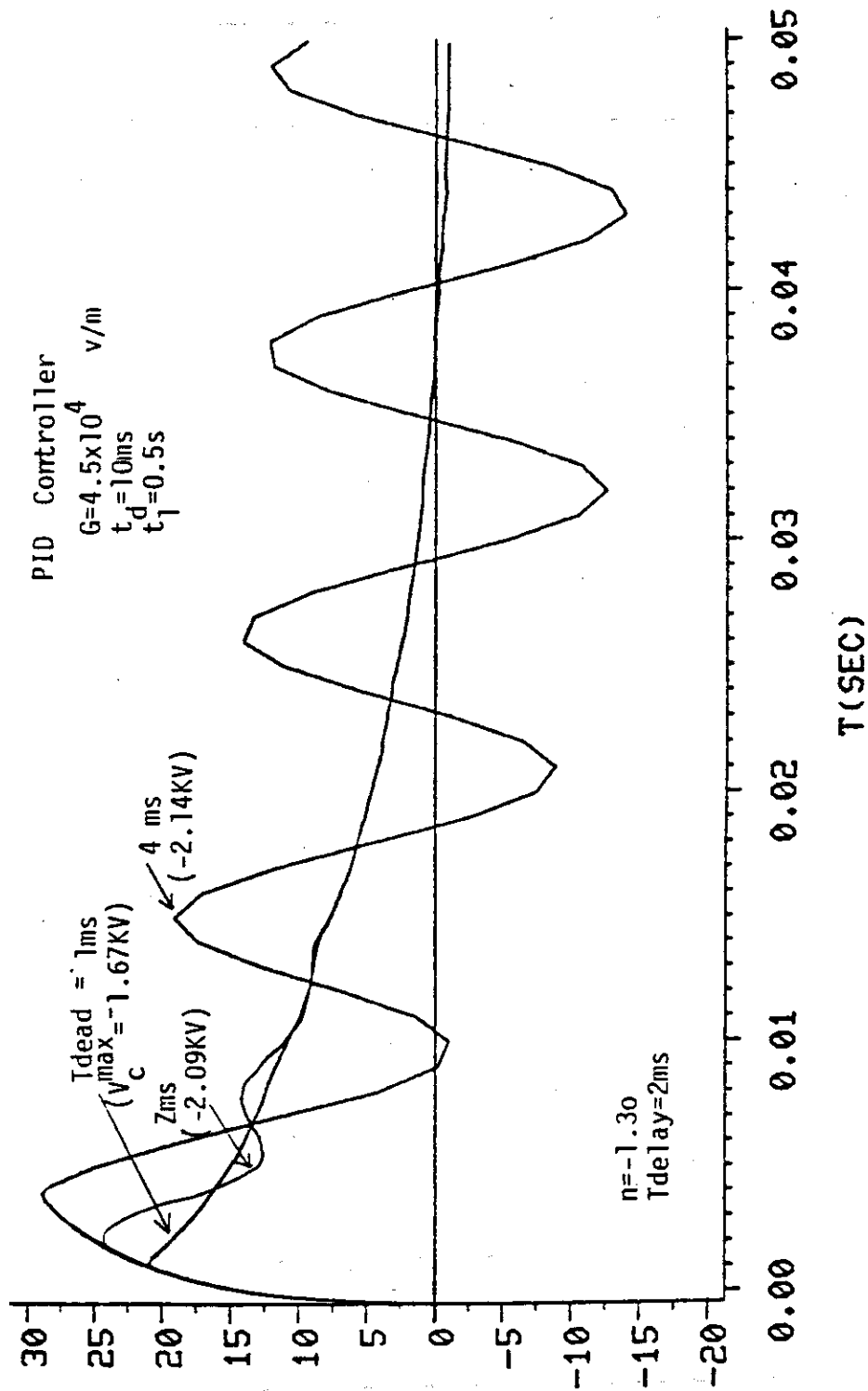


Fig. 2.2.35 Dependence of the position control characteristics upon the dead time of power supply

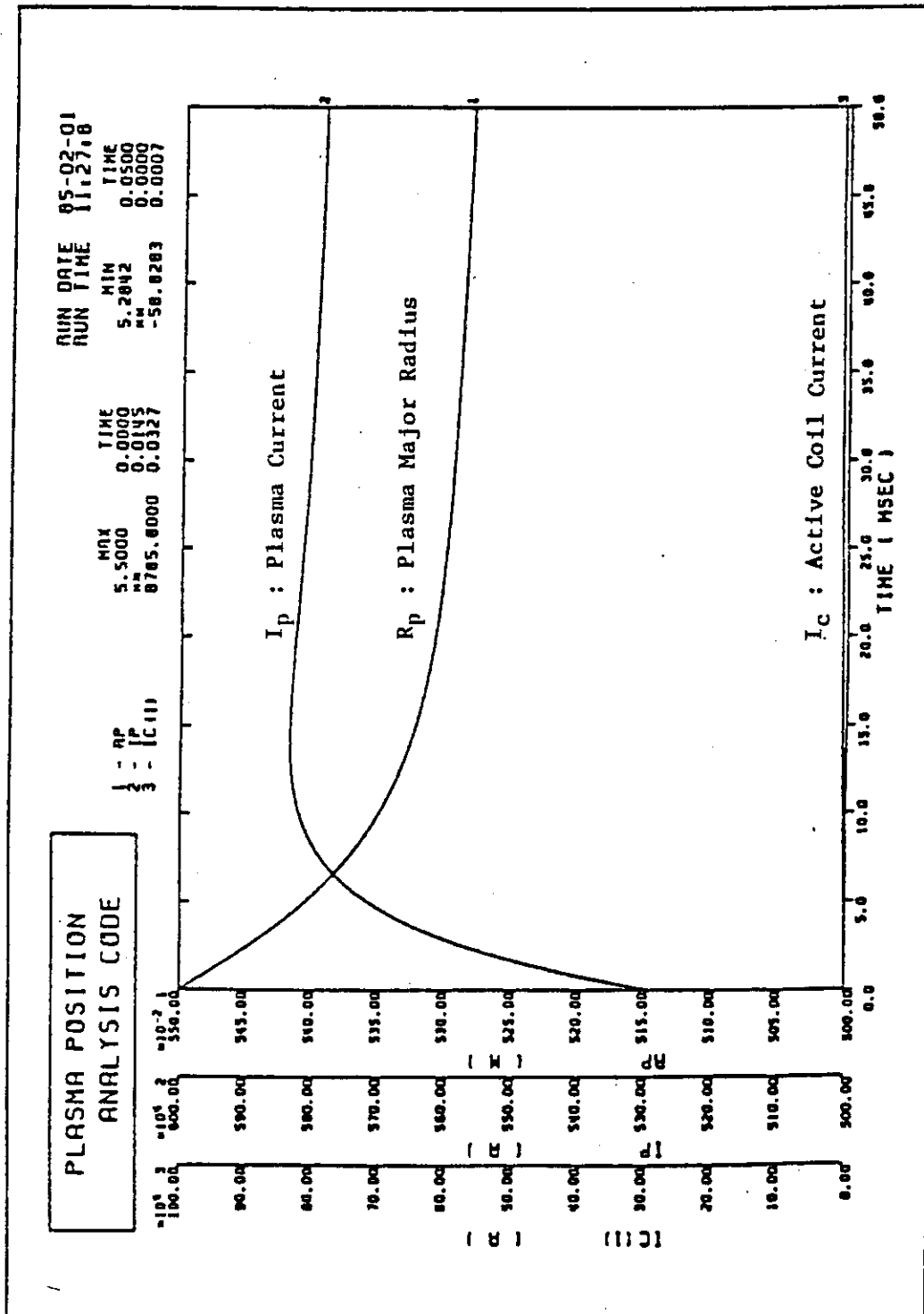


Fig. 2.3.1 Simulation result of plasma radial movement without feedback control. High conductive shells are installed in blanket modules for plasma vertical position control.  $\beta_p$  and  $\beta_i$  are reduced by 40 % with a time constant of 5 msec.

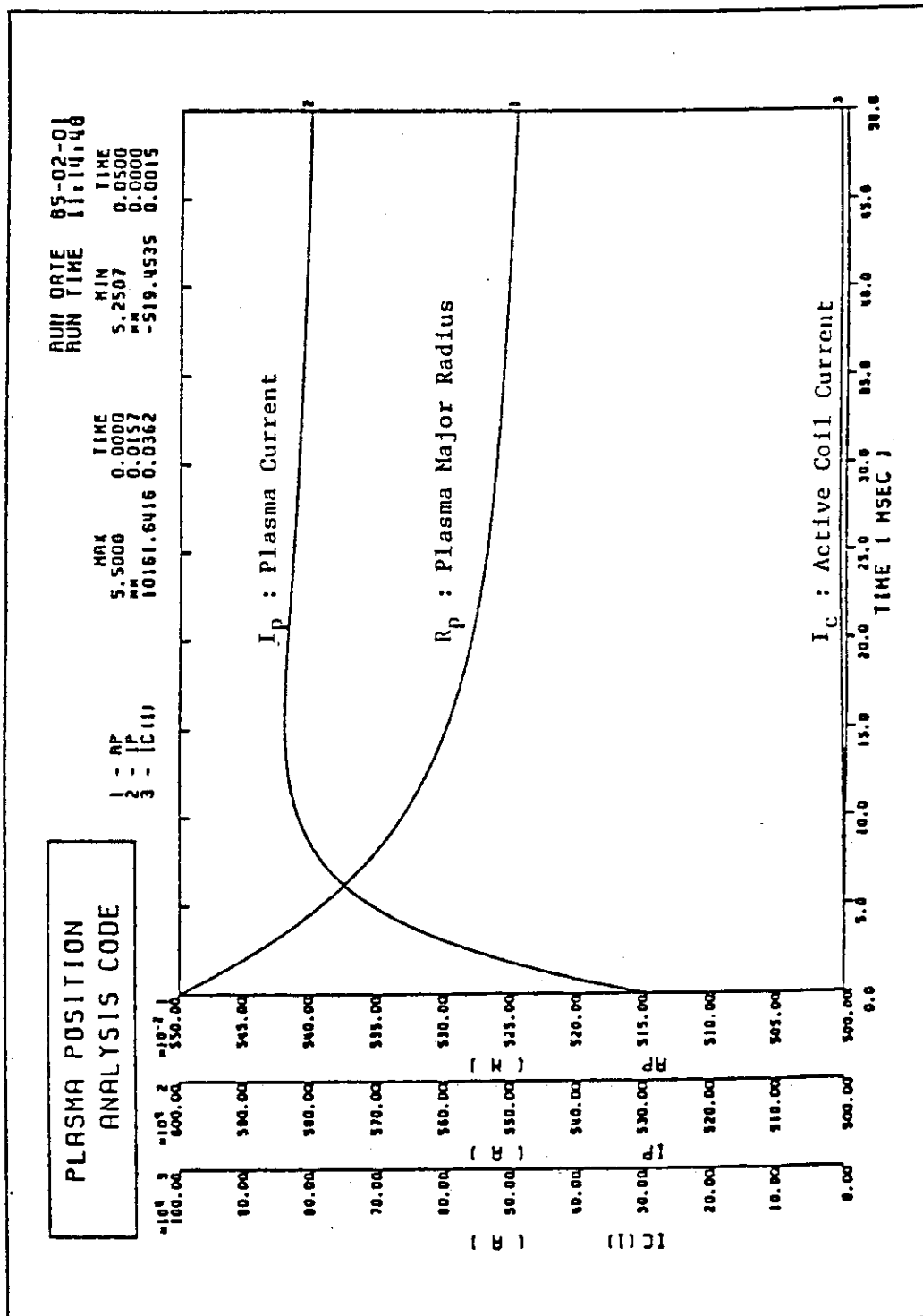


Fig. 2.3.2 Simulation result of plasma radial movement without feedback control. Conductive shells are not installed in blanket modules. Perturbation conditions are the same as those in figure 2.3.1 .

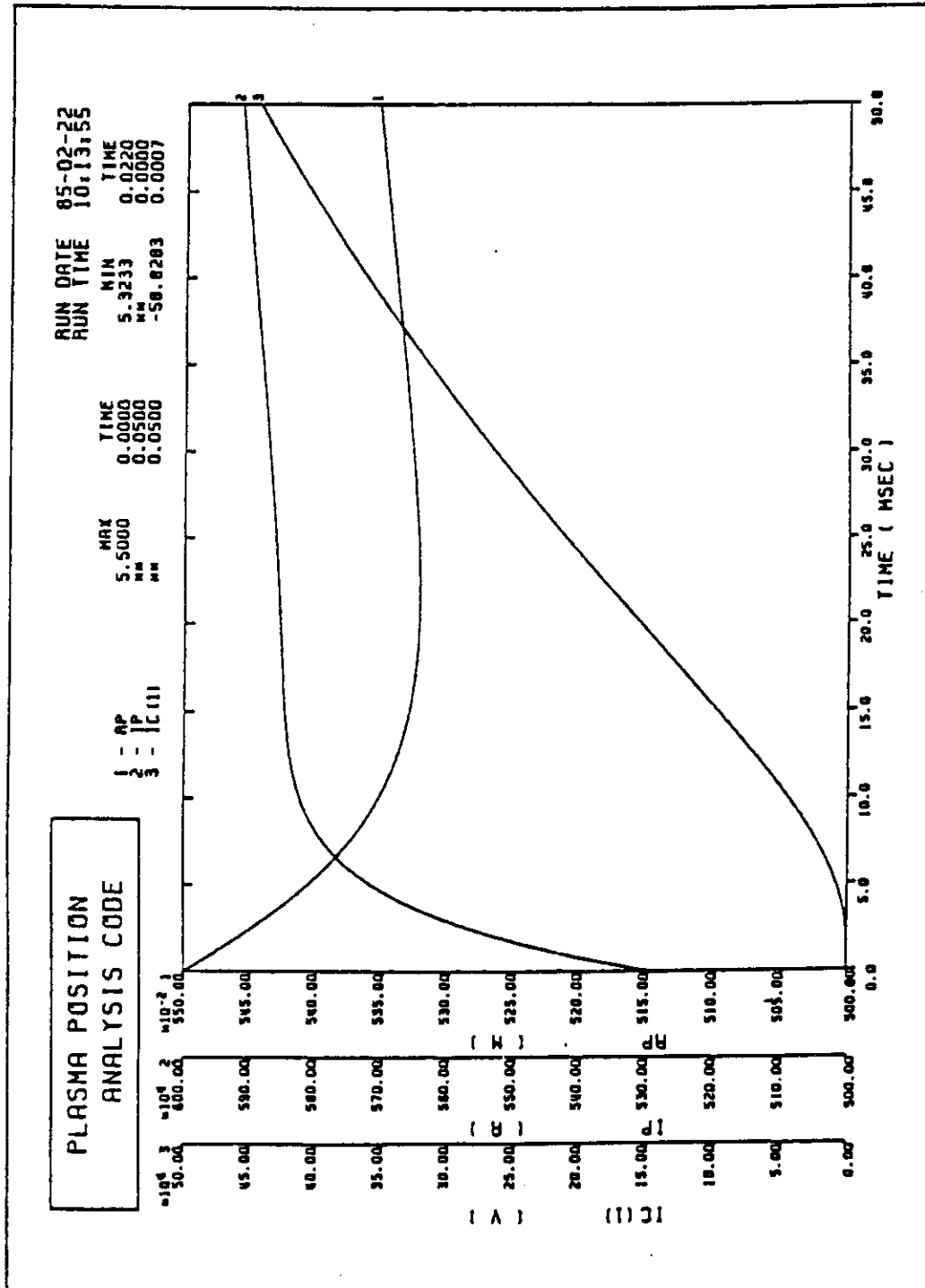


Fig. 2.3.3 Simulation result of plasma radial position control by P action. High conductive shells are installed in blanket modules for plasma vertical position control. Perturbation conditions are the same as those in figure 2.3.1.  
 Max. Coil Voltage: 4400 V, Max. Coil Current: 450 kA, Max. Power: 2000 MVA.

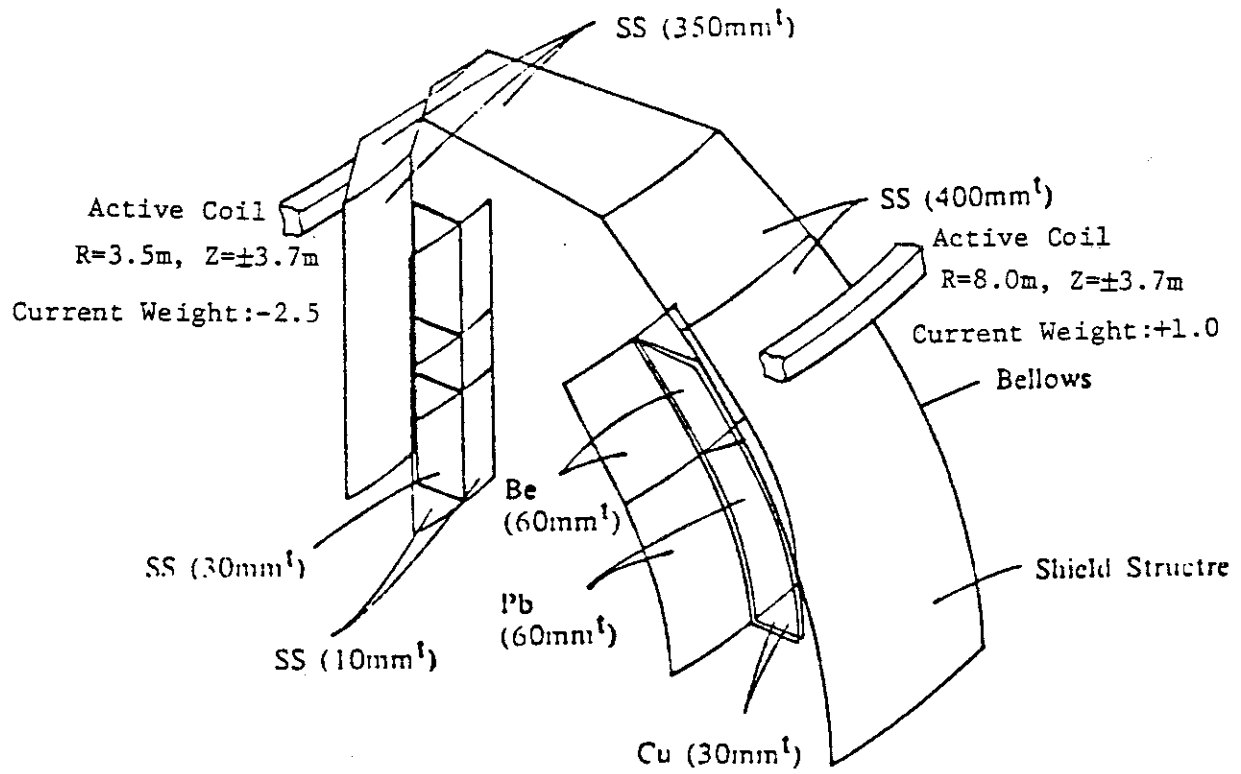


Fig. 2.3.4 Model of feedback control system for plasma radial position.

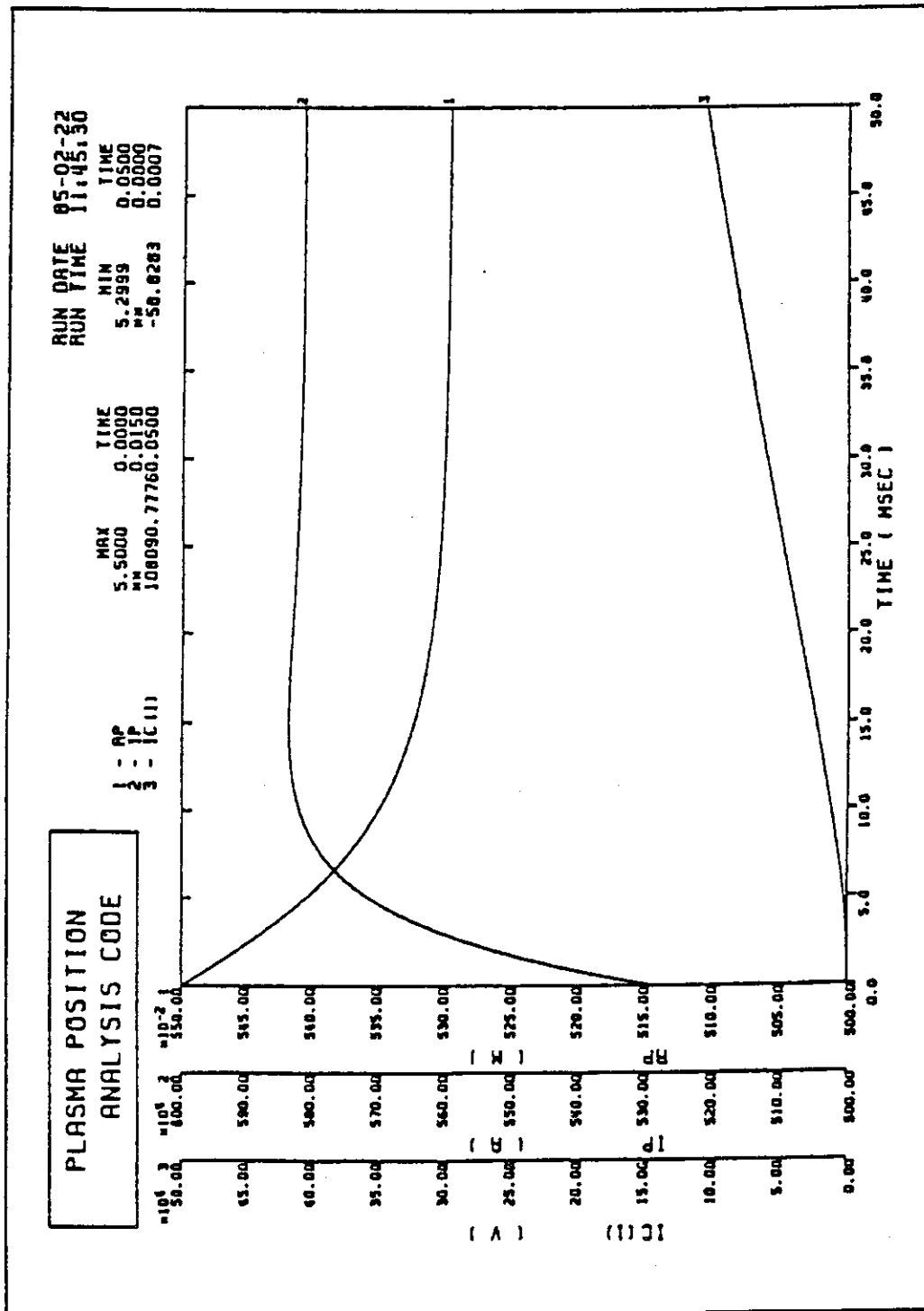


Fig. 2.3.5 Simulation result of plasma radial position control by P action. High conductive shells are installed in blanket modules for plasma vertical position control. Perturbation conditions are the same as those in figure 2.3.1. Max. Coil Voltage: 1000 V, Max. Coil Current: 110 kA, Max. Power: 110 MVA.

### 3. Start-up Effects

#### 3.1 Magnetic and Electrical Field Penetration

In future tokamak reactors, plasmas are surrounded by multiple layers such as first wall, blanket, shield, coil vacuum chamber and toroidal field magnet including shear panels. This multilayer structure will prevent fast penetration of magnetic and electrical field at start-up phase.

The magnetic field and loop voltage applied to plasma are reduced by eddy currents and represented by eqs. (3.1.1) and (3.1.2), respectively, using Laplace transform as described in section 2.3.

$$B_v(s) = \frac{M_{pc}^* I_c}{2\pi R_p} \{1 - M(s)\} \quad (3.1.1)$$

$$V_p(s) = -M_{pOH} s I_{OH} \{1 - V(s)\} \quad (3.1.2)$$

$$M(s) = \sum_k \frac{s\tau_k}{1+s\tau_k} m_k, \quad m_k = \frac{M_{pk}^* M_{ck}}{M_{pc}^* \tau_k}$$

$$V(s) = \sum_k \frac{s\tau_k}{1+s\tau_k} v_k, \quad v_k = \frac{M_{pk} M_{OHk}}{M_{pOH} \tau_k}$$

where,  $M_{pk}$  : mutual inductance between plasma and k-th eddy current mode

$M_{ck}$  : mutual inductance between magnetic field coil and k-th eddy current mode

$M_{pOH}$  : mutual inductance between plasma and OH coil

$M_{OHk}$  : mutual inductance between OH coil and k-th eddy current mode

$\tau_k$  : time constant of k th eddy current mode

$M^*$  : derivative of mutual inductance

$I_c$  : Laplace transform of magnetic field coil current

$I_{OH}$  : Laplace transform of ohmic coil current decay rate

When plasma is perfectly surrounded by conductive structures and coils are located outside the structures,  $M(s \rightarrow \infty)$  and  $V(s \rightarrow \infty)$  are equal to one. So the fields are completely shielded at  $t = 0$ .

### Magnetic field penetration

Figures 3.1.1 and 3.1.2 show  $M(s)$  of JAERI FER system. The comparison between the cases with and without high conductive shells is represented in Fig. 3.1.1, and the effect of one turn resistance is shown in Fig. 3.1.2. The configuration of passive elements is shown in Fig. 3.1.3. The presence of high conductive shells does not largely affect the shielding effect,  $M(s)$ , since the high conductive shells does not perfectly enclose the plasma. On the other hand, the  $M(s)$  depends strongly on one turn resistance of radiation shield/coil vacuum chamber. This result suggests that the radial magnetic field for plasma vertical position control will be shielded by radiation shield/coil vacuum chamber, and the control property will be deteriorated, if the active coils are located outside the toroidal field coils, and toroidal one turn resistance will not be required because of plasma current ramp up by LHRF current drive. On the other hand, low one turn resistance of radiation shield/coil vacuum chamber will improve the property of stabilizing plasma vertical position instability, whether the control property of feedback system will be improved or not depends on which property (shielding property or stabilizing property) is effective to the overall control property. Figure 3.1.4 shows the gain-phase diagrams of the feedback control system of the JAERI FER. The curve (s) is the diagram for the standard case that one turn toroidal resistance is  $0.2 \text{ m}\Omega$ . In curves (b) and (c), the shield/coil chamber has no bellows and toroidal high electric conductance. The effective thickness of the shield/coil chamber is  $35 \sim 40 \text{ cm}$  in curves (a) and (b). In order to analyze the case that the active coils are located between shield/coil chamber and toroidal field coils, the effective thickness is set at  $10 \text{ cm}$ . A large gain value is required for feedback control when the shield/coil chamber has high electric conductance in the toroidal direction, though low one turn resistance expands the stable range of feedback controller where the phase is larger than  $-180$  degrees when the gain in diagram is equal to one. This result suggests that the required power supply capacity of the active coils will be too large unless the active coils are located inside the shield/coil chamber. The toroidal one turn resistance of the order of  $0.2 \text{ m}\Omega$  will be needed to avoid the difficulties to make the concept of active coils inside the shield compatible with other engineering problems such as remote maintenance, coil insulations, installations, space arrangements, etc.

### Electrical field penetration

Assuming that plasma current is absent and ohmic heating coil current is linearly decreased, the loop voltage can be derived from eq. (3.1.2) as

$$V_p = -M_{pOH} \dot{I}_{OH} \left\{ 1 - \sum_k \frac{M_{pk} M_{OHk}}{M_{pOH} \tau_k} e^{-\frac{t}{\tau_k}} \right\} \quad (3.1.3)$$

$$V_p^{\text{Norm}} = 1 - \sum_k \frac{M_{pk} M_{OHk}}{M_{pOH} \tau_k} e^{-\frac{t}{\tau_k}}$$



Figure 3.1.5 shows the normalized loop voltage,  $V_p^{\text{Norm}}$ , as a parameter of one turn bellow resistance. The penetration of loop voltage is largely delayed, when one turn bellow resistance is lower than approximately 0.1-0.2 m $\Omega$ .

Summarizing the magnetic and electrical field penetration, one turn resistance of approximately 0.2 m $\Omega$  is reasonable, when active coils are located outside the shield or plasma current is inductively ramped up by ohmic heating coils.

### 3.2 Components Specification

Components specification on one turn resistance in the toroidal direction,  $R_{\text{loop}}$  depends on the basic concept and design choice such as the scenario of plasma current ramp up and the selection of active coil locations. One turn toroidal resistance of approximately 0.2 m $\Omega$  will be required unless plasma current is ramped up by non inductive method such as LHFR current drive for all operation phases including early stage operations, or if active control coils are located outside radiation shield/coil vacuum chamber. Summarizing components specification, two choices can be considered from the view point of field penetration as shown in Table 3.2.1.

Table 3.2.1 Components Specification

	Inductive $I_p$ Rise	Noninductive $I_p$ Rise
One-turn Resistance	$\sim 0.2\text{m}\Omega$	no specification
Active Coil Location	no specification	inside radiation shield (if $R_{\text{loop}} \ll 0.1\text{m}\Omega$ )

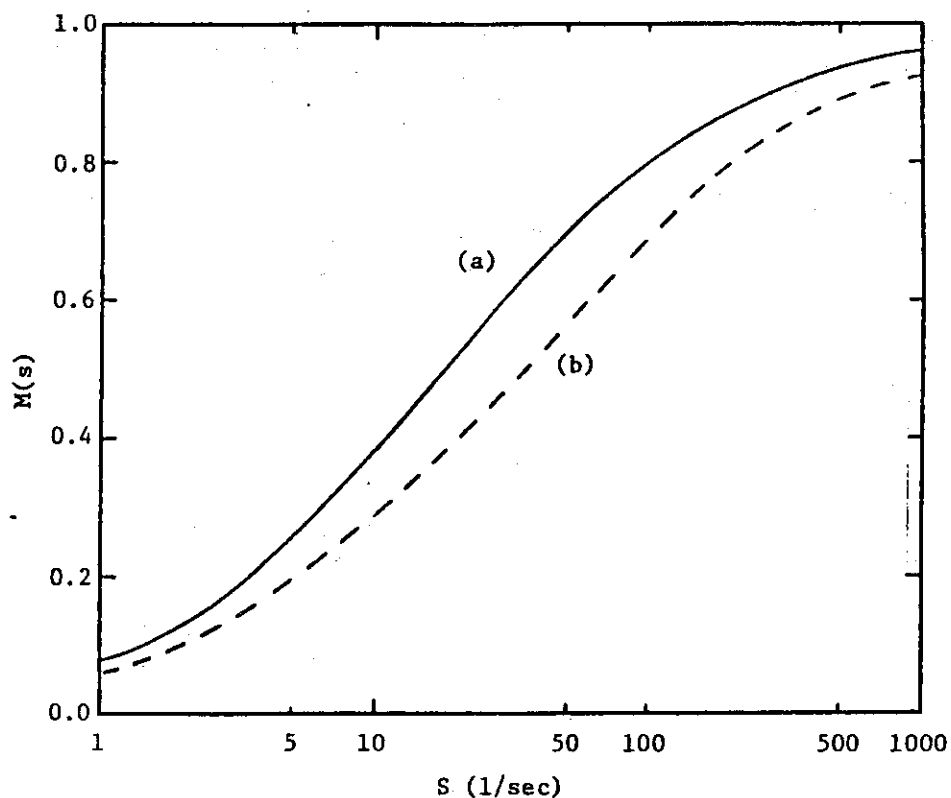


Fig. 3.1.1 Effects of magnetic field shielding as a function of conductive shell thickness.  
(a) Standard shells, (b) Half of standard shell thickness.

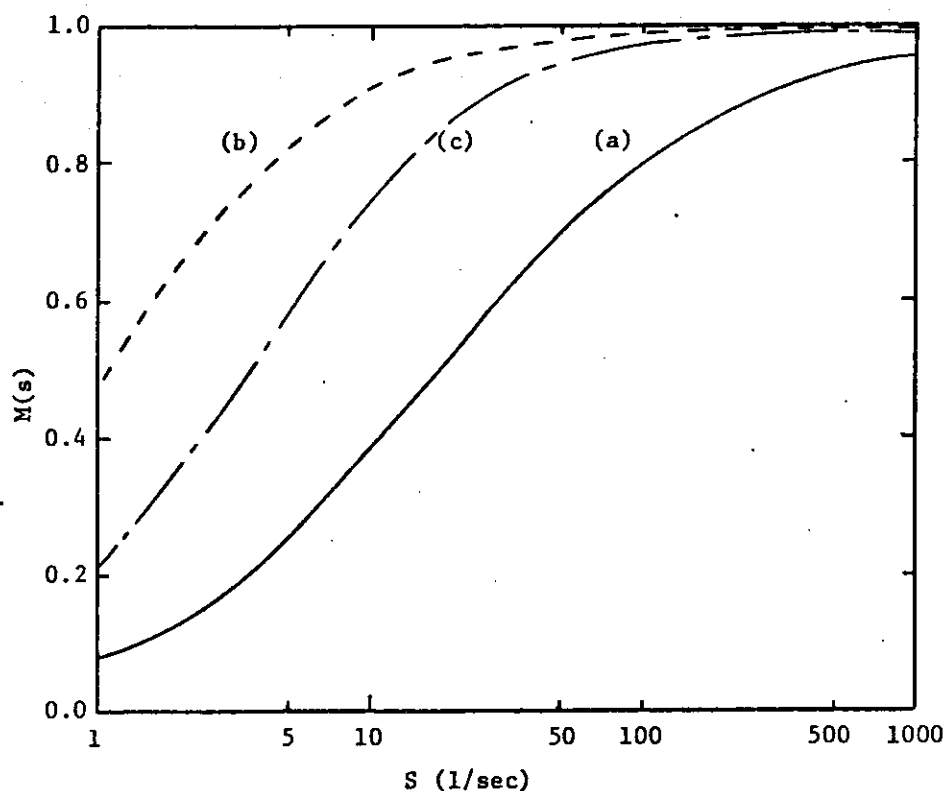


Fig. 3.1.2 Effects of magnetic field shielding as a function of toroidal one turn resistance and effective electrical thickness of shield.  
(a) 0.2  $m\Omega$  of bellows resistance and 35-40 cm of effective shield thickness.  
(b) 0  $m\Omega$  of bellows resistance and 35-40 cm of effective shield thickness.  
(c) 0  $m\Omega$  of bellows resistance and 10 cm of effective shield thickness.

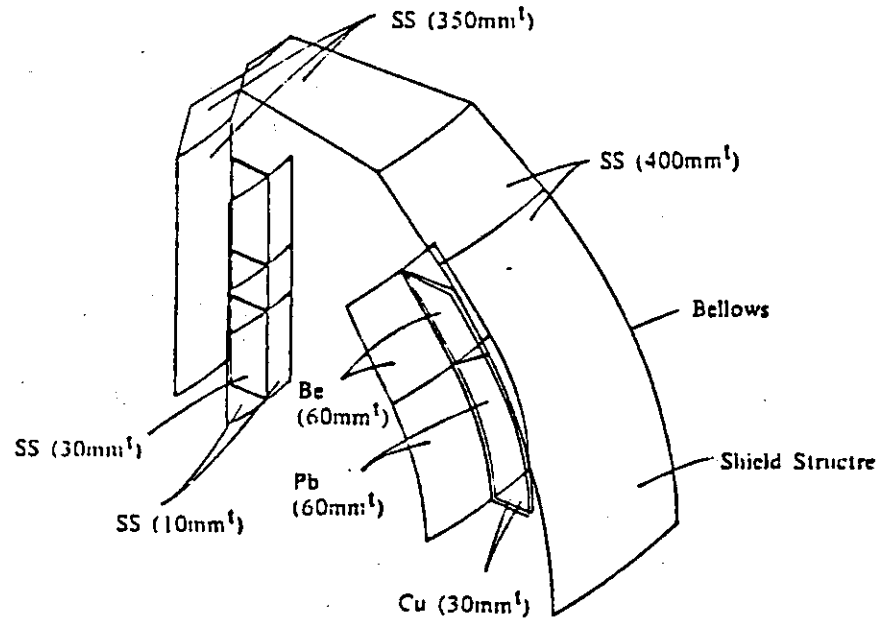


Fig. 3.1.3 Analysis model of passive elements

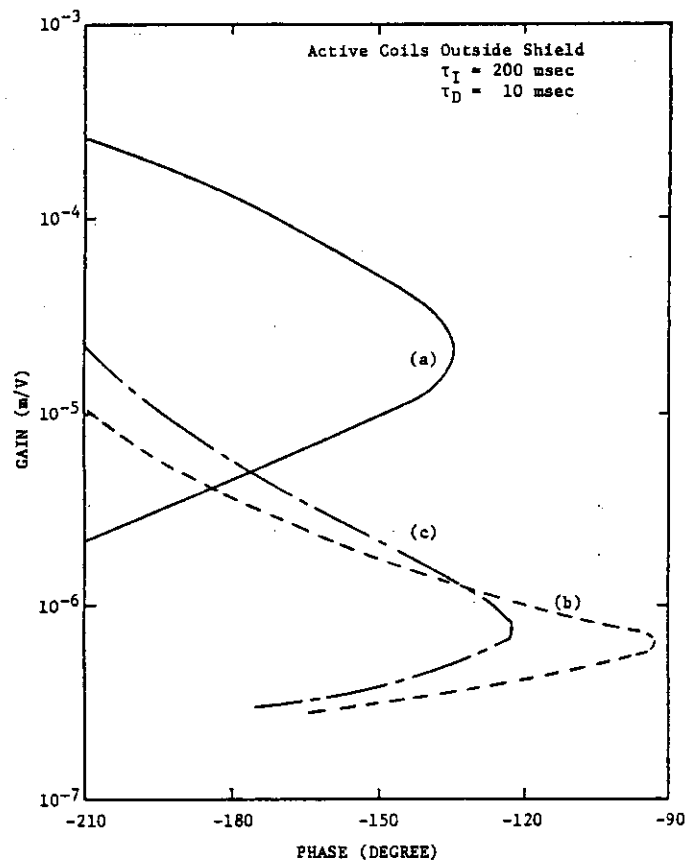


Fig. 3.1.4 Gain-phase diagrams of open-loop transfer functions for PID control systems.

- (a) 0.2 m $\Omega$  of bellows resistance, 35-40 cm of effective shield thickness.
- (b) 0 m $\Omega$  of bellows resistance, 35-40 cm of effective shield thickness.
- (c) 0 m $\Omega$  of bellows resistance, 10 cm of effective shield thickness.

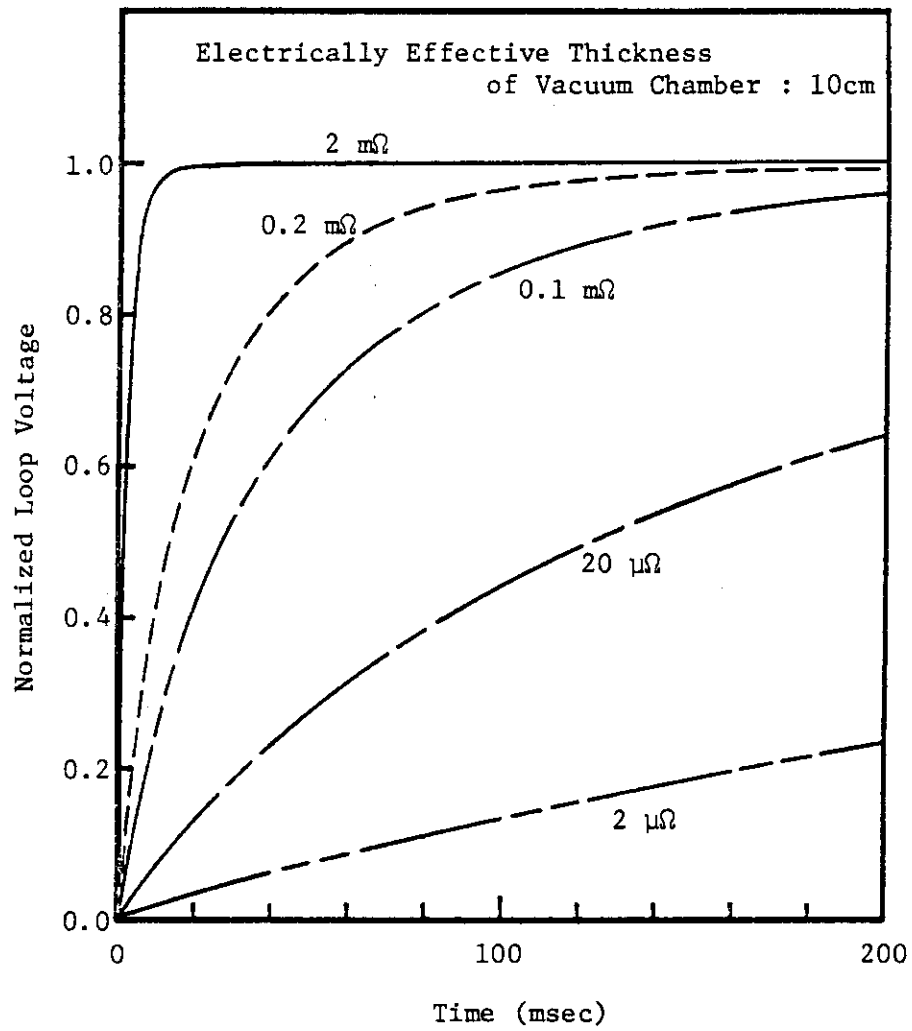


Fig. 3.1.5 Time evolution of normalized plasma loop voltage as a function of bellows resistance.

## 4. Plasma Disruption Effects

### 4.1 Plasma Models

The disruption effects such as electromagnetic forces depend on plasma models. In this section, we restrict our discussions to the models which deal with plasmas as a electric circuit.

Two models can be considered from the view point of plasma toroidal current distribution as

- i) Single filament model which represents plasma as a filament ring coil.
- ii) distributed current model in accordance with MHD plasma equilibrium.

Model i) provides higher maximum electromagnetic force in FER blankets than model ii) by approximately 10 ~ 30%, if plasma position does not change at disruption.

The following models on the decaying pattern of plasma current can be cited as typical examples.

- a. Without plasma movement.
  - a-1. Current decay with a constant rate,  $I_p/\tau_d$ .
  - a-2. Exponential decay with a time constant of  $\tau_d$ .
- b. Current decay with plasma movement.

Here,  $I_p$  is plasma current before disruption, and  $\tau_d$  is disruption time.

The electromagnetic forces of FER blankets calculated with model a-1 are larger than those with model a-2 by approximately 50% as shown in section 4.3. In order to analyze disruption effects with model b, the disruption scenario such as moving speed and direction, disruption time should be determined consistently with shell properties of passive elements, clearance around plasma, etc. The disruption analyses by model b are important future works, since our preliminary results suggest that model b brings significant changes in forces and its distribution.

### 4.2 Induced Voltage

During plasma disruptions, a high voltage would be induced in the components surrounding plasmas. This high voltage may produce arcing between the adjacent modules of blanket and divertor plate, etc., unless the components are properly insulated. For the design specification, it does not necessarily need to evaluate accurate voltages, but enough to estimate the order of values. So we roughly estimate the voltage value induced between the adjacent blanket modules.

The current decay rate is approximately  $4 \times 10^8$  A/S, since plasma current is 6 MA, and current decay time is  $\sim 15$  msec. On the approximation that plasma and blanket have concentric circular crosssectional shapes, the mutual inductance between plasma and blanket,  $M_{pB}$ , is estimated as

$$M_{pB} \approx \mu_0 R_p \left( \ln \frac{8R_p}{a_B} - 2 \right) = 10 \mu H,$$

where,  $R_p$  is a plasma major radius (5.3 m), and  $a_B$  is a minor radius of the first wall (1.25 m).

Therefore, the one turn voltage induced at blanket,  $V_B$ , is evaluated as

$$V = M_{pB} \left| \dot{I}_p \right| \approx 4000 \text{ V}$$

If the number of blanket modules is twenty four, the induced voltage between adjacent modules is approximately 170 V.

On the other hand, assuming that the radiation shield/coil vacuum chamber can be considered as perfect conductor, the induced voltage will be reduced. In this case, the one turn induced voltage can be approximately expressed as

$$V_B = \mu_0 R_p \left| \dot{I}_p \right| \ln \frac{a_s}{a_B} \approx 1850 \text{ V}$$

where,  $a_s$  is a minor radius of the shield/coil vacuum chamber, and  $a_s/a_B = 2$  is assumed. In the case of 24 blanket modules, the induced voltage between adjacent modules is estimated as  $\sim 75$  Volts.

From above estimations, we specify the one turn induced voltage at disruptions as 3600 Volts: 24 modules  $\times$  75 Volts  $\times$  factor 2 of safety margin.

#### 4.3 Forces

The analyses of eddy currents and associated electromagnetic forces at plasma disruptions were performed by the EDDYTRAN program [17]. The eddy currents and electromagnetic forces were calculated for the blanket of the Fusion Experimental reactor (FER). We considered two kinds of type models, the reference and the alternative shell designs [18,19], for two kinds of the plasma current decay scenarios. The plasma of FER has double null points for divertor operation and its major and minor radii are 5.5 m and 1.1 m with elongation 1.5, respectively. The plasma current, toroidal field on the axis and thermal fusion power are 5.3 MA, 5.7 T and 440 MW, respectively.

For the reference shell design, the eddy currents on the outboard front wall detour around the ducts. So the electromagnetic forces on the outboard front wall near the ducts are larger than those for the alternative shell design. The electromagnetic forces decaying linearly between 0 sec and 15 msec are one and half times as large as those decaying exponentially with a time constant of 15 msec.

### Analytical Model

In the FER design, the first wall is integrated into the blanket for better tritium breeding ratio and easy remote maintenance. Considering plasma vertical stabilization, the blanket assembly is divided into 28 modules along the torus and highly conductive shells are installed on the outboard blanket modules, as described in section 2.2.2. The blanket assembly is composed 14 center modules and 14 side modules. Each blanket module has a stainless steel can-structure filled with  $\text{Li}_2\text{O}$  pellets and water cooling tubes. The highly conductive shell is installed on the stainless steel wall.

Fig. 4.3.1 shows the analytical model. This analysis is based on partial model between 0 and  $\pi/14$  in toroidal angle, due to symmetry of electrical conditions and blanket configuration. As the boundary conditions, current potential is assumed to be plane-symmetry in the toroidal direction across the 0 and  $\pi/14$  planes and antisymmetry in the Z-direction across the plasma mid-plane. EDDYTRAN program [11] prepares a whole torus model by rotating the above partial model. Eddy current and electromagnetic force analyses have been performed for two models, the reference shell design and the alternative shell design [16]. Fig. 4.3.2 shows electrical properties for each model. The plasma current of 5.3 MA is assumed to be multi-filamentary according to plasma equilibrium analysis (Table 4.3.1) and decay as described in the following two cases. As case 1, the plasma current decays exponentially with a time constant of 15 msec. As case 2, the plasma current decays linearly between 0 sec and 15 msec. The other components such as shield, cryostat vacuum chamber and others are neglected.

### Eddy Currents and Electromagnetic Forces

The eddy current distribution for the reference shell design with the conditions of case 1 at 20 msec after a plasma disruption is shown in Fig. 4.3.3. The principal eddy currents are induced on the outboard front wall near the plasma mid-plane in the plasma current direction. These currents flow through the side wall and return on the outboard end wall and the beryllium part of the outboard front wall. The eddy current on the lead part of the outboard front wall detour around the duct part. The other remarkable eddy currents occur on the inboard end wall. The eddy current distribution for the alternative shell design is shown in Fig. 4.3.4. This eddy currents are with the condition of case 1 at 20 msec after a plasma disruption. The eddy currents on the outboard front wall near the plasma mid-plane are smaller in comparison with the reference shell design. However the eddy currents on the outboard end wall near the plasma mid-plane are rather larger. The other eddy current modes are similar to those for the reference shell design.

Fig. 4.3.5 shows the joule loss distribution due to the eddy currents for the reference shell design with the condition of case 1 at 20 msec after a plasma disruption. The major joule heat appears on the lead part of the outboard front wall near the plasma mid-plane due to the high current density and the high resistivity of lead. Fig. 4.3.6 shows the joule loss distribution for the alternative shell design with the same condition as for the reference shell design. In this case, the joule heat on the lead part is reduced.

The electromagnetic force distribution, due to interaction with the toroidal field, shown in Fig. 4.3.7 is for the reference shell design with the condition of case 1 at 20 msec after a plasma disruption. The toroidal field is assumed to be 5.7 T at  $R=5.5$  m and dependent on  $1/R$  ( $R$  is the coordinate in the major radius direction). R-direction components of these electromagnetic forces are reversed across the plasma mid-plane. The R-direction and Z-direction components are also reversed across 0 and  $\pi/14$  planes in the toroidal direction. Large electromagnetic forces are observed on the outboard side wall, the outboard front wall near the duct and the inboard end wall. The force on the outboard side wall, due to saddle-like currents, is about  $30 \text{ kgf/cm}^2$ , in-plane force. The force on the outboard front wall, due to detouring currents around the duct, is about  $14 \text{ kgf/cm}^2$ , out-of-plane force. That on the inboard end wall is about  $29 \text{ kgf/cm}^2$ , out-of-plane force. The similar electromagnetic force distribution for the alternative shell design is shown in Fig. 4.3.8. On the outboard front wall, the electromagnetic force is smaller in comparison with the reference shell design while it is larger on the outboard end wall. The other electromagnetic forces are similar to those of the reference shell design. If importance should be given to the fact that the stiffness of the first wall is smaller than that of the end wall, the alternative shell design could be preferred than the reference shell design.

The time dependencies of electromagnetic forces due to the eddy currents and the toroidal field are shown in Fig. 4.3.9 and Fig. 4.3.10 for the condition of case 1 and case 2, respectively. In case 1, the maximum electromagnetic forces on the inboard front wall and the inboard side wall appear at 5 msec. On the outboard front wall and the inboard end wall, they appear at 20 msec. At the outboard end wall, it appears at 30 msec. The electromagnetic forces on the outboard end wall decay with larger time constants compared with those on the other walls. In case 2, most of the electromagnetic forces reach the maximum earlier and decay with shorter time constants than for case 1. The electromagnetic forces on most of the walls reach the maximum at 15 msec and one and a half times as large as in case 1.

#### Concluding Remarks

We performed eddy current and electromagnetic analyses for two candidate shell designs [18] with two kinds of the plasma current decay scenarios. The major conclusions are as follows.

1. The eddy currents and joule heat on the outboard front wall near the plasma mid-plane for the reference shell design are larger than those for the alternative shell design.



2. The electromagnetic forces on the outboard front wall near the duct for the reference shell design are larger than those for the alternative shell design.
3. The maximum electromagnetic force for the reference design is about 30 kgf/cm<sup>2</sup> which appears on the outboard side wall (in-plane force) and the inboard end wall (out-of-plane force).
4. The electromagnetic forces decaying linearly between 0 sec and 15 msec are one and half times as large as those decaying exponentially with a time constant of 15 msec. The former increase and decay with shorter time constants than the latter.

Table 4.3.1 Plasma Current Distribution

$i_p$ (A)	$9.55709 \times 10^2$	$2.27255 \times 10^4$	$7.46850 \times 10^5$		
R (m)	4.70000	5.01703	5.35943		
Z (m)	1.83367	1.87327	1.81705		
$i_p$ (A)	$9.62207 \times 10^3$	$9.20979 \times 10^4$	$1.40868 \times 10^5$	$3.78005 \times 10^4$	
R (m)	4.67341	5.03898	5.49459	5.86714	
Z (m)	1.44778	1.47487	1.45326	1.35840	
$i_p$ (A)	$2.21881 \times 10^4$	$1.40597 \times 10^5$	$2.59948 \times 10^5$	$2.51907 \times 10^5$	$2.20546 \times 10^4$
R (m)	4.64461	5.03843	5.51377	5.97538	6.31767
Z (m)	0.97662	0.98883	0.98668	0.96685	0.85532
$i_p$ (A)	$3.12003 \times 10^4$	$1.70315 \times 10^5$	$3.23987 \times 10^5$	$3.88272 \times 10^5$	$1.36859 \times 10^5$
R (m)	4.63274	5.03774	5.51653	5.99685	6.36971
Z (m)	0.49251	0.49530	0.49480	0.49164	0.46559
$i_p$ (A)	$3.49432 \times 10^4$	$1.80360 \times 10^5$	$3.45022 \times 10^5$	$4.28295 \times 10^5$	$1.93643 \times 10^5$
R (m)	4.62508	5.03751	5.51713	5.99960	6.38947
Z (m)	0.0	0.0	0.0	0.0	0.0

WALL AND PLATE NAME	PART NUMBER	REFERENCE MATERIAL (THICKNESS[mm])	ADJUTIVE MATERIAL (INTERIE S.S. [mm])
FRONT WALL OF OUTBOARD CENTER NOUURE	1, 4, 5, 8, 9	Pb(60.0)	S.S.(10.0)
FRONT WALL OF OUTBOARD SIDE NOUURE	2, 3, 6, 7, 10, 11	Be(60.0)	Be(60.0)
FRONT WALL OF INBOARD CENTER NOUURE	12, 13, 16, 17	Pb(60.0)	S.S.(10.0)
FRONT WALL OF INBOARD SIDE NOUURE	14, 15, 18, 19	Be(60.0)	Be(60.0)
END WALL OF OUTBOARD CENTER NOUURE	20 ~ 25	S.S.(10.0)	S.S.(10.0)
END WALL OF OUTBOARD SIDE NOUURE	26 ~ 28	S.S.(10.0)	S.S.(10.0)
END WALL OF INBOARD CENTER NOUURE	30 ~ 40	S.S.(100.0)	Cu(30.0)
END WALL OF INBOARD SIDE NOUURE	41 ~ 48	S.S.(100.0)	Cu(30.0)
END WALL OF INBOARD CENTER NOUURE	{ 50, 51, 53, 54 52, 55	S.S.(130.0) S.S.(180.0)	S.S.(130.0) S.S.(160.0)
END WALL OF INBOARD SIDE NOUURE	{ 56, 57 58	S.S.(130.0) S.S.(180.0)	S.S.(130.0) S.S.(160.0)
SIDE WALL OF OUTBOARD CENTER NOUURE	65 ~ 68	Cu(30.0)	Cu(16.0)
SIDE WALL OF OUTBOARD SIDE NOUURE	69 ~ 72	Cu(30.0)	Cu(30.0)
SIDE WALL OF INBOARD CENTER NOUURE	80 ~ 82	S.S.(10.0)	S.S.(10.0)
SIDE WALL OF INBOARD SIDE NOUURE	83 ~ 85	S.S.(10.0)	S.S.(10.0)
SIDE WALL AT DUCT	61	S.S.(10.0)	S.S.(10.0)
UPPER PLATE AT DUCT	60	S.S.(10.0)	S.S.(10.0)
END PLATE OF OUTBOARD CENTER NOUURE	90 ~ 92	S.S.(10.0)	S.S.(10.0)
END PLATE OF OUTBOARD SIDE NOUURE	93 ~ 94	S.S.(10.0)	S.S.(10.0)
END PLATE OF INBOARD CENTER NOUURE	95 ~ 96	S.S.(10.0)	S.S.(10.0)
END PLATE OF INBOARD SIDE NOUURE	97	S.S.(10.0)	S.S.(10.0)

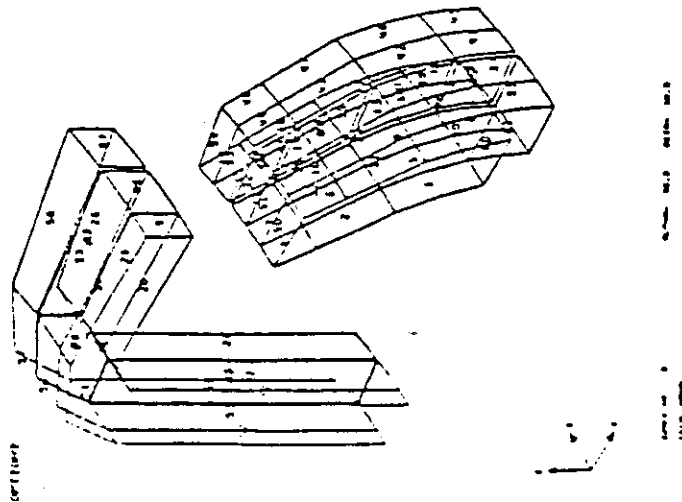


Fig. 4.3.1 Model and Electrical property of Blanket for Structural Analysis

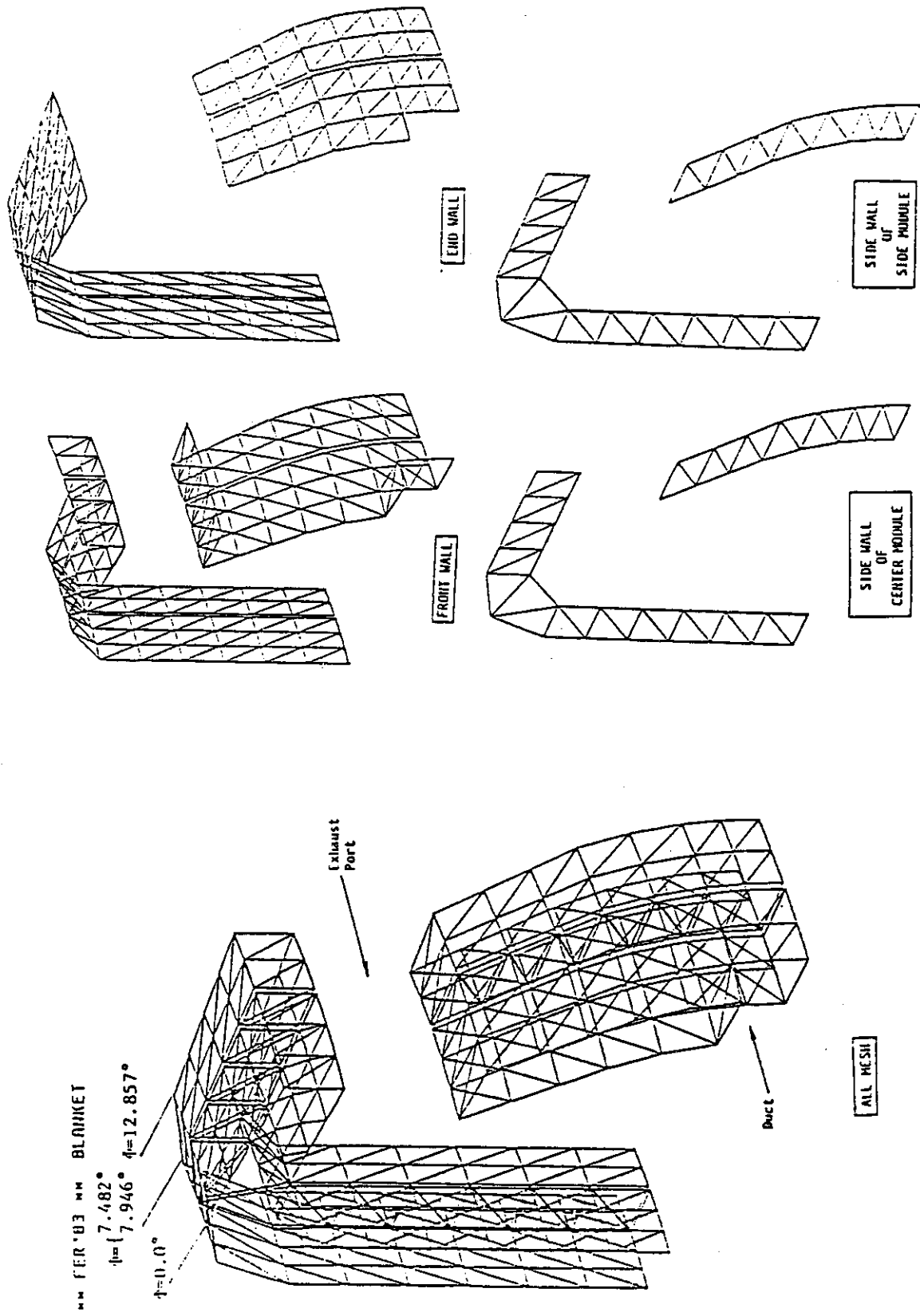
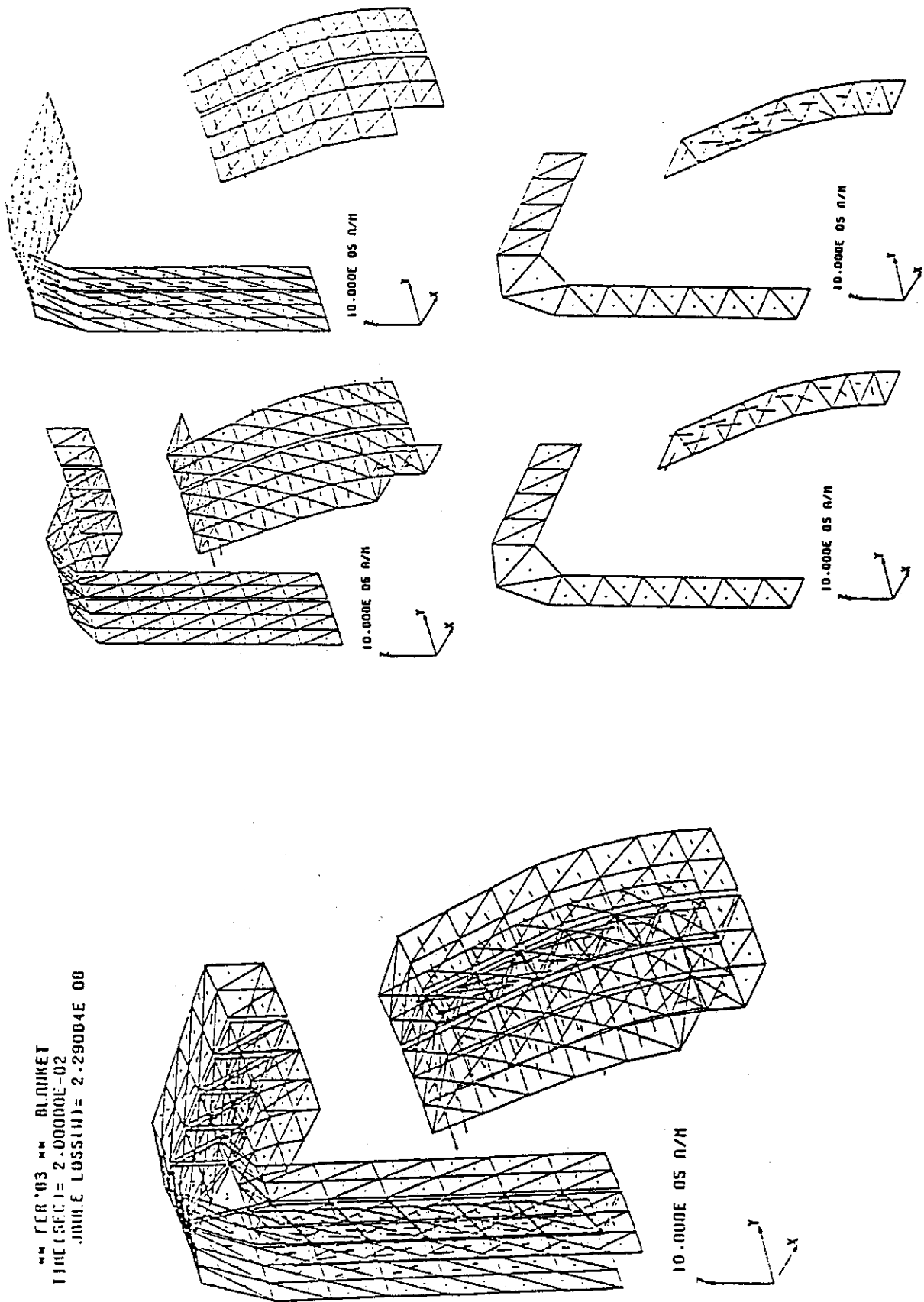


Fig. 4.3.2. Mesh of Blanket for Eddy Current and Electromagnetic Force



MM FER '03 MM BLINKET  
 TIME (SEC) = 2.00000E-02  
 J (A/M) = 2.29004E 00

Fig. 4.3.3 Eddy Current Distribution During Plasma Disruption  
 (Reference) (Time: 20msec) (Case 1)

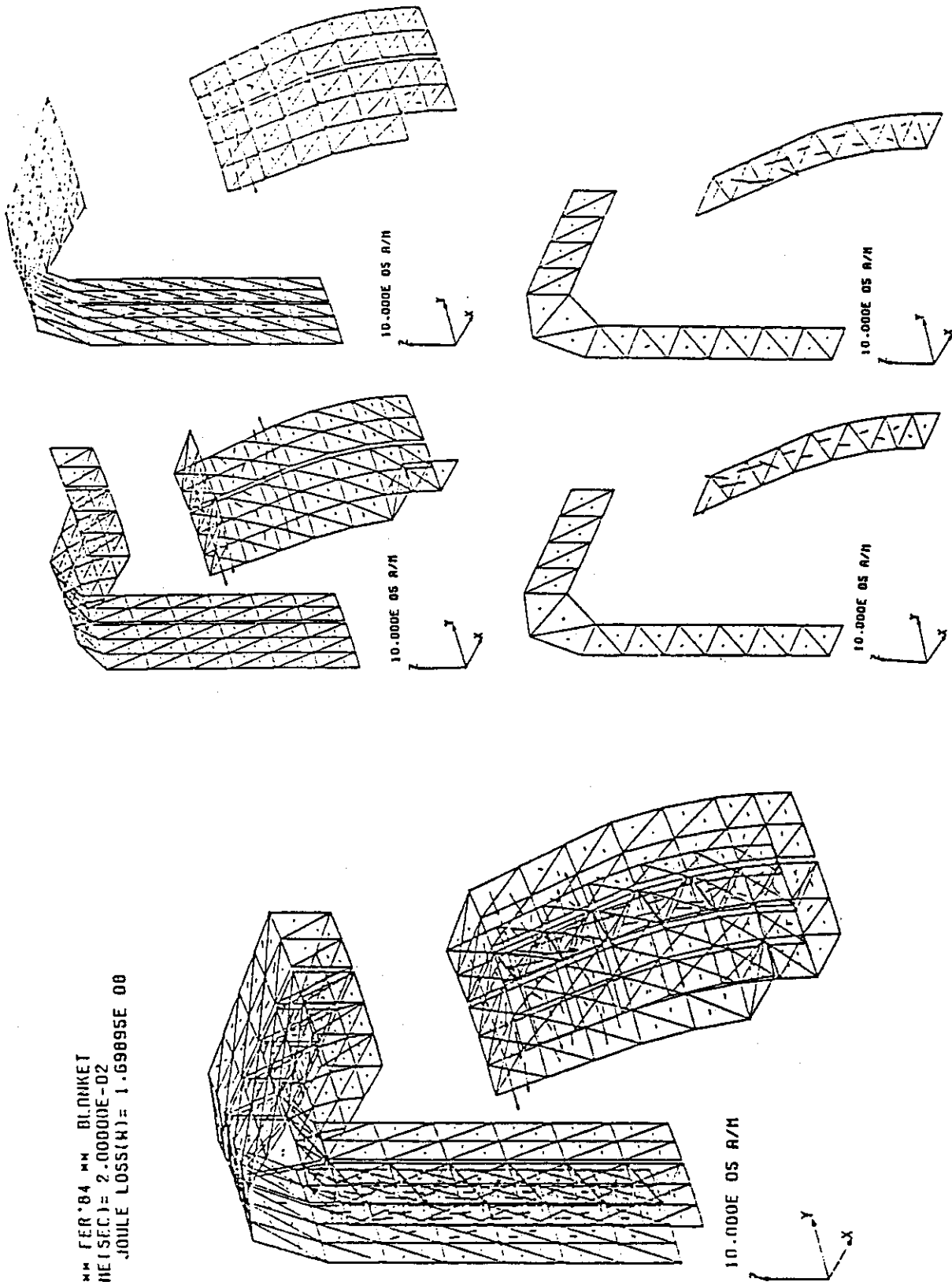


Fig. 4.3.4 Eddy Current Distribution During Plasma Disruption  
(Alternative) (Time: 20msec) (Case 1)

MM FER'84 MM BLANKET  
TIME(SEC) = 2.00000E-02  
JOULE LOSS(M) = 1.69895E 00

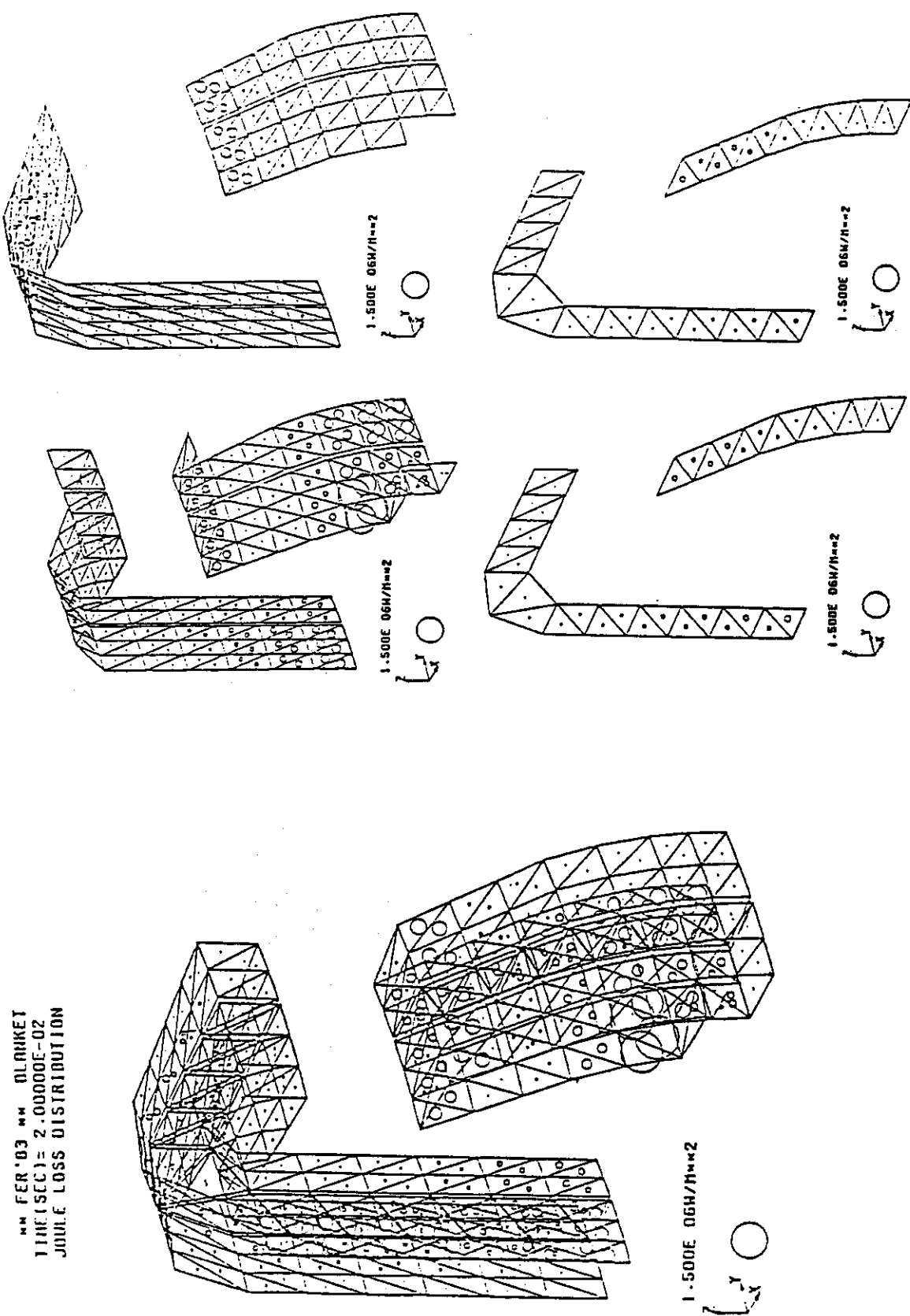


Fig. 4.3.5 Joule Loss Distribution During Plasma Disruption  
(Reference) (Time:20msec) (Case 1)

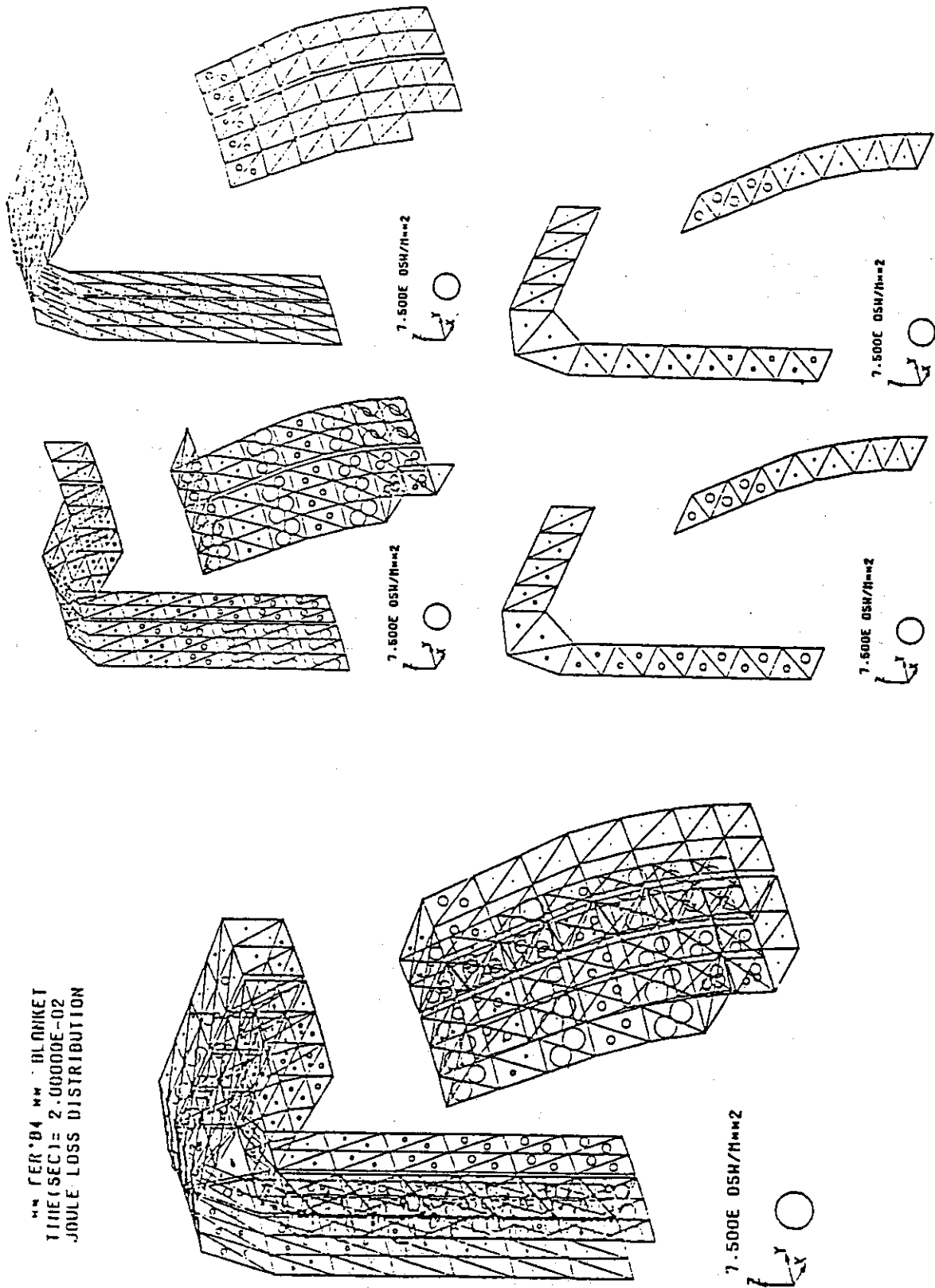


Fig. 4.3.6 Joule Loss Distribution During Plasma Disruption  
(Alternative) (Time:20msec) (Case 1)

PER'04 MM BLANKET  
TIME(SEC)= 2.00000E-02  
JOULE LOSS DISTRIBUTION

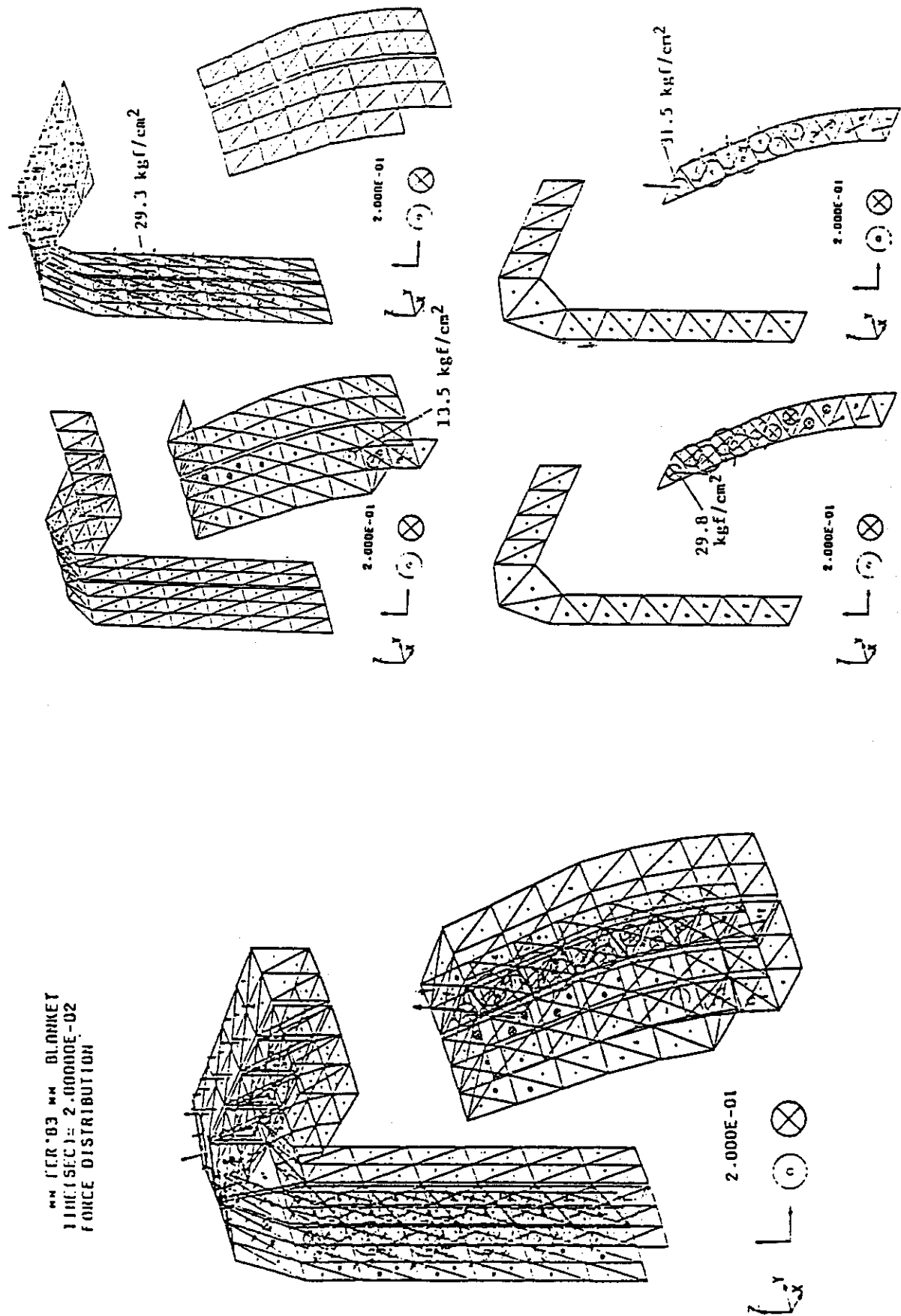


Fig. 4.3.7 Electromagnetic Force Due to Toroidal Magnetic Field  
(Reference) (Time: 20msec) (Case 1)



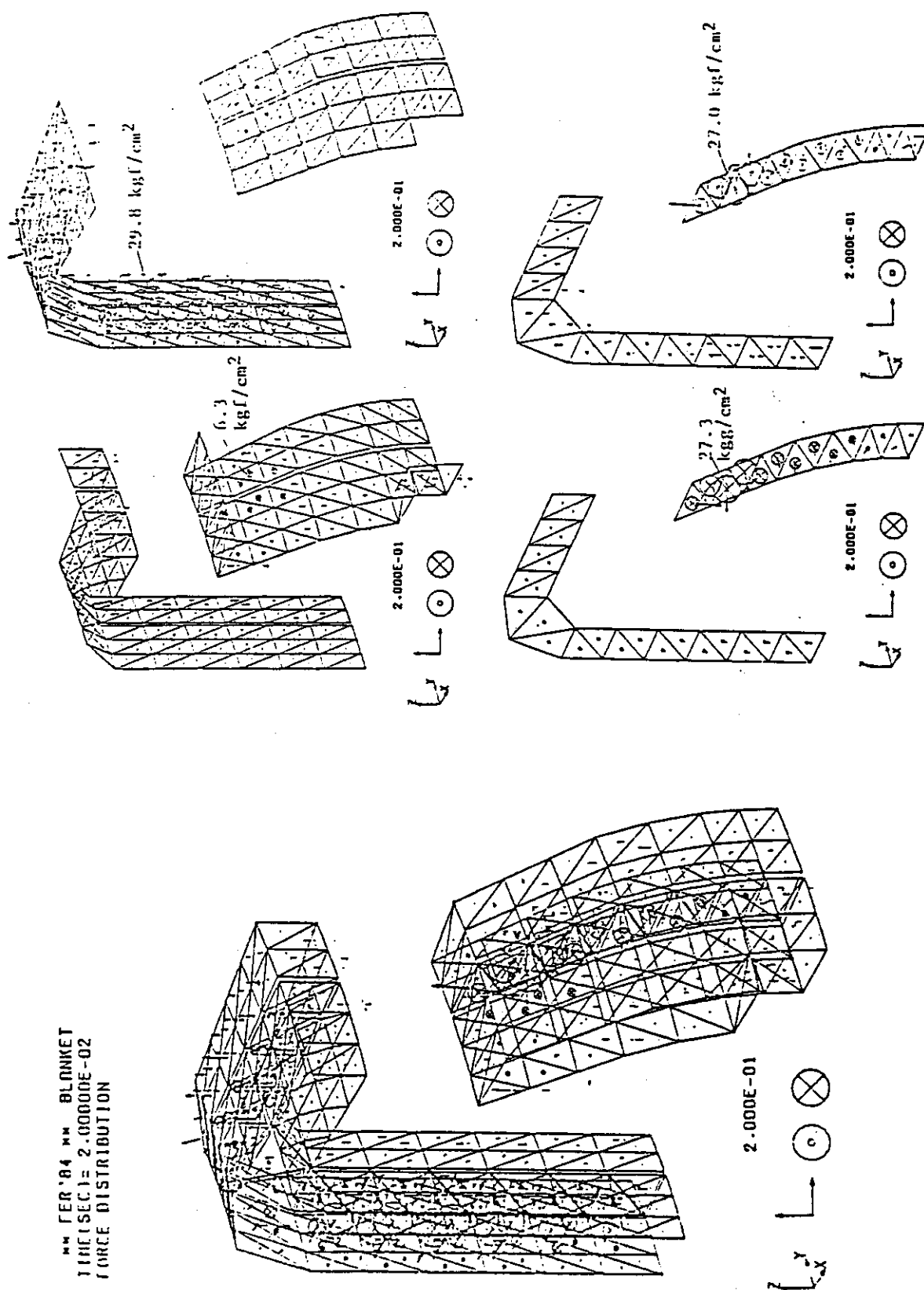


Fig. 4.3.8 Electromagnetic Force Due to Toroidal Magnetic Field  
(Alternative) (Time:20msec) (Case 1)

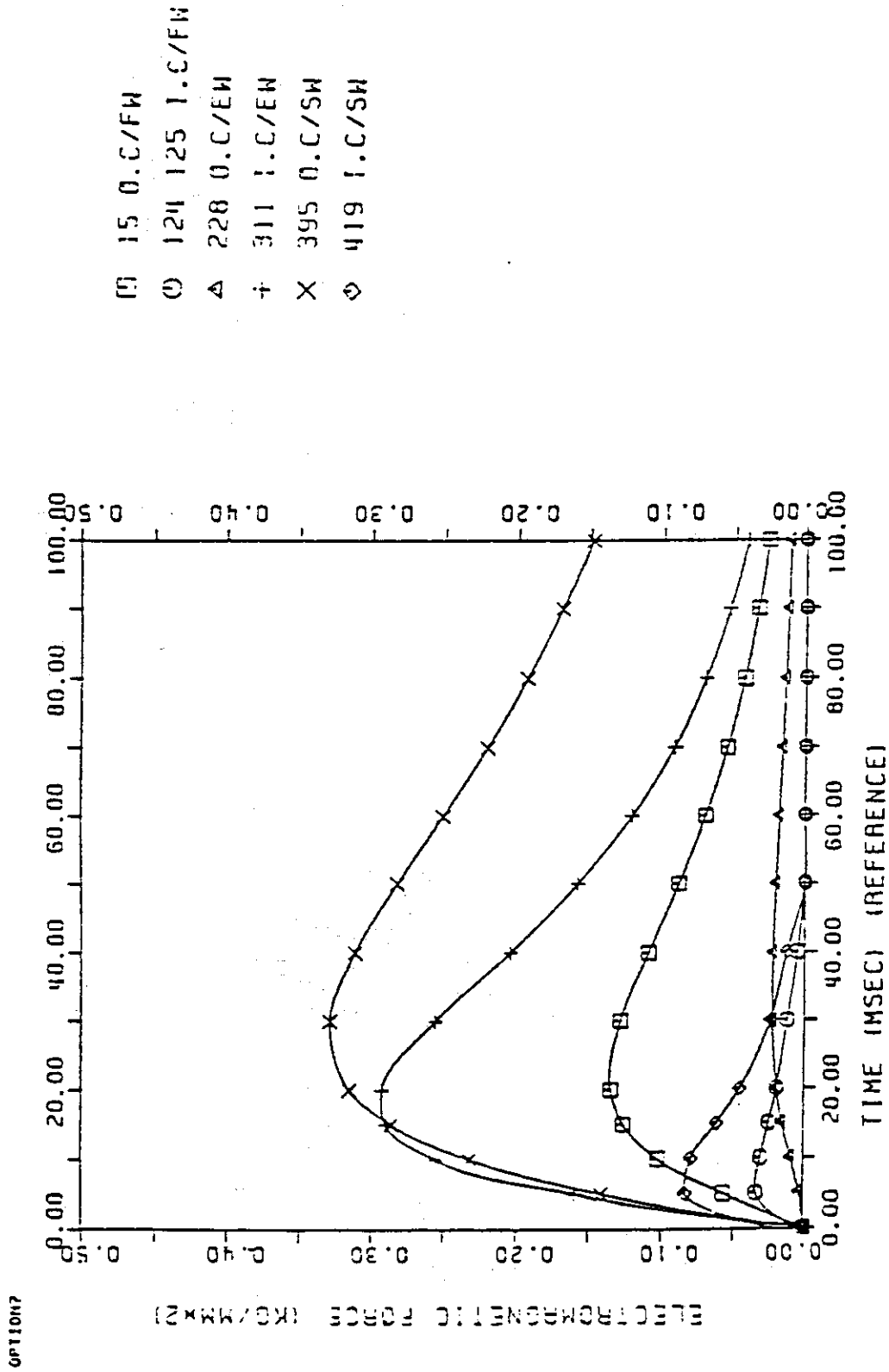


Fig. 4.3.9 Time Dependence of Electromagnetic Force Due to Toroidal Magnetic Field During Plasma Disruption (Reference) (Case 1)

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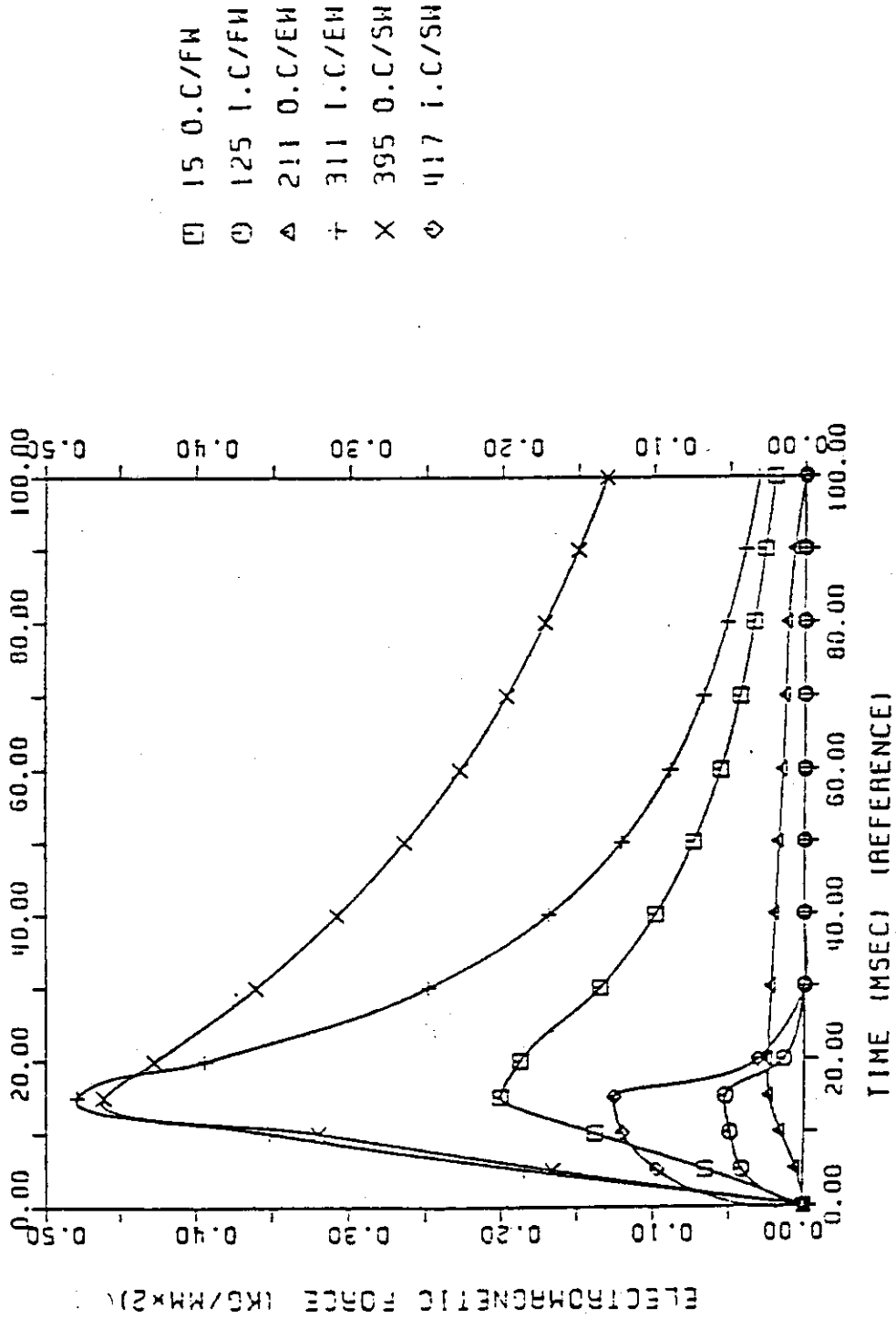


Fig. 4.3.10 Time Dependence of Electromagnetic Force Due to Toroidal Magnetic Field During Plasma Disruption (Reference), (Case 1)

## 5. Conclusions

### 5.1 Plasma Stabilization

Since the small number of conductive shell segmentation is preferable for stabilizing plasma vertical instability, the electrical connections between the adjacent conductive shells significantly improve the stabilizing property of passive shells. However, this kind of connection by remote handling makes remote maintenance procedure complicate and unreliable.

The conductive shells installed in inboard blanket modules is effective to stabilize the positional instability. However, the electromagnetic forces at plasma disruption would induce large stresses and deformations in the inboard blanket structures containing conductive shells, since their support could not be sufficient because of restricted accessibility behind the inboard blanket modules. Therefore, it should be better not to set up the inboard shells.

Saddle shape shells installed only in outboard blanket modules has been analytically found to stabilize vertical position instability of the JAERI FER, even when the active control coils are located outside the toroidal coils. This kind of vertical control system will provide the least impacts on overall tokamak reactor design.

### 5.2 Start-up

Magnetic and electrical field will be prevented by eddy currents to penetrate the radiation shield/coil vacuum chamber which surrounds plasma perfectly.

Since toroidal one turn resistance of larger than  $0.2\text{m}\Omega$  is not effective to penetrate plasma loop voltage as shown in section 3.1, it is adequate to set it at approximately  $0.2\text{m}\Omega$  if plasma current is inductively ramped up by Ohmic heating coils.

A large power supply capacity would be required for feedback control of plasma position when the radiation shield/coil vacuum chamber has high electrical conductance in the toroidal direction and the active control coils are located outside it, since a large gain value is needed to apply control magnetic field in plasma due to the magnetic shielding effect.

Toroidal one turn resistance of  $\sim 0.2\text{m}\Omega$  would also be necessary in order to prevent the excessive increase in power supply capacity of active control coils if they are installed outside the shield/coil chamber.

Summarizing the requirement on toroidal one turn resistance, approximately  $0.2\text{m}\Omega$  would be required if plasma current is ramped up inductively or active coils are located outside radiation shield/coil chamber to make active coil design and installation easy.

### 5.3 Disruption

Plasma disruptions induce high voltage and large electromagnetic forces in the components surrounding plasma. Rough estimation of induced loop voltage at blanket module would be less than  $\sim 2,000\text{V}$  which is equivalent to the voltage of 75V between adjacent blanket modules in the case of 24 blanket modules. However, it is desirable to take a safety margin of a few factor in order to specify the induced voltage at plasma disruptions.

The high conductive shells installed in blanket modules produce large electromagnetic forces at disruptions if the decay time of plasma current does not depend on shell properties of passive elements. Especially in the inboard blanket modules, the large induced force would generate excessive stresses and deformations in the blanket structures when they contain high conductive shells. Since the electromagnetic forces at disruptions depend on plasma models and current decay time and are generally too large in the next generation machine, it is necessary to specify the realistic disruption scenario in order to avoid their over estimation. And it will be fruitful to develop the control methods to increase plasma current decay time.

### 5.4 Design Guidelines

For plasma vertical movement stability, installation of active control coils located outside the toroidal field coils along with high-conductive shells in the FWBS is recommended based on the following considerations.

#### A. Design Guidelines for Active Control Coils

1. From the view-point of plasma vertical movement stabilization, preferably active control coils should be located as close to plasma as possible. When active control coils are to be installed inside the shield, passive shells are not necessarily required. However, such active control coils, being saddle-shaped modular type, could not be reliable systems because of the reasons mentioned below, and would bring in substantial risk into the reactor design. The problems associated with the internal active control coils include the following items;
  - 1) Insufficiency of data and practices for the inorganic insulator to be applied in the plasma vacuum region.
  - 2) Integrity of the coil structures including feeders against large electromagnetic forces.
  - 3) Connection/disconnection operations (including insulation work) of many feeders on assembling/disassembling of the FWBS modules for maintenance.
2. On the other hand, active control coils located outside the toroidal field coils require high-conductive shells installed in the FWBS to be able to increase the time scale of plasma

movement,  $\gamma^{-1}$ , to approximately 35 msec. The external active control coils, being designed with relaxed environmental conditions and not affected by the construction and maintenance of the FWBS modules, can be of high reliability, and the expected control power is about 35 MVA which is seemingly permissible.

#### B. Design Guidelines for Passive Shells

1. The most effective way for passive shell installation is to build saddle-shaped shell modules made of a high-conducting material in blanket modules, considering remote assembling and disassembling of torus structures.
2. It is not necessarily required to connect adjacent shell modules electrically. However, when assembly and maintenance procedures permit electrical connections between shell modules, the connections should be located just behind the blanket modules since those made at the outside of the shield would not contribute to improvement of shell effects so much.
3. It should be better not to install high conductive shells in inboard blanket modules, since the electromagnetic forces at disruptions would induce large stresses and deformations in the inboard blanket structures if they contain high conductive shells.

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7. Appendix

7.1 Data Base Assessment

7.1.1 Experimental Data for the Position Control Systems

JT-60 Plasma Control System Parameters

JFT-2M Equilibrium Control System Parameters

JT-60 Plasma Control System Parameters

T. Kimura  
I. Kondo

## I. Major parameters of JT-60

Table I Major parameters of JT-60

Major plasma radius	$R_p$	3.0(3.2)	m
Minor plasma radius	$a_p$	0.93(0.90)	m
Plasma current	$I_p$	2.7(2.1)	MA
Volt second	$\phi$	28.5	Vs
Toroidal magnetic field	$B_t$	4.5	T
NBI heating			
power into torus	$P_{NB}$	20	MW
accelerating voltage	$E_{NB}$	75-100	keV
pulse length	$T_{NB}$	10	s
LH heating			
power into torus	$P_{LH}$	18	MW
frequency	$f_{LH}$	1.7-2.3	GHz
pulse length	$T_{LH}$	10	s
ICRF heating			
power into torus	$P_{IC}$	5	MW
frequency	$f_{IC}$	0.11-0.13	GHz
pulse length	$T_{IC}$	10	s
Control magnetic field			
Vertical field	$B_v$	0.41(0.32)	T
coil current	$mI_v$	32 turns $\times$ 57.5	kA
Quadrupole field at $r = a_p$	$B_Q$	0.037	T
coil current	$mI_Q$	9 turns $\times$ 25	kA
Main divertor coil current	$mI_M$	8 turns $\times$ 120	kA
Maximum poloidal beta at			
maximum plasma current	$\beta_p$	2.5	
Maximum mean plasma pressure at			
maximum plasma current	$\Sigma_k \overline{n_k T_k}$	$2 \times 10^{21} (1.3 \times 10^{21})$	keV/m <sup>3</sup>
Elongation		none	
Equilibrium parameter	$\beta_p + 2I_1/2$	0.5 - 2.5	
Normalized vertical field	$B_v / (\mu_0 I_p / R)$	0.36(0.39)	
Plasma inductance	$L_p$	6.8	$\mu H$

( ) = divertor operation (magnetic limiter operation)

## II. Poloidal field system

Table II.1 Poloidal field system parameters (Windings)

Parameter	OH Coil	V Coil	H Coil	Q Coil	M Coil
Ampere Turns (MAT)	5.5	$\pm 1.84$	$\pm 0.12$	$\pm 0.5$	$\pm 0.755$
Total Turns	60	64	12	36	16
Max. Current (KA)	91.7	57.5	20	25	94.4
L ( $\mu$ H)	8.3	9.9	0.44	3.2	0.88
R(m $\Omega$ )* (at 75°C)	4.6	12.1	5.0	17.4	3.2
Time Constant (sec)	1.8	0.82	0.09	0.18	0.28
R·I <sup>2</sup> (MW)	38.7	40.0	2.0	10.9	28.5
1/2·LI <sup>2</sup> (MJ)	34.9	16.4	0.09	1.0	3.9
Equivalent Square					
Wave Pulse Length (sec)	8	6	7	7	6

\* not including feeders

Material Cu (0.2°/w Ag)

OH coil (or F coil) : Ohmic heating coil  
V coil : Vertical field coil  
H coil : Horizontal field coil  
M coil : Magnetic limiter coil  
(Divertor coil)

The layout of each coil is shown in Fig. 1.

Table II.2 Poloidal field system parameters (Power supply)

<hr/>		
OH coil power supply		
Type	24-phase thyrister converter	
Rated current	$\pm 92$	kA
Rated voltage	$\pm 2.5$	kV
Peak power	280(230)	MW
V coil power supply		
Type	24-phase thyrister/diode hybrid current	
Rated current	58	kA
Rated voltage	+10, -5	kV
Peak power	290	MW
Current step response (total feedback loop)	2-3	msec
H coil power supply		
Type	12-phase thyrister converter	
Rated current	$\pm 22$	kA
Rated voltage	$\pm 0.5$	kV
Peak power	11	MW
Current step response	2-3	msec
Q coil power supply		
Type	12-phase thyrister converter	
Rated current	$\pm 25$	kA
Rated voltage	$\pm 1.0$	kV
Peak power	25	MW
Current step response (total feedback loop)	2-3	msec
M coil power supply		
Type	12-phase thyrister converter	
Rated current	$\pm 120$	kA
Rated voltage	$\pm 1.0$	kV
Peak power	120	MW
Current step response (total feedback loop)	2-3	msec
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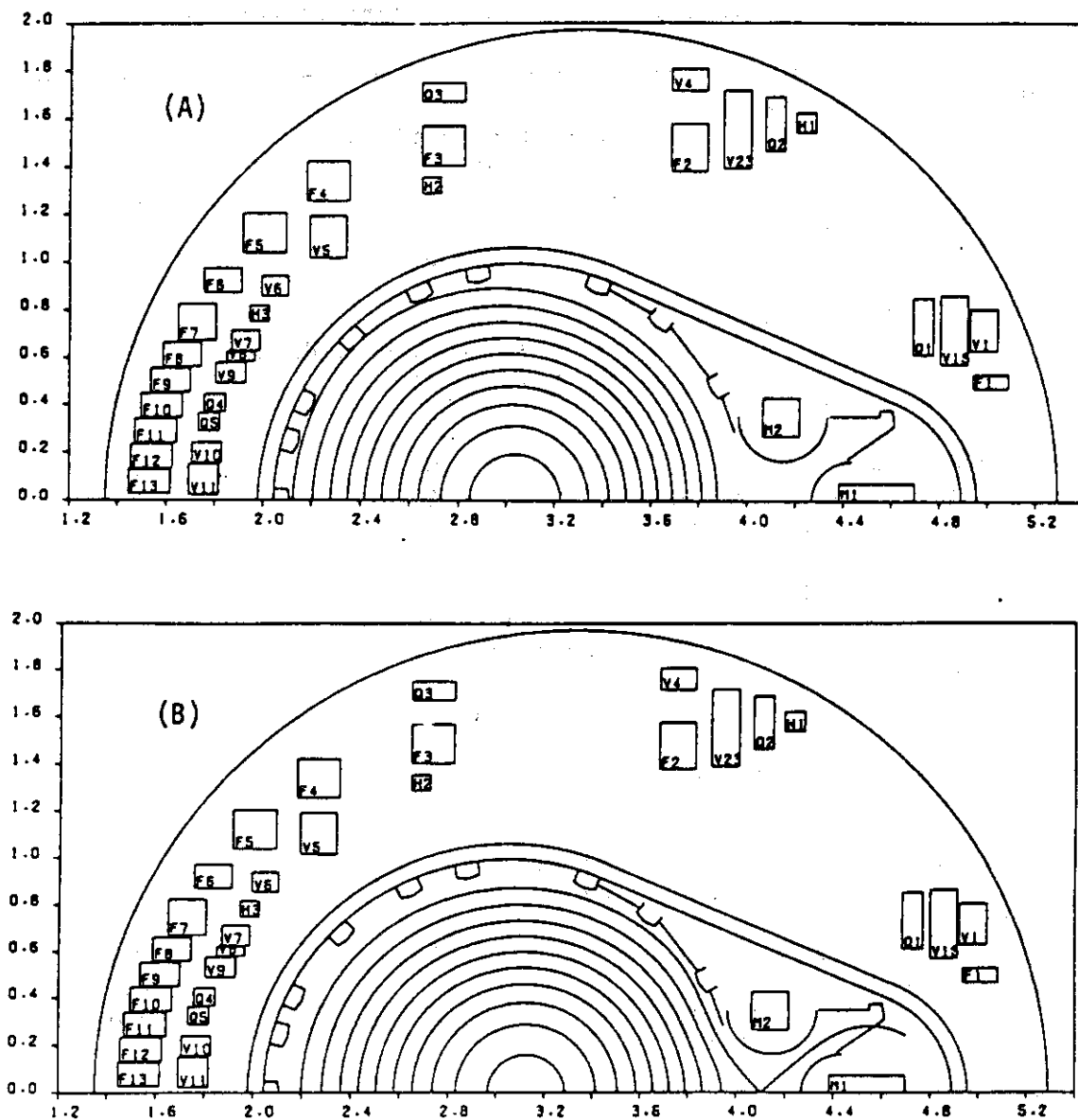


Fig. 1. Equilibrium configurations of JT-60 low  $\beta$  plasma.  
 (A) Fixed limiter operation ( $R_p = 3.0$  m,  $a_p = 0.88$  m,  
 $I_p = 1.5$  MA,  $I_V = 20$  kA,  $I_Q = 0$  kA)  
 (B) Divertor operation ( $R_p = 3.15$  m,  $a_p = 0.87$  m,  
 $I_p = 1.9$  MA,  $I_V = 21.3$  kA,  $I_Q = 18.0$  kA,  $I_M = 80$  kA)

[From M. Kikuchi et al., JAERI-M 84-095 (1984)]

### III. Real-time feedback control system

Conceptual block diagram of plasma real-time feedback control loops is shown in Fig. 2.

Main control purposes of the actuators are as follows:

i) Poloidal feed coil power supplies:

MHD equilibrium configuration

ii) NBI, RF heating systems:

Input power of additional heating

iii) Gas injection system:

Flow rate of gas injection

The real-time control computer  $I_b^R$  supervises each actuator and feedback control computer  $II_b$  with the period of 10 msec in the normal state and the period of 1 msec in the disruptive state.

The feedback control computer  $II_b$  controls the plasma MHD equilibrium configuration in conjunction with the DDC's (Direct Digital Controllers) in the poloidal field coil power supplies.

## A) MHD equilibrium configuration control

## 1. Plasma current control

(1) Current detection Rogowski coils

(2) Feedback control law:

$$V_F = K_F(I_{\text{pref}} - I_p)$$

where  $V_F$  : input control voltage to the power supply, $I_{\text{pref}}$ : reference input (preprogrammed input) for plasma current $I_p$  : plasma current

## 2. Horizontal position control

(1)  $R_p$  detection magnetic probes(2) VF index of curvature  $0 \leq n \leq 1.5$ 

$$n \equiv - \left. \frac{R}{B_z^{\text{ext}}} \frac{\partial B_z^{\text{ext}}}{\partial R} \right|_{R=R_p}$$

(3) Control law: PID control

$$Y_R = K_{RP} \{ \Delta X_R(i) - \Delta X_R(1) \} + \sum_{j=1}^i K_{RI} \Delta X_R(j) \\ + K_{RD} \{ \Delta X_R(i) - \Delta X_R(i-1) \},$$

where  $\Delta X_R = R_{\text{pref}} - R_p$  $\Delta X_R$  : deviation between reference input of plasma radial position  $R_{\text{pref}}$  and measured plasma radial position  $R_p$  at the time  $t_i$  ( $t_i = i \cdot \Delta t$ ) $K_{RP}$ ,  $K_{RI}$  and  $K_{RD}$  denote the PID gains for radial position control. Then, the input command for the power supply is

$$V_V^{\text{com}} = V_V^{\text{REF}} + Y_R \quad (\text{for AVR}),$$



$$\text{or } I_V^{\text{com}} = I_V^{\text{REF}} + Y_R \quad (\text{for ACR}),$$

where  $V_V^{\text{REF}}$  and  $I_V^{\text{REF}}$  are reference (or preprogrammed) input voltage and current for the power supply respectively.

### 3. Vertical position control

(same as horizontal position control)

- (1)  $Z_p$  detection : magnetic probes
- (2) Control law : PID Control

$$Y_Z = K_{ZP} \{ \Delta X_Z(i) - \Delta X_Z(1) \} + \sum_{j=1}^i K_{ZI} \Delta X_Z(j) \\ + K_{ZD} \{ \Delta X_Z(i) - \Delta X_Z(i-1) \} ,$$

where  $\Delta X_Z = Z_{\text{pref}} - Z_p$

Then,

$$V_H^{\text{com}} = V_H^{\text{REF}} + Y_Z \quad (\text{for AVR})$$

$$I_H^{\text{com}} = I_H^{\text{REF}} + Y_Z \quad (\text{for ACR})$$

### 4. Plasma shape control

- (1)  $\Delta Q_p$  detection : magnetic probes

$$\Delta Q \equiv \frac{1}{\mu_0 I_p} \frac{r_m^2}{2 r_p} (\lambda_2 + \mu_2)$$

$r_p$  : plasma minor radius,

$r_m$  : probe position

$\lambda_2$  : 2nd harmonic parameter of  $B_\theta$

$\mu_2$  : 2nd harmonic parameter of  $B_n$

## (2) Control law : PID control

(same as horizontal position control)

$$Y_Q = K_{QP} \{ \Delta X_Q(i) - X_Q(1) \} + \sum_{j=1}^i K_{QI} \Delta X_Q(j) \\ + K_{QD} \{ \Delta X_Q(i) - \Delta X_Q(i-1) \} ,$$

where  $\Delta X_Z = \Delta Q_{pref} - \Delta Q_p$  ,

Then,

$$V_Q^{com} = V_Q^{REF} + Y_Z \quad (\text{for AVR})$$

$$I_Q^{com} = I_Q^{REF} + Y_Z \quad (\text{for ACR})$$

## 5. Magnetic limiter coil current control

(1) Current detection: Rogowski coil and DCCT

(2) Control law

$$V_M^{COM} = V_M^{REF}$$

$$I_M^{COM} = I_M^{REF} \quad (I_M^{REF} = \alpha I_p^{REF})$$

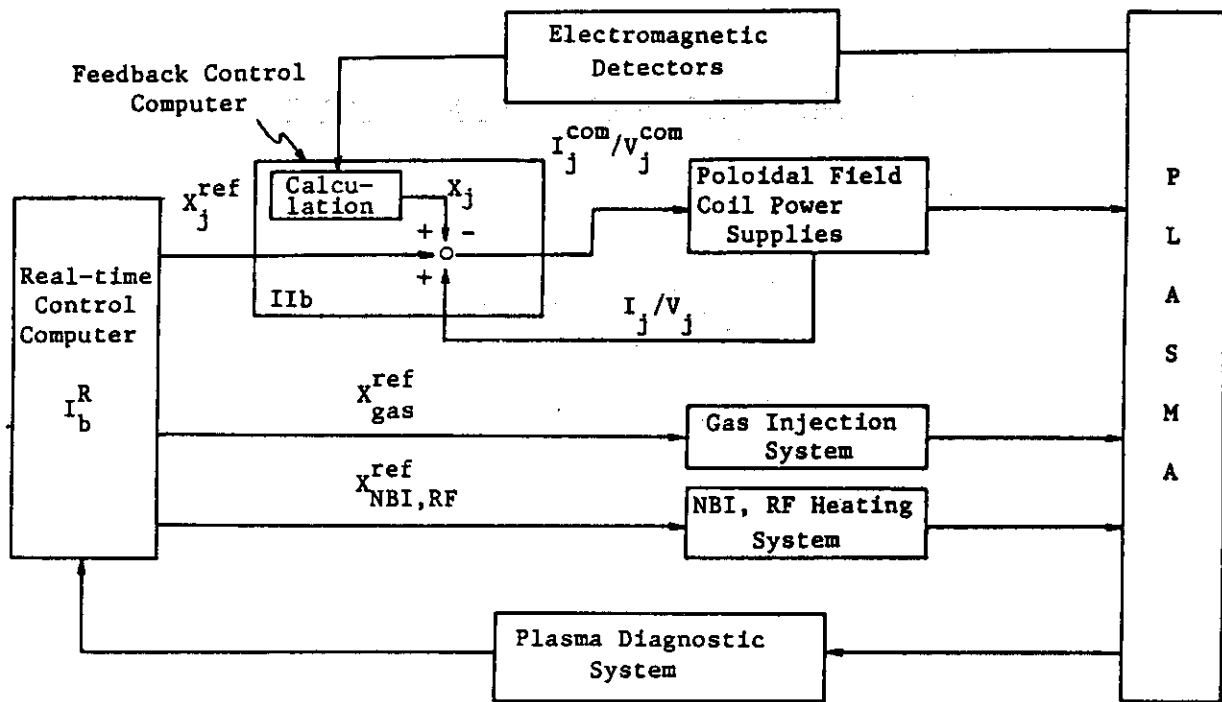


Fig. 2 Conceptual block diagram of plasma real-time feedback control system in JT-60

JFT-2M Equilibrium Control System Parameters

Yoshimi Matsuzaki  
Masahiro Mori

## I. Plasma

- A. Major radius  $R_0 = 1.31$  m
- B. Minor radius  $a, b = 0.35, 0.53$  m
- C. Elongation  $k = 0.9 - 1.4$
- D. Plasma current  $I_p = 500$  KA
- E. Poloidal beta  $\beta_p = 0.3 - 1.1$
- F. Plasma inductance  $L_p \sim 2\mu\text{H}$

## II. Poloidal Field System in JFT-2M

## A. Iron Core Transformer

- 1. maximum flux swing 2V.s
- 2. structure 1 core + 1 return yoke

## B. Poloidal Coils

- 1. coil configuration Fig. 1
- 2. coil parameters Table 1
- 3. vertical field strength and configuration Table 2

## C. Poloidal Power Supply

## 1. OH power supply

- a. main load coils OH coils
- b. capacitance bank for start-up
  - max. voltage 6 kV
  - max. capacitance 7.5 mF
  - storage energy 135 kJ
  - switching thyristor

## c. thyristor power supply

- phase 12
- max. DC current 3.6 kA

max. DC voltage	0.7 kV
peak power	2.52 MW
transformer	3.712 kVA
control type	analog, ACR/AVR

## 2. S power supply

a. main load coils  $S_1 - S_2 - S_{3/1}$  coils

## b. capacitance bank for start-up

max. voltage	5 kV
max. capacitance	28.8 mF
storage energy	360 kJ
switching	ignitron

## c. thyristor power supply

phase	12
max. DC current	8 kA
max. DC voltage	1.2 kV
peak power	9.6 MW
transformer	13,900 kVA
control type	analog, ACR/AVR

## 3. V up power supply

a. main load coils  $V_1 - V_2 - S_{3/2}$  upside coils

## b. thyristor power supply (positive)

phase	12
max. DC current	8.25 kA
max. DC voltage	0.275 kV
peak power	2.27 MW
transformer	4,096 kVA
control type	analog, ACR/AVR

## c. thyrister power supply (negative)

phase	12
max. DC current	7.5 kA
max. DC voltage	0.275 kV
peak power	2.06 MW
transformer	3,725 kVA
control type	analog, ACR/AVR.

## 4. V low power supply

a. main load coils  $V_1 - V_2 - S_{3/2}$  downside coils

## b. thyrister power supply (positive)

phase	12
max. DC current	8.25 kA
max. DC voltage	0.275 kV
peak power	2.27 MW
transformer	4,096 kVA
control type	analog, ACR/AVR.

## c. thyrister power supply (negative)

phase	12
max. DC current	7.5 kA
max. DC voltage	0.275 kV
peak power	2.06 MW
transformer	3,725 kVA
control type	analog, ACR/AVR.

## 5. Q power supply

a. main load coils  $Q_{1/1} - Q_{1/2} - Q_2$  coils

## b. thyrister power supply

phase	12
max. DC current	10 kA

max DC voltage	0.25 kV
peak power	2.5 MW
transformer	3,646 kVA
control type	analog, ACR/AVR.

## 6. H power supply

a. main load coils	H coils
b. thyristor power supply	
phase	12
max. DC current	7.5 kA
max. DC voltage	0.48 kV
peak power	3.6 MW
transformer	5,520 kVA
control type	digital, ACR/AVR.

## 7. B power supply

a. main load coils	B coils
b. thyristor power	
phase	6
max. DC current	0.27 kA
max. DC voltage	0.58 kV
peak power	0.16 MW
transformer	200 kVA
control type	analog, ACR.



### III. Plasma Control System

#### A. Plasma Current Control

1. current detection : Rogowski loops
2. current control : pre-programmed currents of (S, OH, Q) coils. ACR operation of thyrister power supply.

#### B. Position Control

1.  $\Delta R$  detection : magnitic probes and saddle loops (iso-flux method)
2.  $\Delta Z$  detection : saddle loops
3. principle of control : horizontal and vertical hybrid control (Fig. 2)
4. VF index of curvature  $-1.3 < n < 0$
5. feedback control law

$$V_o^U + \tau_f \dot{V}_o^U = A I_p (\Delta R + \tau_D^R \dot{\Delta R}) + B I_p (\Delta Z + \tau_D^Z \dot{\Delta Z})$$

$$V_o^L + \tau_f \dot{V}_o^L = A U_p (\Delta R + \tau_D^R \dot{\Delta R}) - B I_p (\Delta Z + \tau_D^Z \dot{\Delta Z})$$

$V_o^U$  : input control voltage to thyrister of  $V_{up}$

$V_o^L$  : input control voltage to thyrister of  $V_{low}$

$$\tau_f = 0.5 \text{ ms} \quad \tau_D^R = 30 \text{ ms} \quad \tau_D^Z = 15 \text{ ms}$$

A, B : feedback gains (variable)

$I_p$  : plasma current

6. feedback calculation : analog circuit

#### C. Shape Control

preprogramed currents of (S, OH, Q) coils.

#### D. Passive Element

vacuum vessel time constant  $\sim 7 \text{ ms}$

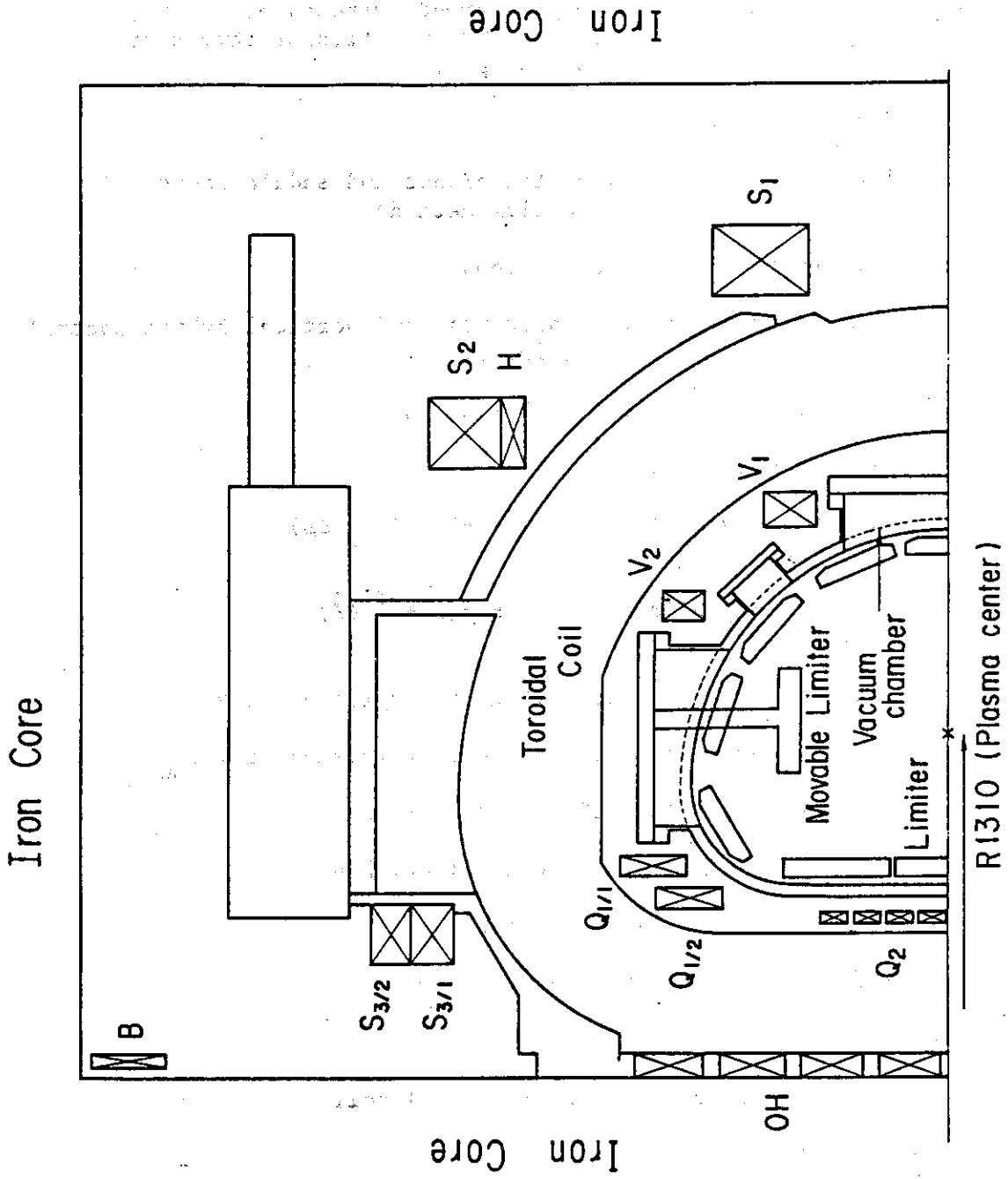


Fig. 1 Quarter cross section of JFT-2M

Table 1 Poloidal coils parameters

Coils	Position of coils		Turns (upside) (changable turns)	Max. Current (kA)	Max. AT (kAT)	Resistance (m $\Omega$ )	Inductance (air core) (mH)	Cooling	Main Objects	Main Power Supply
	R (m)	Z (m)								
S <sub>1</sub>	2.34	0.4325	56 (28, 42)	4.3	241	44.1	32.3	Water	joule plasma shaping and control	S
S <sub>2</sub>	1.945	1.115	42 (14)	4.3	181	27.5	15.0	Water		
S <sub>3/1</sub>	0.785	1.1905	24 (12, 18)	4.3	103	6.3	16.7	Water		
S <sub>3/2</sub>	0.785	1.2865	14 (8, 10, 12)	7.5	105	2.0	0.41	Water	fast control of plasma position	V up & V low
V <sub>1</sub>	1.770	0.365	10 (6, 8)	7.5	75	3.2	0.86	—		
V <sub>2</sub>	1.545	0.61	6 (4)	7.5	45	1.7	0.28	—		
Q <sub>1/1</sub>	0.9375	0.68	8	8.5	68	1.4	0.24	—	non- circularity control	Q
Q <sub>1/2</sub>	0.8625	0.6	8	8.5	68	1.3	0.22	—		
Q <sub>2</sub>	0.82	0.036 , 0.252	16	8.5	136	3.0	0.52	Water		
OH	0.476	0 , 0.722	42	3.6	151	11.9	—	Water	V · s control	OH
H	1.945	1.005	14	4.3	60	18.3	1.88	—	vertical position	H
B	0.480	1.883	15	0.267	4	13.2	0.34	—	V · s compensation	B

(series connections upside and downside coils)

Table 2 Magnetic field of poloidal field coils in JFT-2M

Coils	Radial field $B_r$ (G)	Vertical field $B_z$ (G)	$n^1) B_z$ (G)
$S_1$	- 105	189	- 237
$S_2$	- 208	57	42
$S_3/1$	- 225	- 105	30
$S_3/2$	- 223	- 96	32
$V_1$	- 257	301	- 253
$V_2$	- 336	100	409
$Q_1/1$	- 258	- 162	81
$Q_1/2$	- 238	- 196	- 32
$Q_2$	- 100	- 358	- 810
OH	- 125	- 259	- 370
H	- 212	74	39
B	- 206	- 62	31
Plasma	0	- 180	- 202

\*) This is the value per 100 kAT and of upper coils only at the center of plasma  $R = 1.31$  m

$$1) \quad n = - \frac{R}{B_z} \frac{\partial B_z}{\partial R}$$

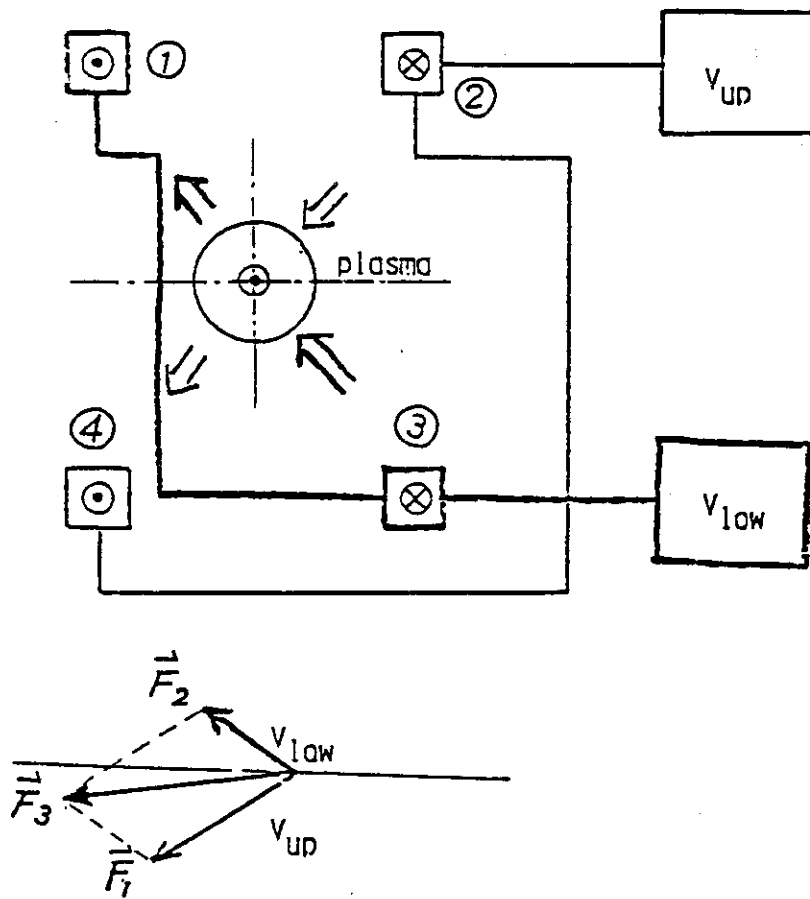


Fig. 2 Schematic diagram of horizontal and vertical hybrid control in the JFT-2M

JFT-2M Fast Pulse Power Supply Technology

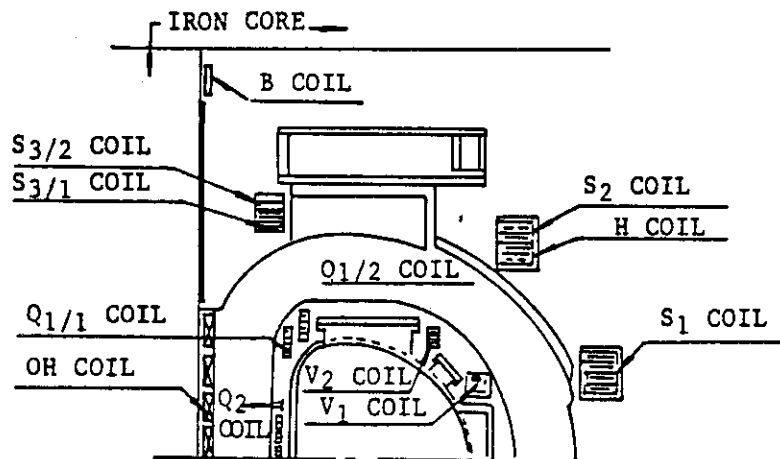
K. Ueda

Table 1 Main Parameters of JFT-2M

Plasma Current (KA)	550
Major Radius (m)	1.31
Minor Radius (m)	0.35
Elongation	1 ~ 1.6
n-Index	0.4 ~ -2.0
Vertical Field (T)	-0.05 ~ -0.10
Poloidal Beta	2.0
OH Flux Swing (Vs)	2.0
NBI Heating Power (MW)	2.0
ICRF       "       (MW)	4.5
LHRH       "       (MW)	1.0
ECRH       "       (MW)	0.2

Table 2  
Name and Function of PF Coils

No.	Name of Coil Group	Member's Name	Function	Power Supply for Excitation
1	Ohmic Heating Coil	OH	Produces main magnetic flux to pass through iron core, and heats plasma.	OH Power Supply
2	Shaping Coil (S)	$S_1, S_2, S_{3/1}$	Produces main magnetic flux to pass through iron core, and produces plasma equilibration magnetic field.	S Power Supply
3	Quadruple Field Coil (Q)	$Q_{1/1}, Q_{1/2}, Q_2$	Produces equilibration magnetic field in non-circular cross-section plasma, and produces magnetic limiter defining magnetic field.	Q Power Supply
4	Vertical Field Coil (V)	$V_1, V_2, S_{3/2}$	Corrects the variable component of the vertical magnetic field required for equilibrating the plasma having the tendency toward high $\beta$ being accompanied with heating power input of NBI, ICRF etc. Feedback controls for locating horizontal direction of plasma.	V Power Supply
5	Horizontal Field Coil (H)	H	Feedback controls for locating vertical direction of plasma, and corrects magnetic field in horizontal direction.	H Power Supply
6	Bias Coil (B)	B	For obtaining the total magnetic flux of 2 v.s. for iron core, gives pre-excitation of -1 v.s.	B Power Supply



The arrangement of PF coils



Table 3 Main Parameters with PF Coils

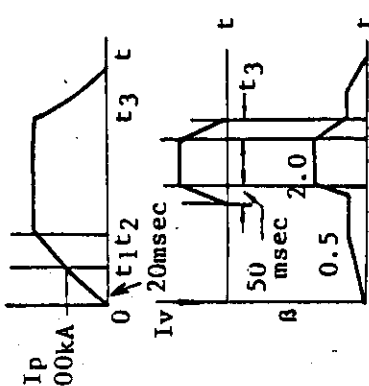
Coil Name \ Items	Exciting current (KA)	Number of Turns	Ampere Turns (KAT)	Intermediate Taps
S	S <sub>1</sub>	54	240.8	0, 28, 42, 56
	S <sub>2</sub>	42	180.9	0, 14, 42
	S <sub>3/1</sub>	24	103.2	0, 6, 12, 24
V	V <sub>1</sub>	10	150.0	0, 6, 8, 10
	V <sub>2</sub>	6	90.0	0, 4, 6
	S <sub>3/2</sub>	24	198.0	0, 5, 10, 12, 14
Q	Q <sub>1</sub>	8	136.0	0, 8
	Q <sub>1/2</sub>	8	136.0	0, 8
	Q <sub>2</sub>	8	136.0	-
OH	1.8	42	75.0	0, 42
N	1.1	24	26.0	0, 14

Table 4 Power Supply of PF Coils

Name of Power Supply	Power Supply for Excitation	Power Supply Bank	Power Supply	Thyristor Control System
S	Thyristor	S	480 V 7.5 KA (2420 V 4.3 KA)	P.P
	Capacitor	S <sub>0</sub>	20 KV 50 mF	
		S	10 KV 7.2 mF or 5 KV 28.8 mF	
OH	Thyristor	OH	350 V 3.6 KA (700 V 3.6 KA)	P.P
	Capacitor	OH	6 KV 7.5 mF or 3 KV 30.0 mF	
		S-OH	10 KV 3.2 mF	
V	Thyristor (cross connection rectifier circuit)	V <sub>up</sub> (+) V <sub>up</sub> (-) V <sub>Low</sub> (+) V <sub>Low</sub> (-)	275 V 8.25 KA (-275 V 5.75 KA) 275 V 8.25 KA (-275 V 5.75 KA)	P.P - F.B
Q	Thyristor	Q	(250 V 8.5 KA)	P.P
H	Thyristor (Same as V)	M <sub>up</sub> (+) M <sub>Low</sub> (-)	(500 V 6.0 KA) (-530 V 6.6 KA)	P.P - F.B
B	Thyristor	B	520 V 0.27 KA	D.C

\* P.P : Pre-programming, F.B : Feedback control

Table 5

ITEMS CROSS- SECTION	TIME INTERVAL	POWER (MVA)	TURNS OF COILS	CURRENT (KA)	MAXIMUM VOLTAGE		CALCULATION CONDITION												
					$0 < t \leq t_1$ (V)	$t_2 < t$ (V)													
CIRCULAR	$0 - t_1$	8.3	OH -21	2.46	1,514	338	<div>PLASMA CURRENT BUILD-UP TIME</div> 												
	$t_1 - t_2$	15.5	<table><tr><td>S1</td><td>S2</td><td>S3/1</td></tr><tr><td>-14</td><td>-42</td><td>24</td></tr><tr><td>-14</td><td>-42</td><td>12</td></tr></table>	S1	S2	S3/1		-14	-42	24	-14	-42	12	3.86	6,945	1,765			
	S1	S2	S3/1																
	-14	-42	24																
-14	-42	12																	
$t_2 - t_3$	20.0	<table><tr><td>V1</td><td>V2</td><td>S3/2</td></tr><tr><td>-6</td><td>-6</td><td>12</td></tr></table>	V1	V2	S3/2	-6	-6	12	4.75	-	<table><tr><td>351</td></tr><tr><td>-232</td></tr></table>	351	-232						
V1	V2	S3/2																	
-6	-6	12																	
351																			
-232																			
$0 - t_1$	6.6	OH -47	2.5	2,881	626														
ELLIPSE	$t_1 - t_2$	14.9	<table><tr><td>S1</td><td>S2</td><td>S3/1</td></tr><tr><td>-56</td><td>21</td><td>-</td></tr><tr><td>-56</td><td>21</td><td>-21</td></tr></table>	S1	S2	S3/1	-56	21	-	-56	21	-21	3.42	9,089	2,236				
	S1	S2	S3/1																
	-56	21	-																
-56	21	-21																	
$t_2 - t_3$		<table><tr><td>V1</td><td>S3/2</td><td></td></tr><tr><td>-10</td><td>10</td><td></td></tr></table>	V1	S3/2		-10	10		4.5	-	<table><tr><td>317</td></tr><tr><td>-204</td></tr></table>	317	-204						
V1	S3/2																		
-10	10																		
317																			
-204																			
D SHAPE	$0 - t_1$		OH -42	3.48	2,724	626	<div>CIRCULAR</div> $t_1 = 20 \text{ msec}$ $t_2 = 100 \text{ msec}$ $I_p = 200 \text{ KA AT } t_1$ $350 \text{ KA AT } t_2$  <div>ELLIPSE</div> $t_1 = 20 \text{ msec}$ $t_2 = 150 \text{ msec}$ $I_p = 200 \text{ KA AT } t_1$ $450 \text{ KA AT } t_2$  <div>D SHAPE</div> $t_1 = 20 \text{ msec}$ $t_2 = 200 \text{ msec}$ $I_p = 200 \text{ KA AT } t_1$ $550 \text{ KA AT } t_2$												
	$t_1 - t_2$	19.4	<table><tr><td>S1</td><td>S2</td><td>S3/1</td></tr><tr><td>-56</td><td>-21</td><td>18</td></tr><tr><td>-56</td><td>-</td><td>6</td></tr><tr><td>-56</td><td>-21</td><td>24</td></tr></table>	S1	S2	S3/1		-56	-21	18	-56	-	6	-56	-21	24	3.2	8,350	2,140
	S1	S2	S3/1																
	-56	-21	18																
-56	-	6																	
-56	-21	24																	
		<table><tr><td>V1</td><td>V2</td><td>V3/2</td></tr><tr><td>-10</td><td></td><td>10</td></tr></table>	V1	V2	V3/2	-10		10	5.48	-	<table><tr><td>355</td></tr><tr><td>-236</td></tr></table>	355	-236						
V1	V2	V3/2																	
-10		10																	
355																			
-236																			
$t_2 - t_3$	24.3	<table><tr><td>Q1/1</td><td>Q1/2</td><td>Q2</td></tr><tr><td>8</td><td>8</td><td>-4</td></tr><tr><td>0</td><td>8</td><td>8</td></tr></table>	Q1/1	Q1/2	Q2	8	8	-4	0	8	8	7.5	216	216					
Q1/1	Q1/2	Q2																	
8	8	-4																	
0	8	8																	

7.1.2 Irradiation Effects of Active Coil Insulation Materials

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## 1. Introduction

Insulators is necessary to prevent conducting between adjacent conducting wires, turns, and layers in active control coils and to reduce AC loss while magnetic field is changing.

The two classes of materials which can be selected for electrical insulation in fusion reactor magnets are inorganic ceramics and organic polymers. In contrast to inorganic ceramics, organic polymers are superior in fabricability and cost. Organic polymers, however, are much less resistant to radiation damage than inorganic ceramics [1-6]. Leading candidates for magnet insulators are organic polymers.

It is desirable that active coils will be near plasma. But active coils with superconductors will be positioned outside of TF coils while that with normal conductors will be positioned between blankets and TF coils.

The materials data base for mechanical and electrical properties of organic polymers is presented.

## 2. Organic insulation materials

The organic insulators for fusion magnets are required to have good radiation resistance.

Fig.2.1. gives a survey of the radiation resistance of several organic materials at 77K [7]. This survey is based on mechanical properties. According to this survey, epoxies, polyimides, and polystyrenes are possible candidates. Almost all magnet designs for fusion applications incorporate epoxy-fiberglass laminates [1]. But recent data shows that unfilled and glass-filled polyimides have a factor of 5-10 more radiation resistance than the glass-filled epoxies [8].

Candidate organic insulators for fusion magnets are listed in Table 2.1. [8-13].

### 3. Mechanical properties

The irradiation effects on tensile strength, compression strength, and flexure strength of epoxies and polyimides are shown in Fig.3.1.-Fig.3.6. [9, 10, 12-17]. Irradiation sources are neutron and gamma ray. Irradiation temperatures are all liquid helium temperature. Test temperatures are liquid temperature, liquid nitrogen temperature, and room temperature. Dose for insulators are generally reported in Gy (or rad), which is the unit of absorbed dose.

The G-10CR and the G-11CR after irradiation at 4.9K and warm-up to 307K and tested at 77K showed severe decrease in compression strength and flexure strength after a gamma dose of 5.5MGy [15].

The Vessel of a pure polyimide retained initial strength in compression and flexure after a gamma dose of 100MGy [13, 17].

The Norplex and Spaulrad of glass-fabric-filled polyimides showed decrease in compression strength and flexure strength after a gamma dose of 50MGy [17]. These materials remained a half of initial strength in compression and flexure after dose of 100MGy [13, 17].

The comparison of epoxies and polyimides in flexure strength is shown in Fig.3.7. [9, 15, 17].

Radiation effects in several polymers at room temperature is shown in Fig.3.8. [18, 19]. At room temperature, polyimides, epoxies, and polystyrenes are possible candidates for fusion magnet insulators with lifetime dose of a few 10MGy.

Tensile fatigue curves of unirradiated and irradiated G-11CR are shown in Fig.3.9. [20]. The samples were irradiated to gamma dose of 1MGy at 295K and the fatigue testing was performed at 295K and 77K. At both temperatures, the fatigue resistance of the G-11CR was increased below 3000 cycles and decreased above 3000 cycles.

### 4. Electrical properties

DC resistivity of various insulators after irradiation at liquid helium temperature is shown in Fig.4.1. [11, 12]. Resistivity measurements are made at liquid nitrogen temperature and room temperature.

The G-10CR and the G-11CR decreased a tenth of their initial resistivity after a gamma dose of 10MGy.

The Kapton (pure polyimide) retained initial resistivity after 100MGy.

Electrical breakdown of various insulators after irradiation at liquid helium temperature is shown in Fig.4.2. [11, 12]. The voltage-breakdown tests performed at room temperature.

The G-10CR, G-11CR, and stycast showed a significant drops in the breakdown voltage after a dose of 100MGy.

## 5. Material evaluation and future work needed

As shown in section 3 and 4, irradiation data base of organic insulation materials is insufficient to design fusion magnet.

It is difficult to evaluate sufficiently insulators for fusion magnet from existing data base of mechanical and electrical properties.

The glass-fabric-filled epoxies are little damaged by doses of less than 5MGy. There are some concern in using these materials for insulators of magnets over the doses of 20MGy. More studies of irradiation effect on properties, especially mechanical properties, of epoxies for the dose of 20MGy are necessary.

The pure polyimide is unaffected by doses of 100MGy.

The glass-fabric-filled polyimides are not seriously damaged by doses of less than 50MGy. These materials are possible candidates for insulators in fusion magnets for the doses of less than 100MGy at least. The data accumulation of irradiation effects on properties of polyimides for the doses more than 100MGy is necessary.

It is necessary to study radiation response of candidate materials as a function of dose, dose rate, irradiation spectrum, thermal history, atmosphere, and stress condition [21].

And it is necessary to consider gass processing, fabricability, and cost as well as mechanical and electrical properties of organic insulators in selecting insulators for fusion magnets.

Table 2.1. Composition of candidate organic insulators

Materials	Matrix Resines & (Hardener)	Reinforcements	Reinforcements Weight Contents
G-10CR	Bisphenol A (heat activated amine-catalyzed)	E-glass cloth	67wt%
G-11CR	Bisphenol A (aromatic amine)	E-glass cloth	70wt%
Epikote 828 G-20	Bisphenol A (aliphatic amine)	E-glass fiber	72wt%
Epikote 828 G-22	Bisphenol A (aromatic amine)	E-glass fiber	75wt%
GFRP A	Bisphenol A (acid anhydride)	E-glass fiber	74wt%
GFRP B	Bisphenol A (aliphatic amine)	E-glass fiber	75wt%
Stycast 2850FT	Epoxy ; 7% 24LV	Inorganically filled	?
Epon 828	Epoxy ; 0.5% Z6020 Silane couplant	400-mesh SiO <sub>2</sub>	40wt%

Table 2.1. Composition of candidate organic insulators (continue)

Materials	Matrix Resins & (Hardener)	Reinforcements	Reinforcements Weight Contents
Vespel (Kapton)	Unfilled polyimide	-	-
Norplex (Kerimide)	Bismaleimide (aromatic diamine)	E-glass cloth	40-60wt%
TIL-G1000	Polyimide	E-glass cloth	65wt%
Spaulrad (Spauldite)	Aromatic polyimide	E-glass cloth	70-71wt%
Nomex	Aromatic nylon paper	-	-



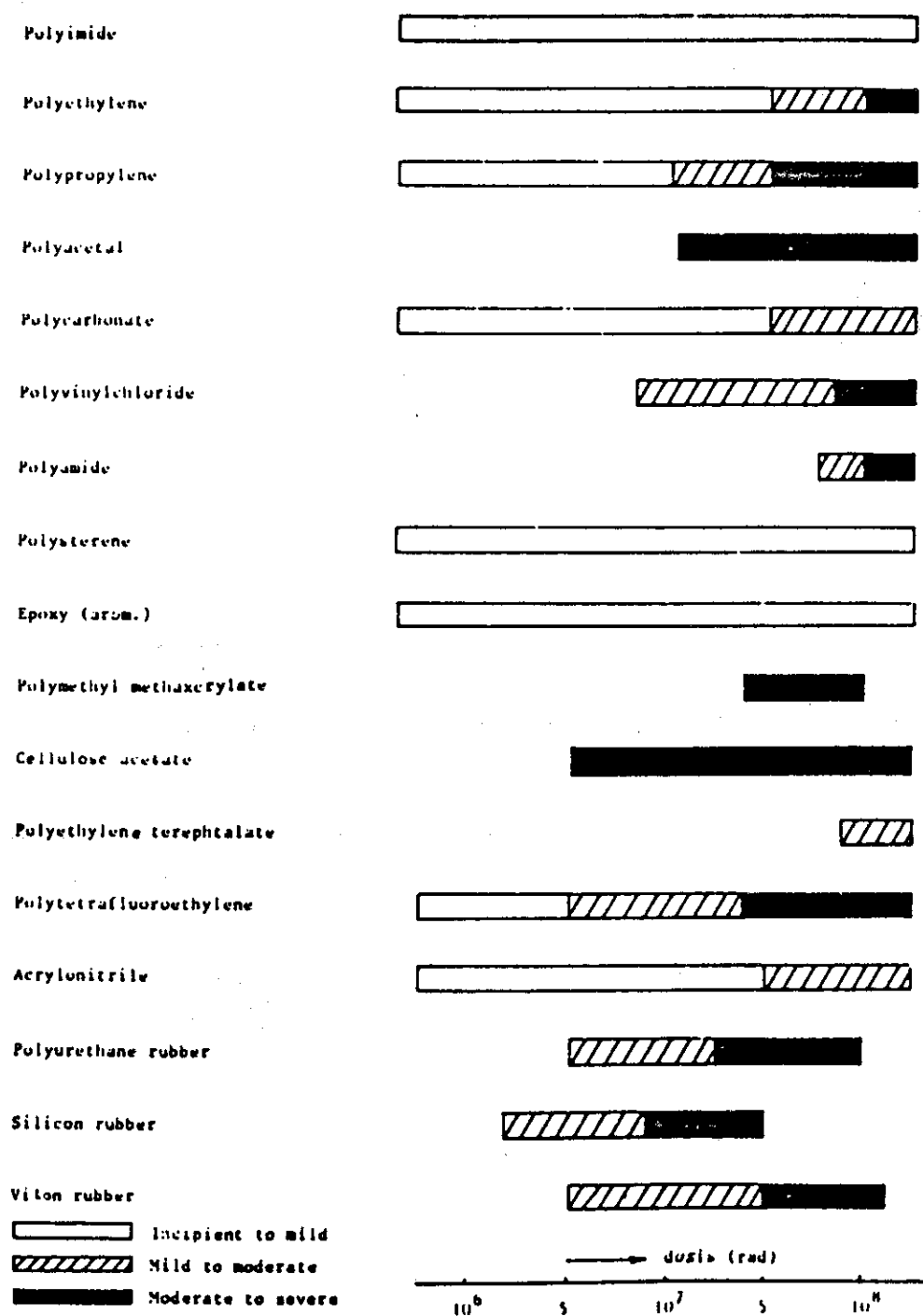


Fig.2.1. Radiation effects in several polymers at 77K.

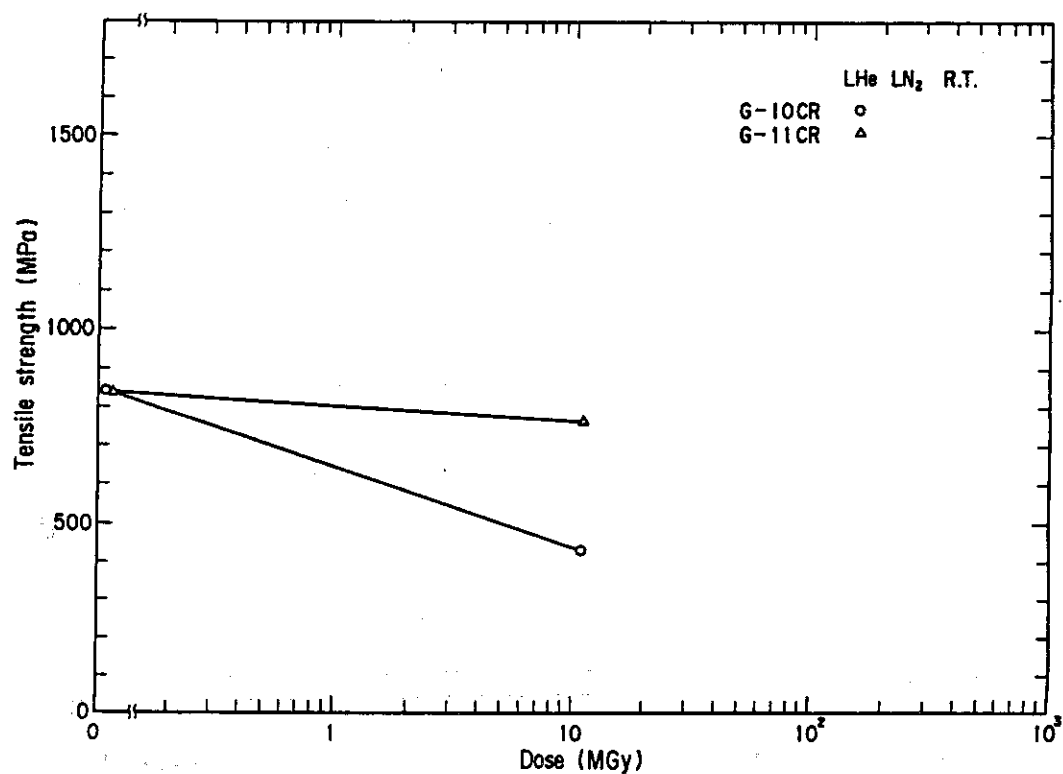


Fig. 3.1 Irradiation effect on tensile strength of glass-fabric-filled epoxies.

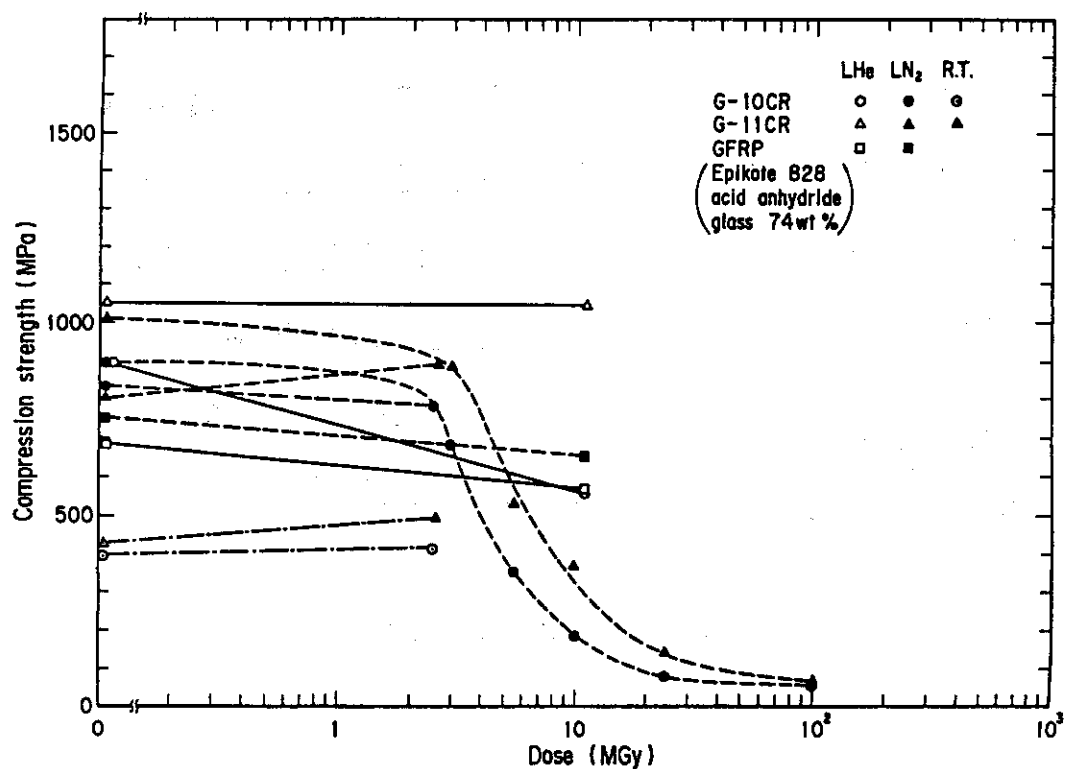


Fig. 3.2 Irradiation effect on compression strength of glass-fabric-filled epoxies.

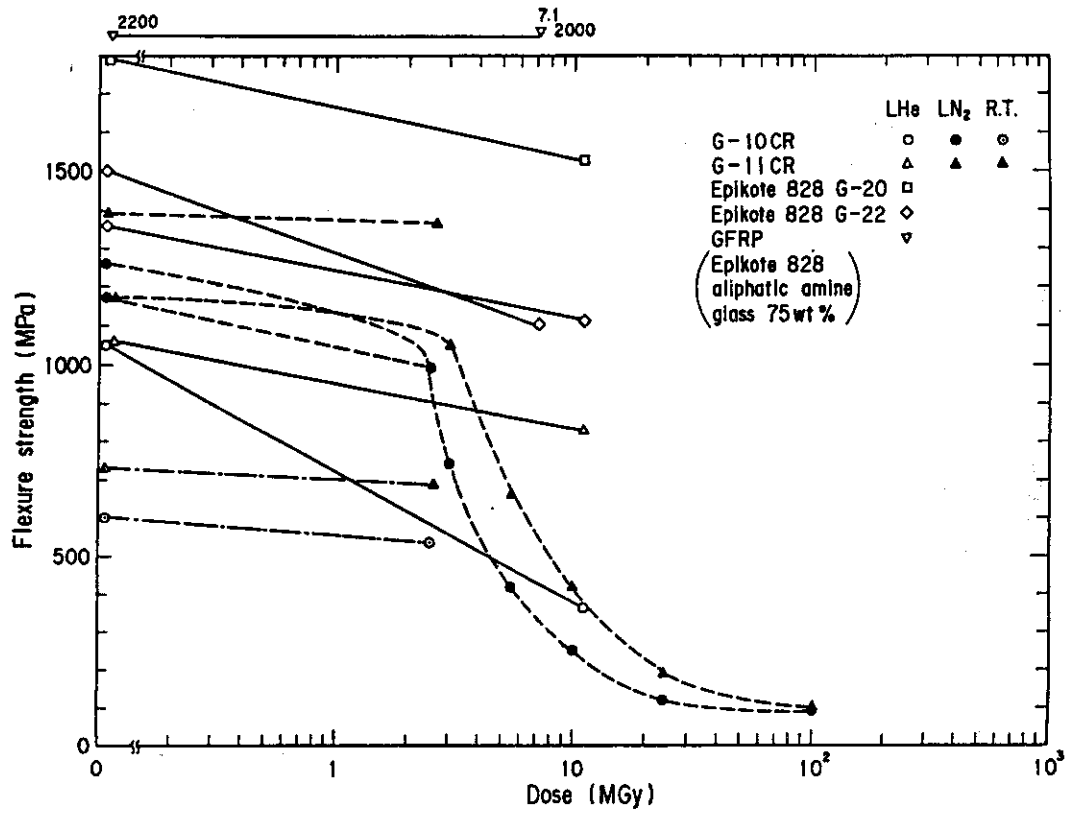


Fig.3.3 Irradiation effect on flexure strength of glass-fabric-filled epoxies.

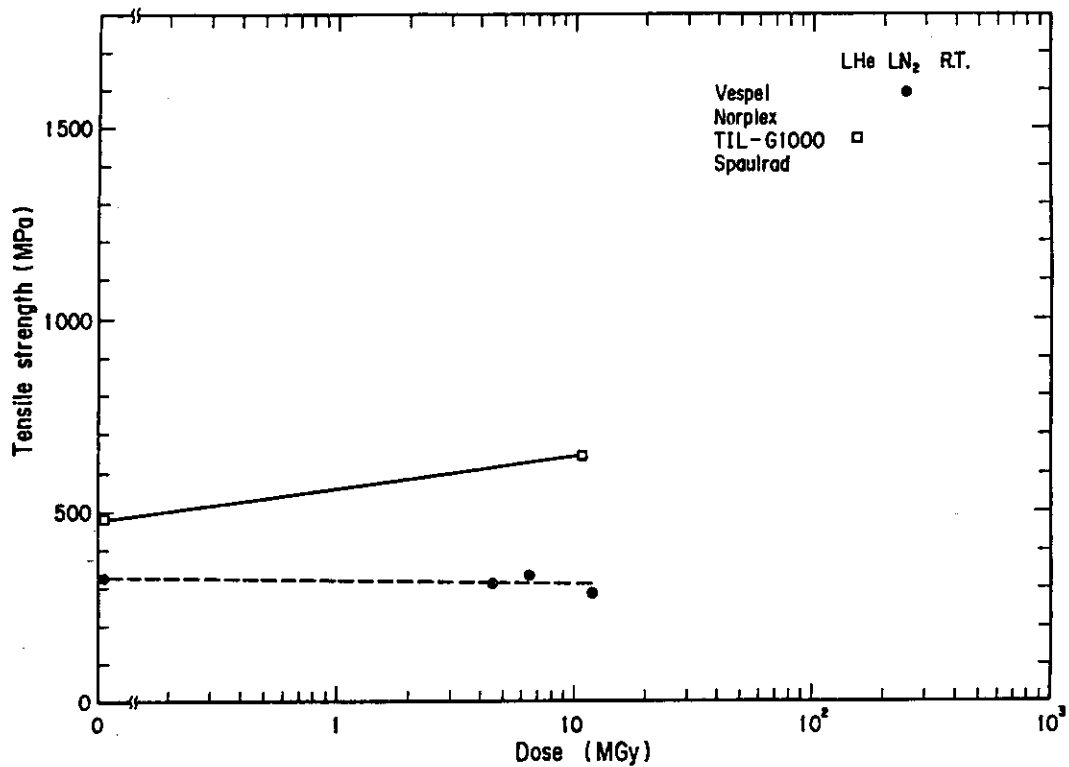


Fig.3.4 Irradiation effect on tensile strength of pure and glass-fabric-filled polyimides.

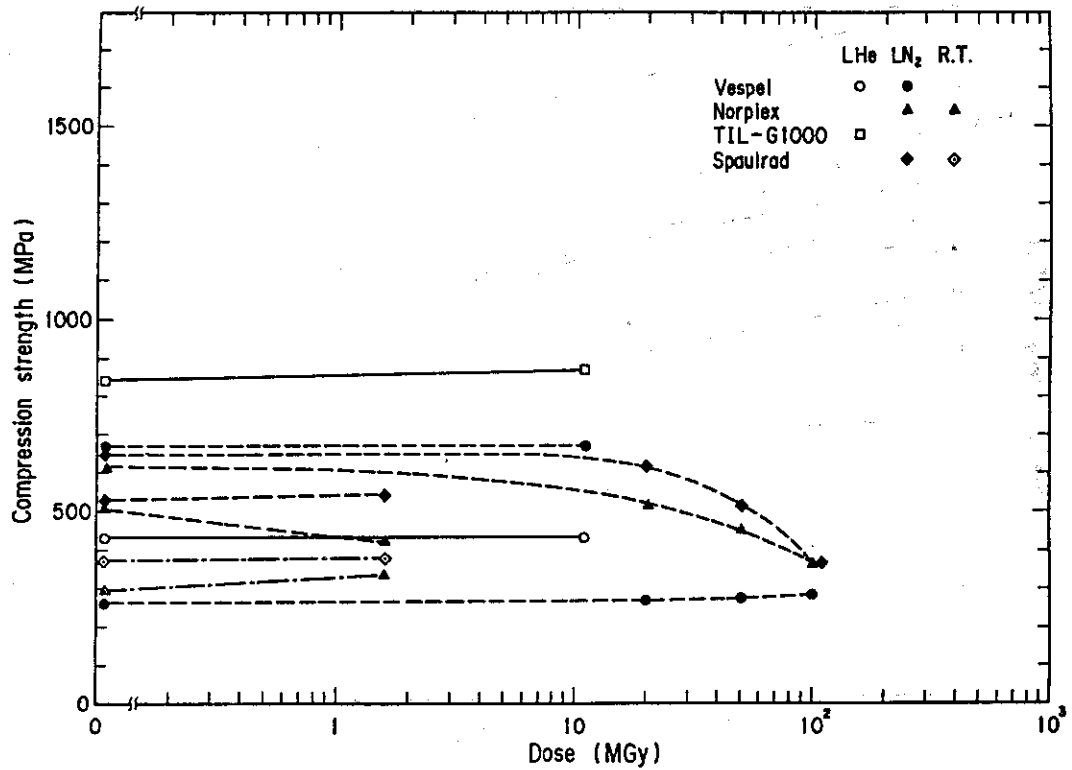


Fig.3.5 Irradiation effect on compression strength of pure and glass-fabric-filled polyimides.

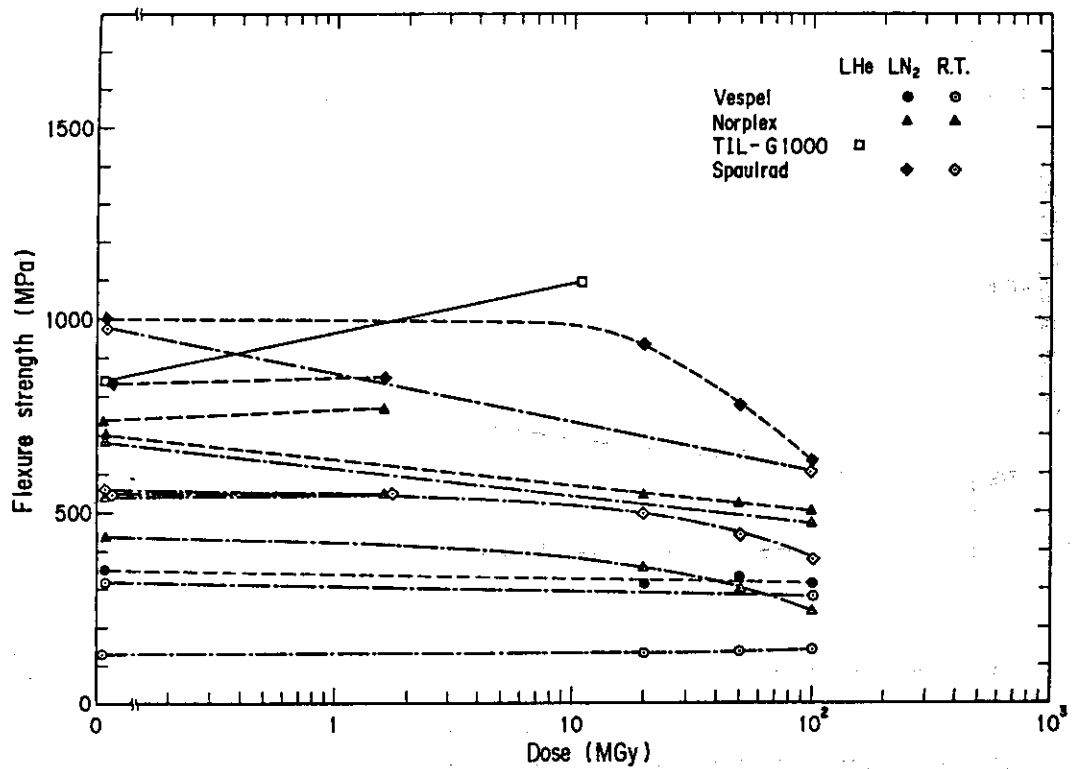


Fig.3.6 Irradiation effect on flexure strength of pure and glass-fabric-filled polyimides.

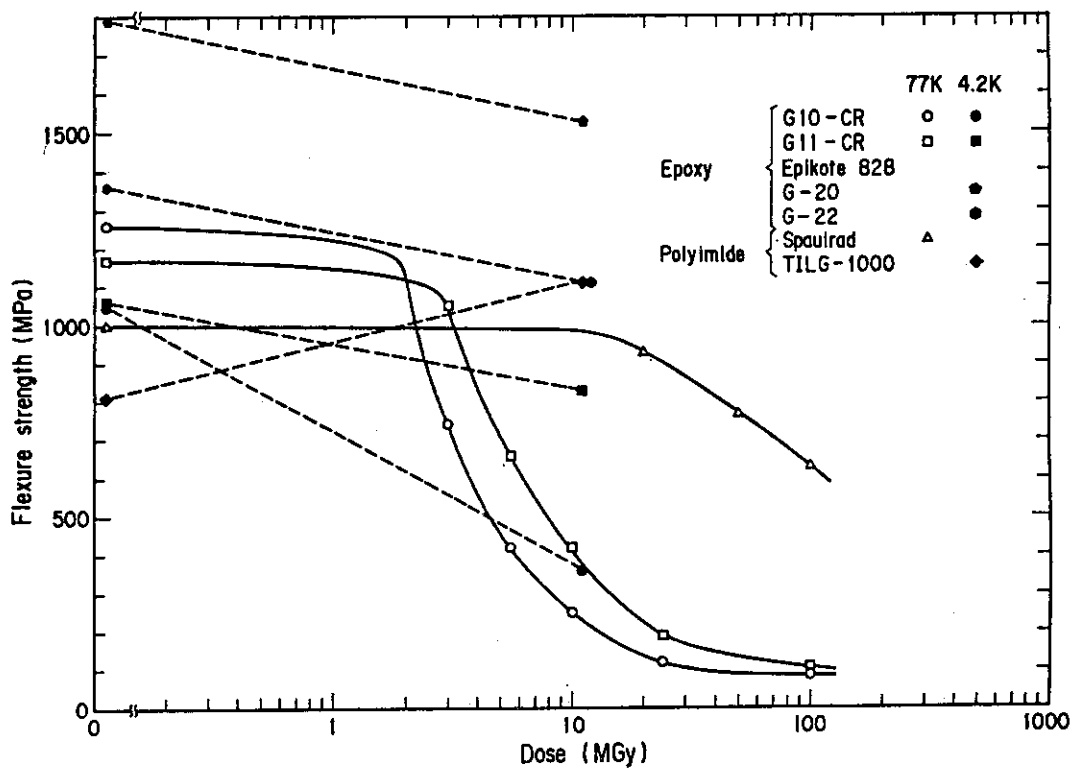


Fig.3.7 Irradiation effect on flexure strength of epoxies and polyimides.

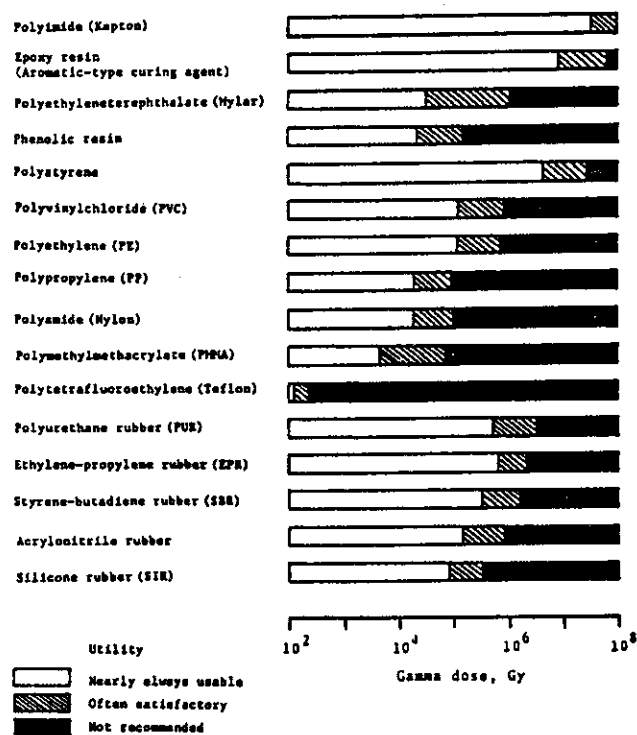


Fig.3.8 Radiation effects in several polymers at room temperature.

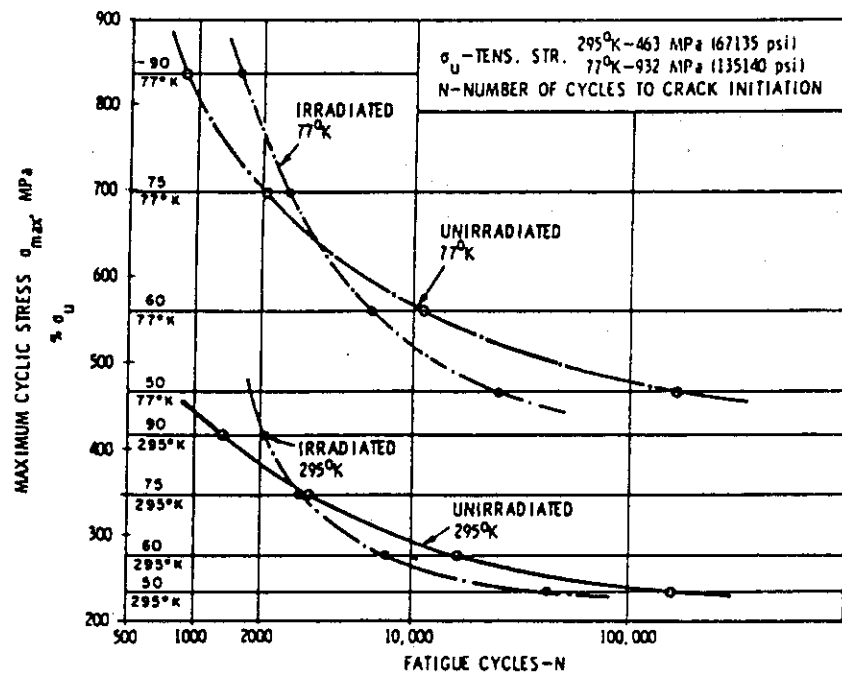


Fig.3.9 Tensile fatigue curves of unirradiated and irradiated G-11CR.

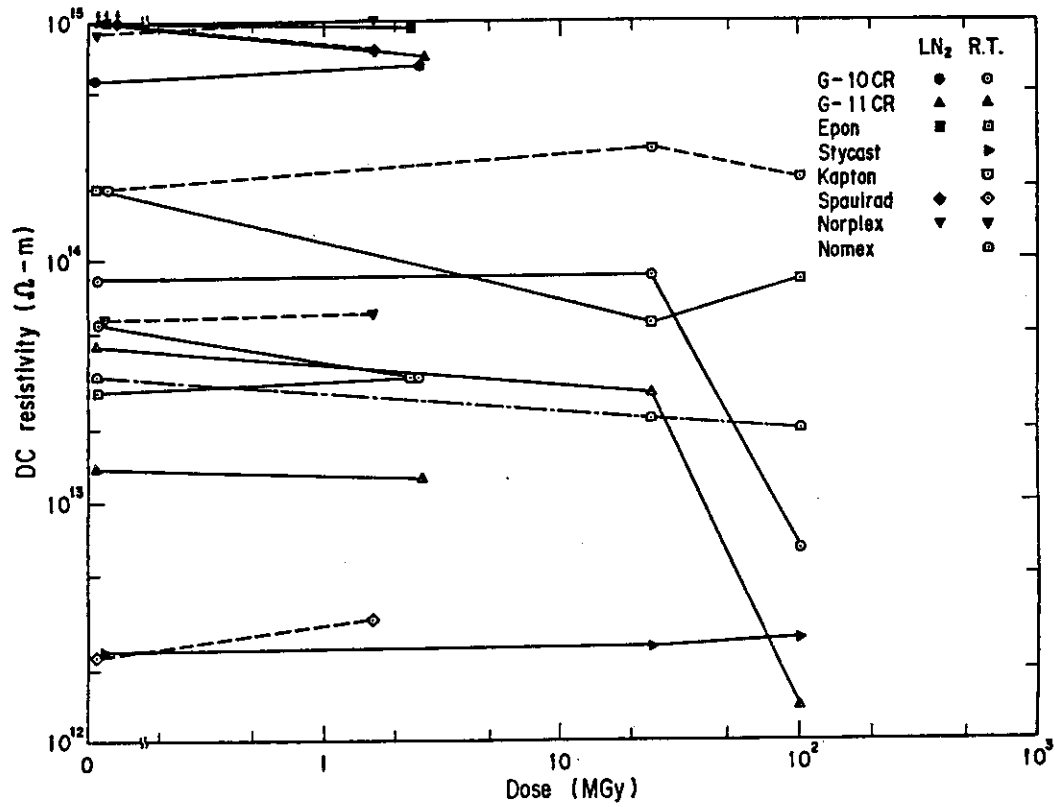


Fig. 4.1 DC resistivity of various insulators after irradiation at liquid helium temperature.

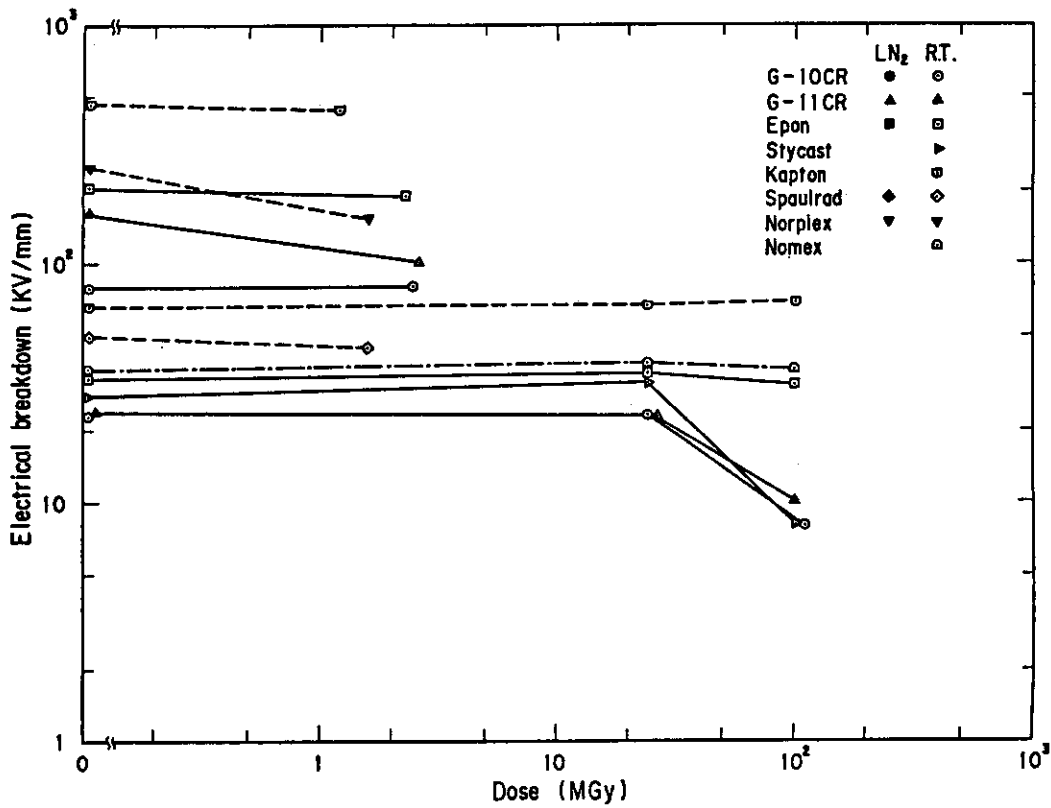


Fig. 4.2 Electrical breakdown of various insulators after irradiation at liquid helium temperature.

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### 7.1.3 Radiation Effects on Passive Shell Materials

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The elongated plasma has a vertical positional instability. In order to stabilize it active and passive coils are required. Active coils are provided out of blankets or shields to control the vertical instability. However it can not cope with rapid movements of plasma. Consequently, it is necessary to place conducting passive shell as near plasma as possible. Since electric resistance of shell determines electric time constant of shell, when shape is fixed the smaller resistance is preferable for the shell material.

Candidate passive shell materials are copper, aluminum, beryllium, and lead. The resistivity of copper and aluminum are small. Other candidates which have small resistivity are silver and gold. But they are too expensive. The resistivities of beryllium and lead are larger than those of copper and aluminum, but beryllium and lead can serve as neutron multipliers as well, and then the larger allowable thickness for shell can compensate to some extent for their larger resistivities.

We will report the materials' data such as electrical resistivities, ductilities, swelling, neutron absorption. And we will evaluate their applicabilities for the passive shell.

The operating temperatures expected for these materials are assumed to be 20 through 200°C. Copper data are for OFHC grade copper. Aluminum data are for 1100 aluminum.

## 1. Copper

### 1.1 Electrical resistivity

The resistivities of copper and copper alloys are shown in Fig. 1.1.1 [1] and Fig. 1.1.2 [2]. The temperature coefficients of electrical resistivities for the copper alloys are little different from that of OFHC copper as shown in Fig. 1.1.1. The resistivity of the alloy doped 1 atom % Co at room temperature is sensitive to the heat treatments [2].

According to Matthiessen's rule the electrical resistivity  $\rho(T)$  of metal is expressed at any temperature as follows:

$$\rho(T) = \rho_0 + \rho_{ph}(T) \quad , \quad (1.1.1)$$

where  $\rho_0$  is the residual resistivity, and  $\rho_{ph}(T)$  is the resistivity at temperature  $T$  due to electron scattering by thermal vibration of lattices. The residual resistivity  $\rho_0$  can be divided further as follows:

$$\rho_0 = \rho_D + \rho_{imp} \quad , \quad (1.1.2)$$

where  $\rho_D$  is the resistivity due to the defects (e.g. point defects, dislocation, and voids), and  $\rho_{imp}$  is the resistivity due to the impurities. The electrical resistivity increases with defects and impurities by neutron irradiation. The impurity resistivity increases due to transmutations. ( $\rho_T \approx \Delta\rho_{imp}$ )

## 1.2 Ductility

Elongation of OFHC copper is shown in Fig. 1.2.1[1]. Large elongation can be obtained by thermal treatment.

Effect of neutron irradiation on room temperature elongation of copper is shown in Fig. 1.2.2[7,8,9,10]. Although the total elongations of copper irradiated at 60-100°C are very scattered, it can be seen that the elongation is reduced by irradiation. The total elongation falls rapidly at the fluence of  $10^{18}$ - $10^{20}$  n/cm<sup>2</sup>. However it seems to decrease slowly for higher fluences. The same tendency is usually seen for other materials that is face-centered cubic metal such as aluminum. The elongation of copper irradiated at 327° does not change by neutron irradiation.

The temperature dependence of the elongation for copper irradiated with  $7.9 \times 10^{19}$  n/cm<sup>2</sup> (epithermal) of fluence at temperature below 100°C is shown in Fig. 1.2.3[8]. Although the data are very scattered, there does not appear any change of the elongation for room temperature to 200°C.

It is difficult to estimate the elongation at the fluence of  $1.0 \times 10^{22}$  n/cm<sup>2</sup> ( $E > 1$  MeV). If the loss of elongation obeys the dashed curves, the elongation of 5% is expected at the fluence of 6 MWyr/m<sup>2</sup> ( $\approx 2 \sim 3 \times 10^{22}$  n/cm<sup>2</sup>).

## 1.3 Swelling

The dependence of the swelling of copper on neutron fluence is shown in Fig. 1.3.1[9,11,12,13,14]. The range of irradiation temperature is 250°C to 450°C. The maximum swelling for the fluence of  $10^{22}$  n/cm<sup>2</sup> ( $E < 1$  MeV) is estimated to be 20-30% extrapolating from this figure.

The temperature dependence of the swelling of copper is shown in Fig. 1.3.2[13,14]. As the fluence increases the peak of swelling appears to shift to the lower side. The swelling at 200°C is a half or a third of the maximum swelling, and there is practically no swelling at temperatures lower than 130°C for the fluence up to  $5 \times 10^{20}$  n/cm<sup>2</sup>. It is necessary to study the swelling at temperatures between 100°C and 200°C and the swelling for fluences above  $10^{21}$  n/cm<sup>2</sup>.

The swellings of copper doped some elements are shown in Table 1.3.1[3]. The irradiation fluences is  $5 \times 10^{20}$  n/cm<sup>2</sup> and the irradiation temperature is 250°C. These elements are effective for the reduction of the swelling of copper.

The swelling of copper on irradiation with 200 keV Cu<sup>+</sup> ion to 60 dpa is shown in Fig. 1.3.3[15]. The maximum swelling of copper alloys on the same irradiation is shown in Fig. 1.3.4[15]. The measurements show a strong dependence on the concentration of solute. Alloys of a 1 atom% solute show more swelling than pure copper, and alloys of 1 atom% show less swelling. Especially, alloy of 1 atom% Au or Be shows no void swelling.

The maximum of swelling in Fig. 1.3.3 occurs about 470°C. Damage rate in the ion irradiation is for more large than that in the neutron irradiation. It is thought that the higher damage rate in ion irradiation causes the higher temperature for maximum swelling and lower swelling rate in comparison with neutron irradiation.

The resistivity measured at 4.2 K after exposure at 60°C in the fission reactor is shown in Fig. 1.1.3 [3]. The increase rate of the defect resistivity  $\rho_D$  reduces as fluence increases. Contrastingly, the increase rate of the impurity resistivity is almost constant against fluence.

The annealing effects on the resistivity of copper at room temperature after neutron irradiation at 40 ~ 45°C is shown in Fig. 1.1.4 [4]. The significant recovery of irradiation-induced defects occurs at room temperature in a rather long period.

The short period annealing characteristics of electrical resistivity in copper irradiated at 60°C is shown in Fig. 1.1.5 [5]. The recovery of defects up to 200°C is not so large.

Neutrons produce Ni, Zn and Co in copper by transmutations. The electrical resistivity increases due to these impurities are shown in Fig. 1.1.6 [6]. The specific resistivities of these element addition by small amount are 125  $\mu\Omega\text{cm}/\text{unit fraction}$  for Ni, 53  $\mu\Omega\text{cm}/\text{unit fraction}$  for Zn and 630  $\mu\Omega\text{cm}/\text{unit fraction}$  for Co respectively. The resistivity increase due to impurities approximately follows the simple addition rule.

We will estimate the resistivity increase of copper irradiated by the neutron with spectrum at first wall. The resistivity increase due to neutron induced defects saturate 0.08  $\mu\Omega\text{cm}$  at room temperature. Amount of transmutations is approximately proportional to fluence. Ni, Zn, and Co produced by the neutron fluence of 3 MWyr/m<sup>2</sup> are 1.2 atom%, 0.723 atom%, and 0.022 atom% respectively. The resistivity increase due to transmutations is calculated to be 2.064  $\mu\Omega\text{cm}$  for a fluence of 3 MWyr/m<sup>2</sup>. The total resistivity increase due to neutron irradiation for 3 MWyr/m<sup>2</sup> is estimated at 2.14  $\mu\Omega\text{cm}$ . The resistivity change by neutron irradiation with first wall neutron spectrum is shown in Fig. 1.1.7. The resistivity of copper irradiated for a fluence of 3 MWyr/m<sup>2</sup> at 150°C is estimated to be 4.7  $\mu\Omega\text{cm}$  and that of 6 MWyr/m<sup>2</sup> at 150°C is estimated to be 7.0  $\mu\Omega\text{cm}$ .

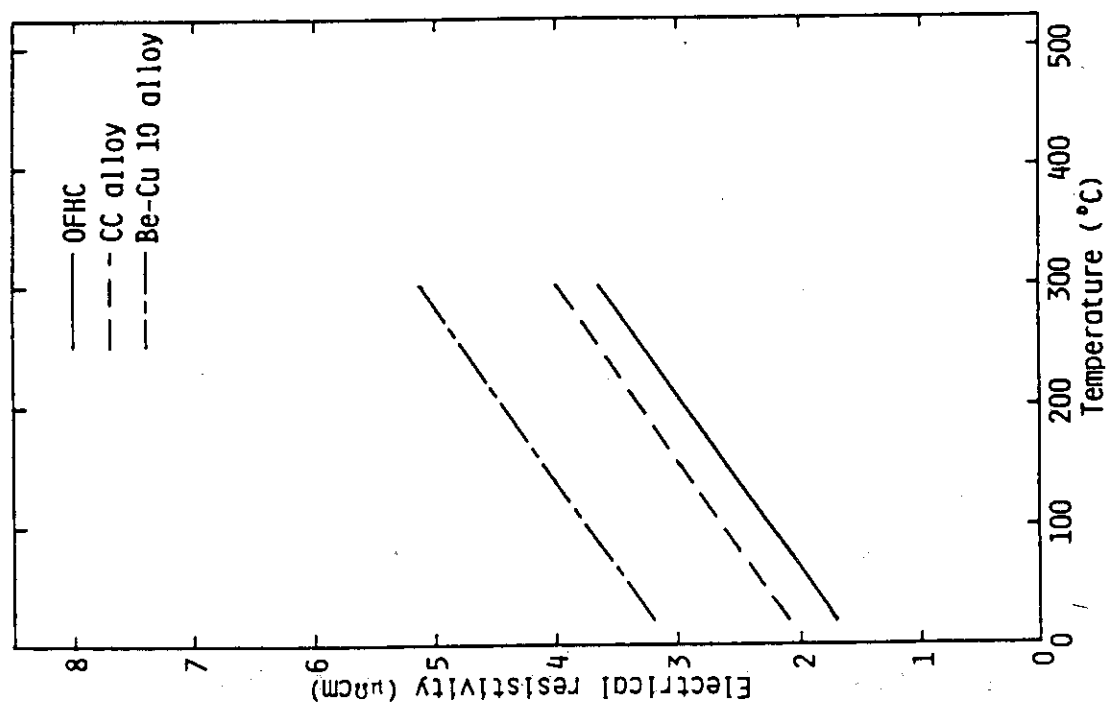


Fig. 1.1.1. Temperature Dependence of Electrical Resistivities of Copper and Copper Alloys.

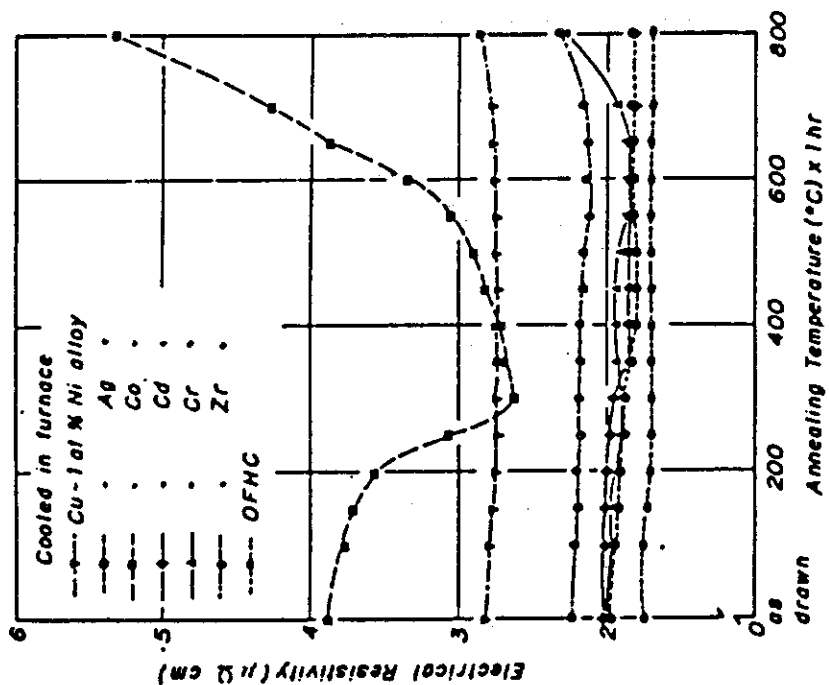


Fig. 1.1.2. Changes in Electrical Resistivity of the Copper Alloys Annealed for 1hr at Various Temperatures after Furnace Cooling and Drawing.

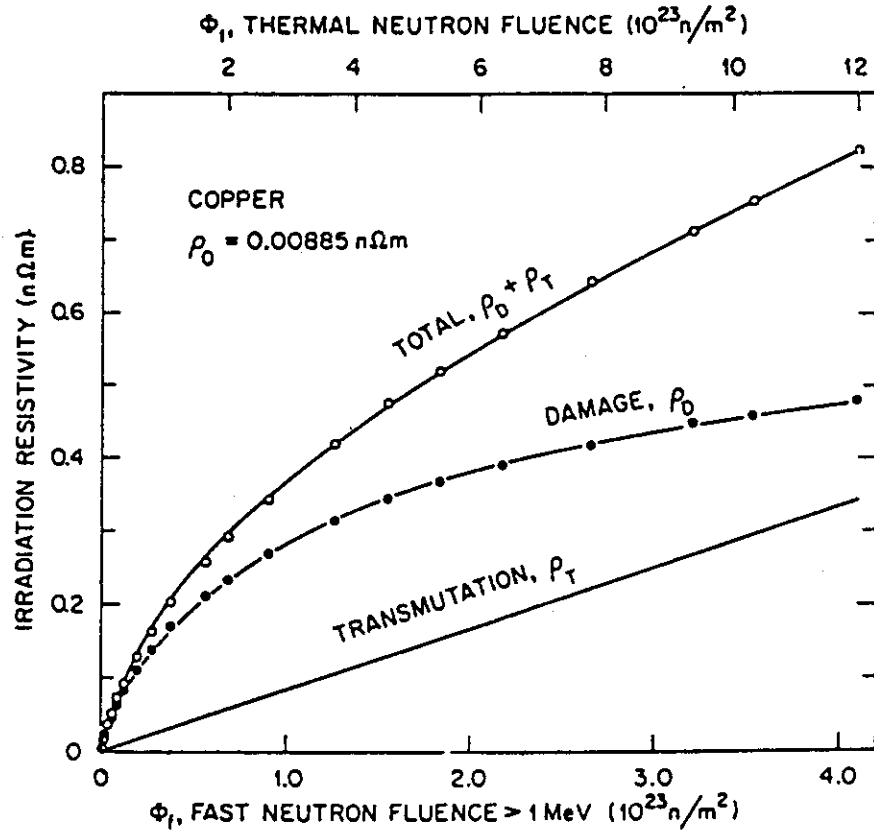


Fig.1.1.3. The total Residual-Resistivity Increase versus Thermal and Fast-Neutron Fluence for Sample Pair No.1. The Resolved Contributions due to Damage and Transmutation Resistivity are Shown for Comparison.

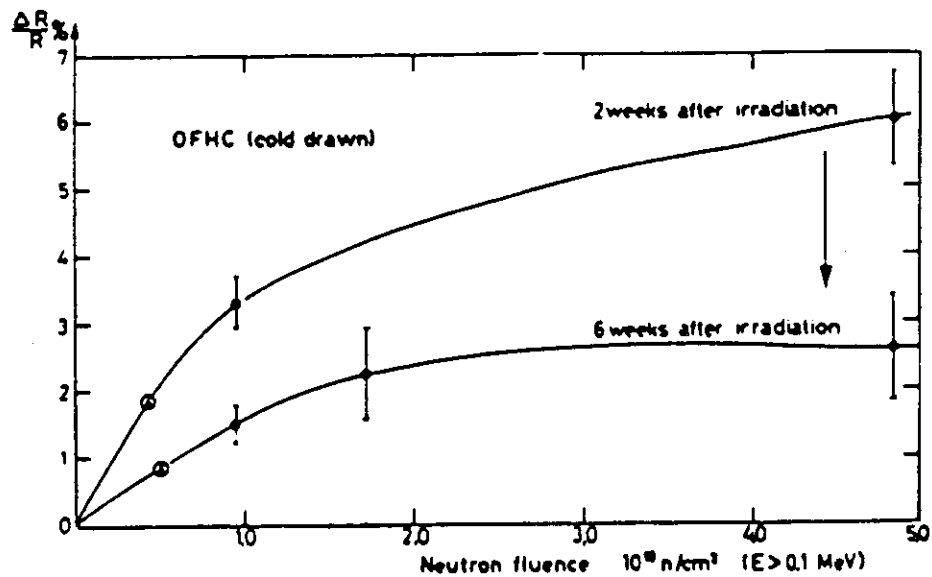


Fig.1.1.4. Annealing Effects in Irradiated OFHC Cold-Drawn Copper Wire at Room Temperature.

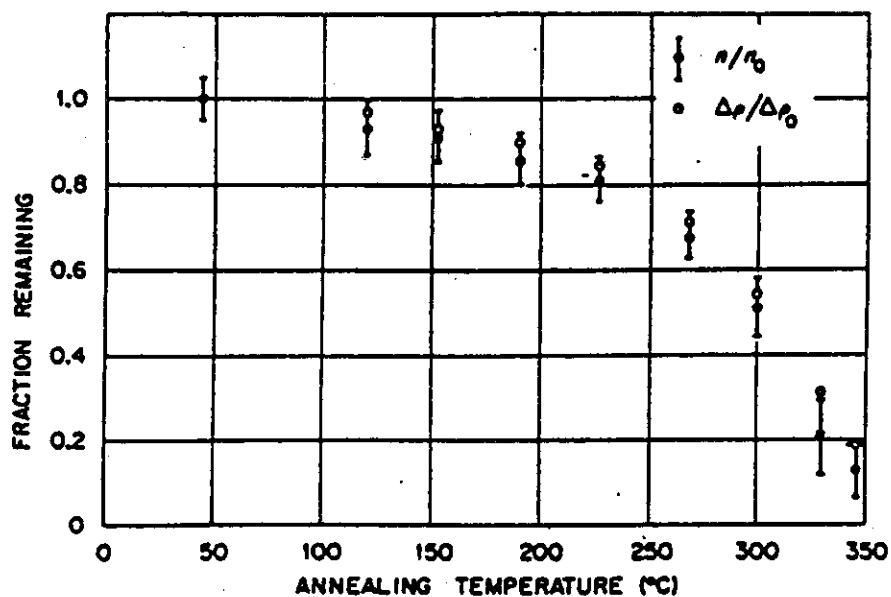


Fig.1.1.5. Annealing characteristics of the number of point defects in loops as determined by integral diffuse scattering and the annealing characteristics of electrical resistivity in ambient-temperature neutron-irradiated copper.

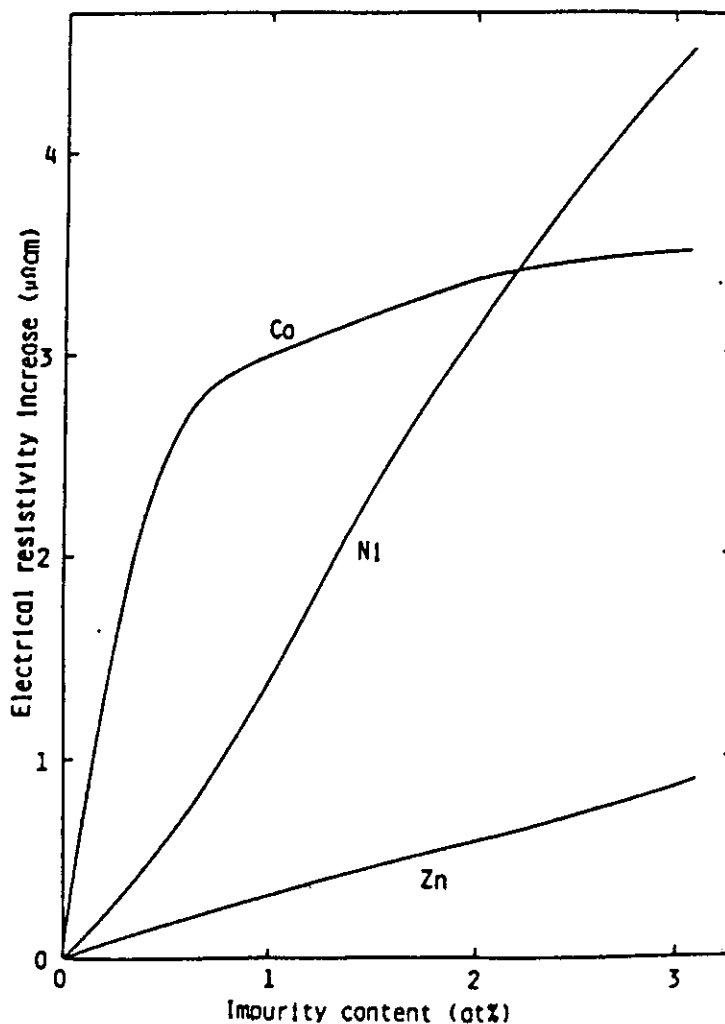


Fig.1.1.6. Electrical Resistivity Increase of Copper by Impurities.

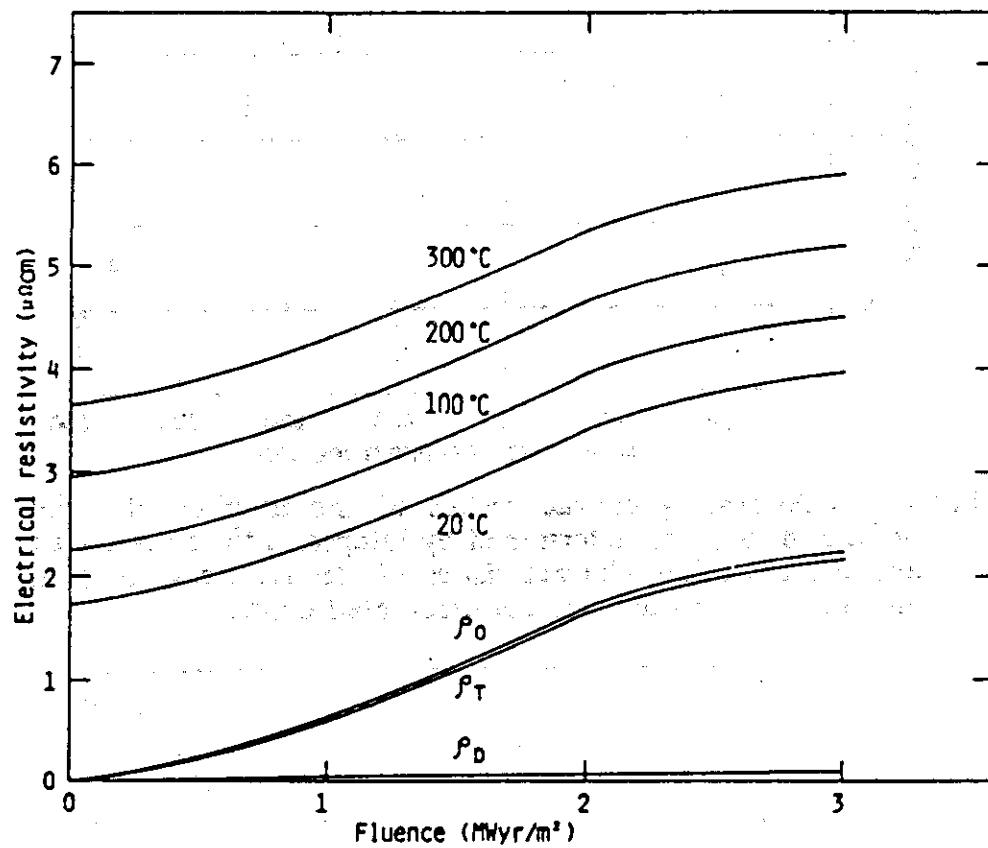


Fig.1.1.7 Effect of Neutron Irradiation on Electrical Resistivity of Copper.



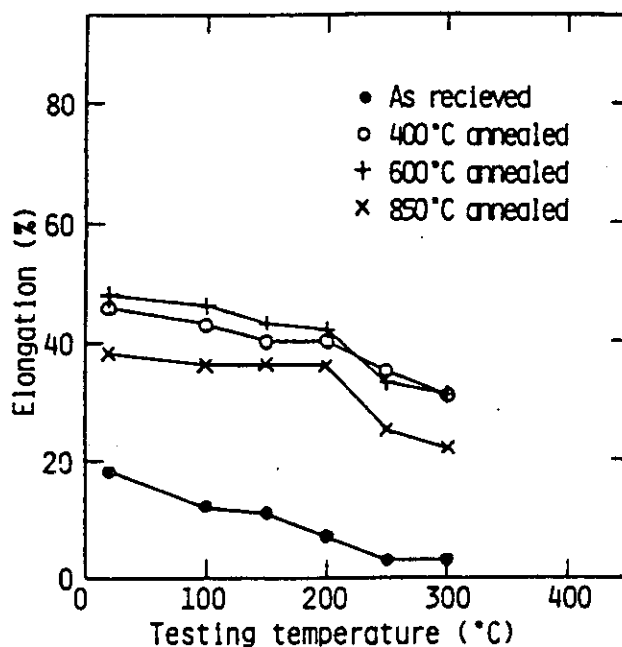


Fig.1.2.1. Effect of Heat Treatment on Elongation of OFHC Copper.

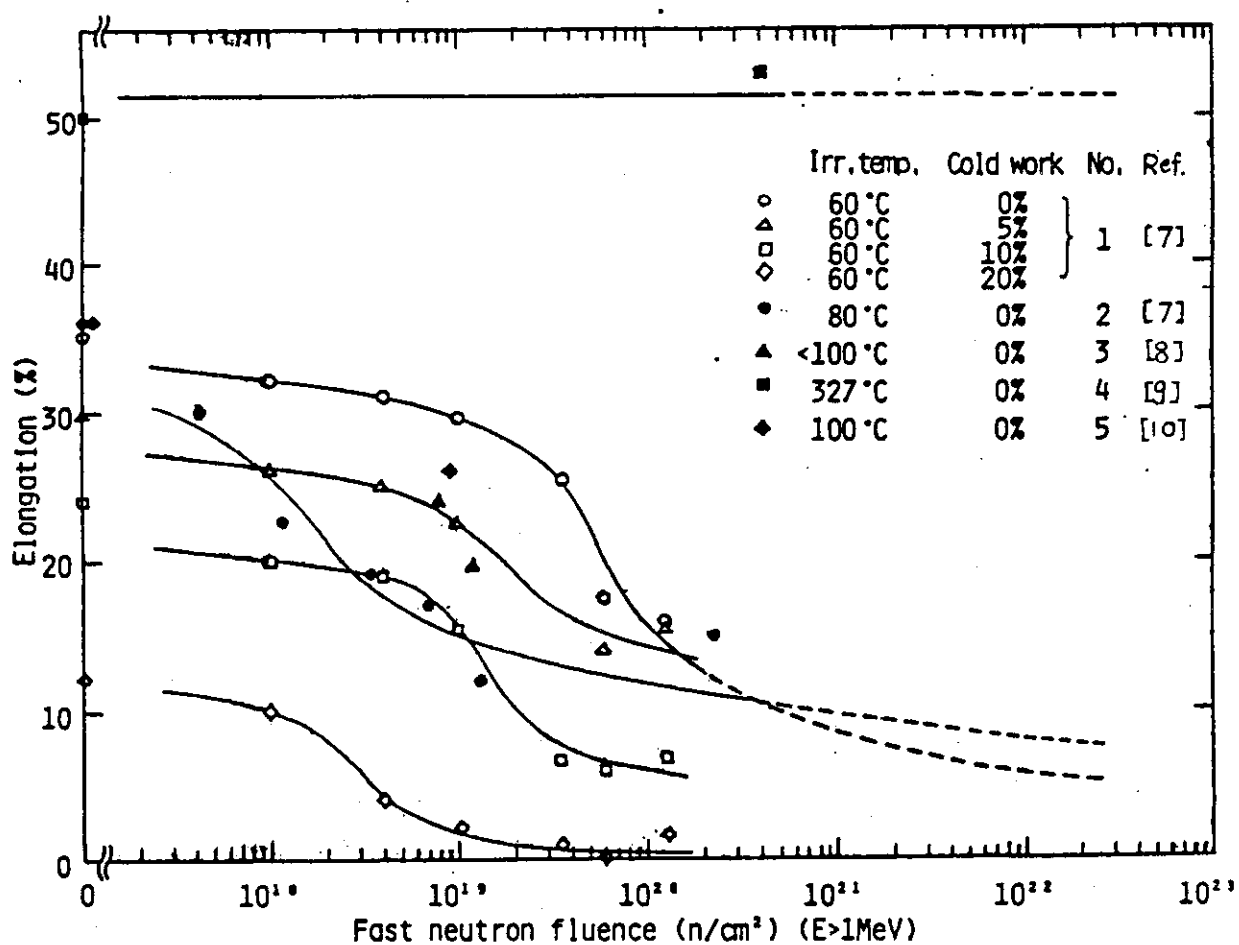


Fig.1.2.2. Effect of Neutron Irradiation on Room Temperature Elongation of Copper.

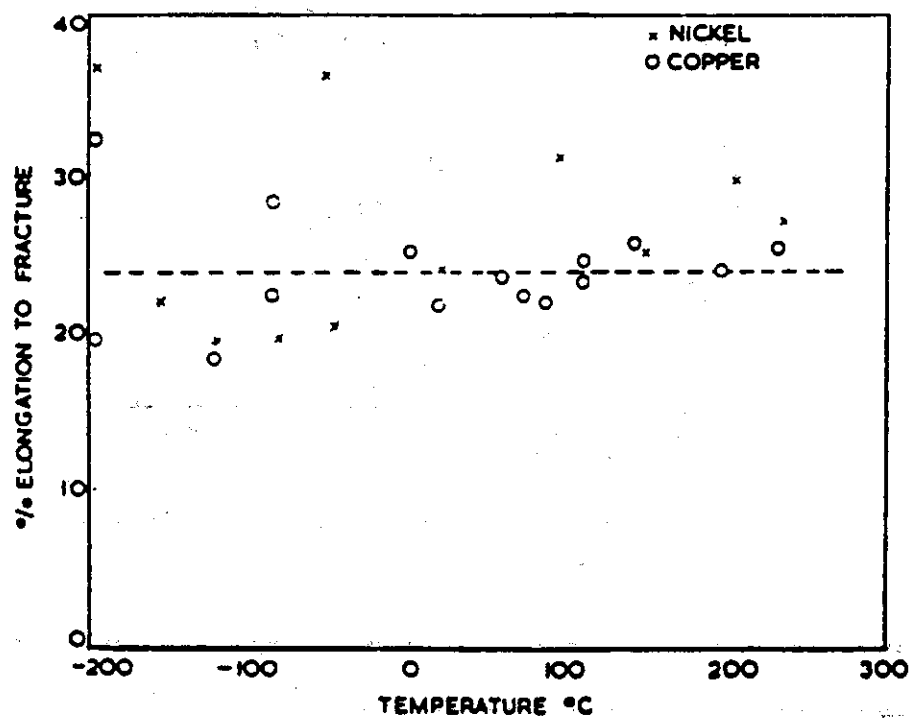


Fig.1.2.3. Elongation to fracture of irradiated copper and nickel.

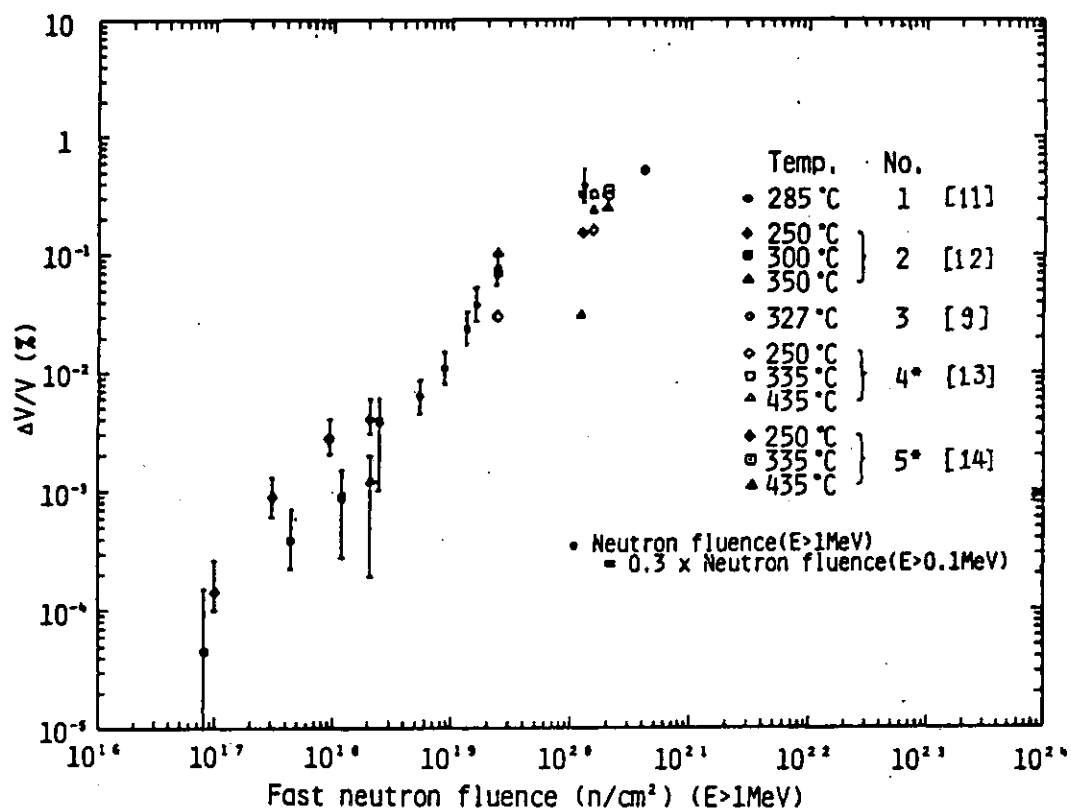


Fig.1.3.1. Neutron Irradiation Induced Swelling in Copper.

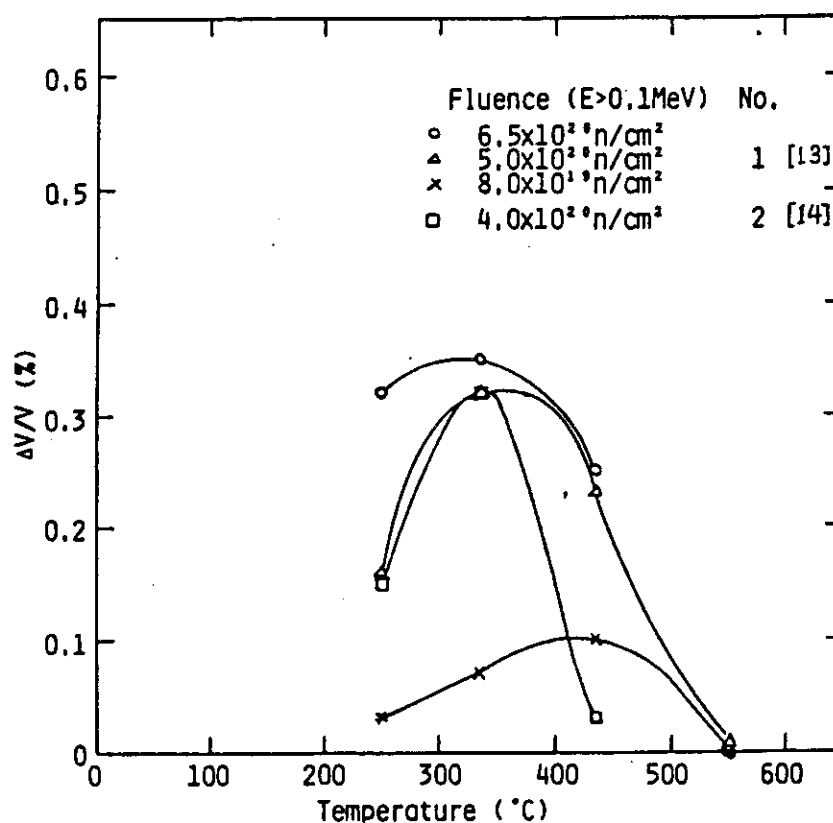


Fig.1.3.2. Temperature Dependence of Neutron Irradiation Induced Swelling of Copper.

Table 1.3.1.

Valeurs du gonflement de la densité moyenne et du diamètre moyen des cavités pour les alliages de cuivre d'énergie de faute différente (irradiation L 15- $T=250^{\circ}\text{C}$ )

Alliage	$\gamma_f$ (erg/cm <sup>2</sup> )	$\Delta V/V$ (%)	$\bar{n}$ (cav/cm <sup>3</sup> )	$\bar{d}$ (Å)
**Cu	55	0,16	$2,2 \times 10^{14}$	275
Cu-Ge 1%	50	0,10	$2,5 \times 10^{14}$	480
Cu-Ge 3%	$\sim 45$	$\sim 0$	quelques cavités	600
Cu-Si 1%	$\sim 45$	$\sim 0$	quelques cavités	120
Cu-Al 1%	$\sim 45$	$\sim 0$	quelques cavités	-
Cu-Al 3%	$\sim 30$	0	quelques cavités	-
Cu-Al 5%	$\sim 5$	0	0	-

\* Les valeurs des énergies de faute d'empilement ont été tirées de ref. <sup>10</sup>).

\*\* L'échantillon témoin de cuivre pur irradié dans L 15 n'ayant pu être observé on a donné ici les valeurs de l'irradiation MP 1, effectuée dans les mêmes conditions nominales.

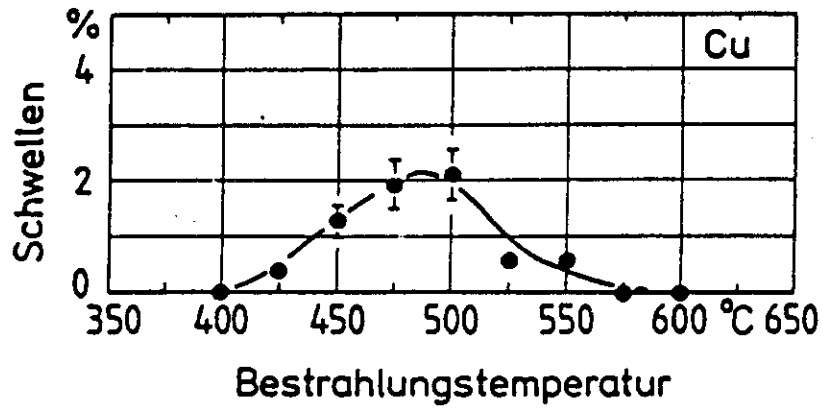


Fig.1.3.3. Schwellen von reinem Kupfer

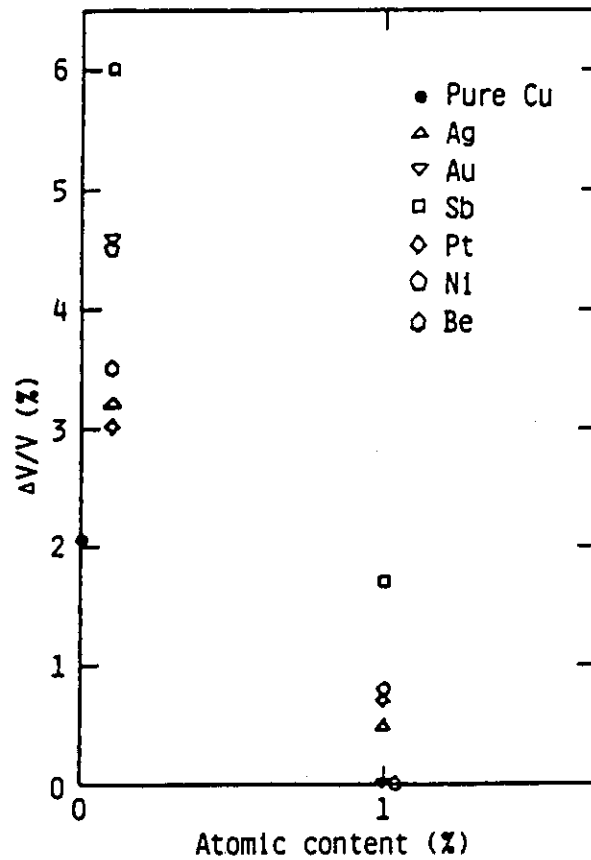


Fig.1.3.4. Effect of Solute on Swelling of Copper.

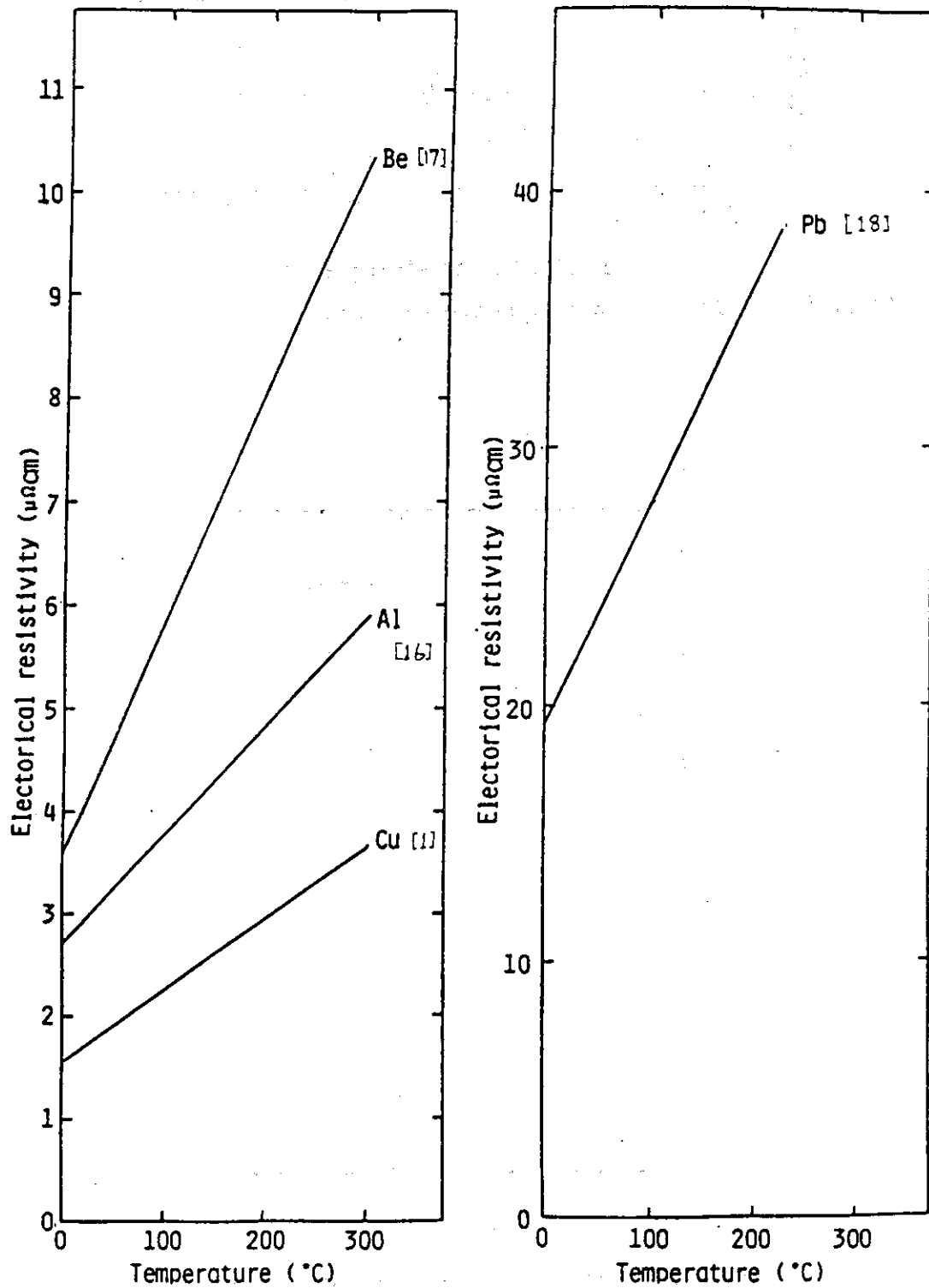


Fig.1.2.1. Temperature Dependence of Electrical Resistivities of Candedated Passive Shell Materials.

## 2. Aluminum

### 2.1 Electrical resistivity

The temperature dependence of electrical resistivity for aluminum in comparison with those for copper, beryllium and lead is shown in Fig. 2.1.1 [1,16,17,18].

The isochronal recovery of aluminum after the neutron irradiation of  $2 \times 10^{18}$  n/cm<sup>2</sup> ( $E > 0.1$  MeV) at 4.5 K is shown in Fig. 2.1.2[19]. The irradiation-induced resistivity of aluminum is recovered completely at 260 K. It can be expected that all of defect resistivity induced by neutron irradiation at temperatures above room temperature is recovered.

Neutron produces Mg, Si and Na in aluminum by transmutations. Amount of Na is very little. The electrical resistivity increase due to the impurities of Mg and Si is shown in Fig. 2.1.3[20].

The resistivity increase of aluminum irradiated by the neutron spectrum at first wall is estimated. Concentration of Mg, Si and Na transmuted by 3 MWyr/m<sup>2</sup> neutron irradiation are 0.214 atom%, 0.0295 atom% and 0.0001 atom%, respectively. The resistivity increase due to transmutation is calculated to be 0.2  $\mu\Omega$ cm for 3 MWyr/m<sup>2</sup> neutron irradiation. The resistivity increase due to irradiation-induced defect is zero. The resistivity change by neutron irradiation with the first wall neutron spectrum is shown in Fig. 2.1.4.

### 2.2 Ductility

The effect of neutron irradiation on the temperature dependence of elongation of 1100 aluminum is shown in Fig. 2.2.1[21,22]. The elongation of aluminum irradiated to  $1 \times 10^{22}$  n/cm<sup>2</sup> ( $E > 1$  MeV) exhibits a minimum at about 250°C.

The effect of neutron irradiation on the elongation of 1100 aluminum is shown in Fig. 2.2.2[21,22]. The loss of elongation at 200°C is larger than that at room temperature. The significant loss of elongation occurs under the fluence of  $10^{22}$  n/cm<sup>2</sup> and then the loss saturates.

### 2.3 Swelling

The swellings of aluminum and aluminum alloys are shown in Fig. 2.3.1 [22,23]. The swelling of 1100 aluminum at the fluence of  $3 \times 10^{22}$  n/cm<sup>2</sup> ( $E > 0.1$  MeV) is about 1%~5%.

The effect of isochronal annealing on swelling of 1100 aluminum irradiated to about  $3 \times 10^{22}$  n/cm<sup>2</sup> is shown in Fig. 2.3.2[24]. The swelling exhibits a minimum at about 300°C.

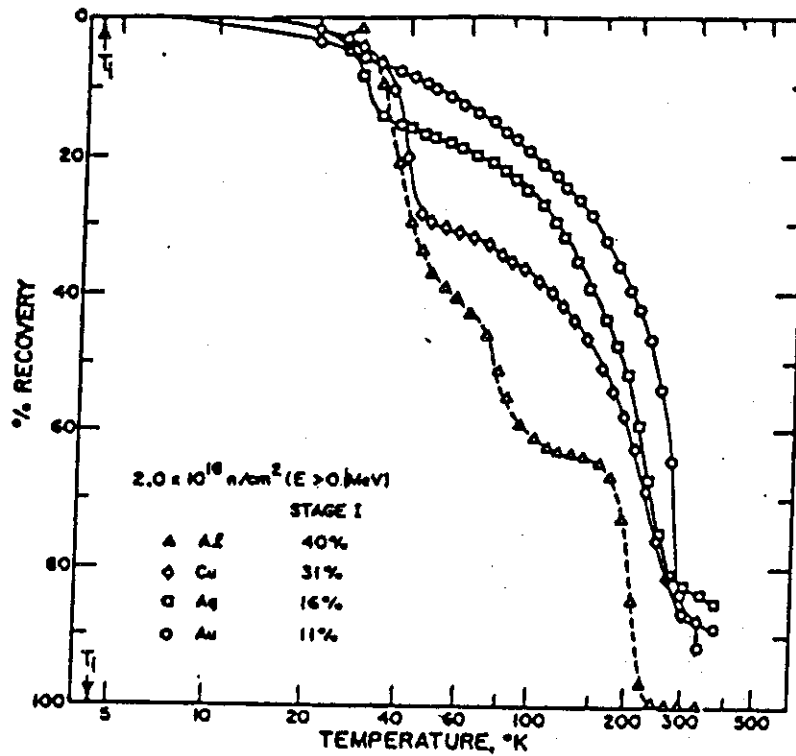


Fig.2.1.2. Recovery versus logarithm of absolute temperature showing decreased stage I and entire recovery spectra with increasing atomic mass for fcc metals.

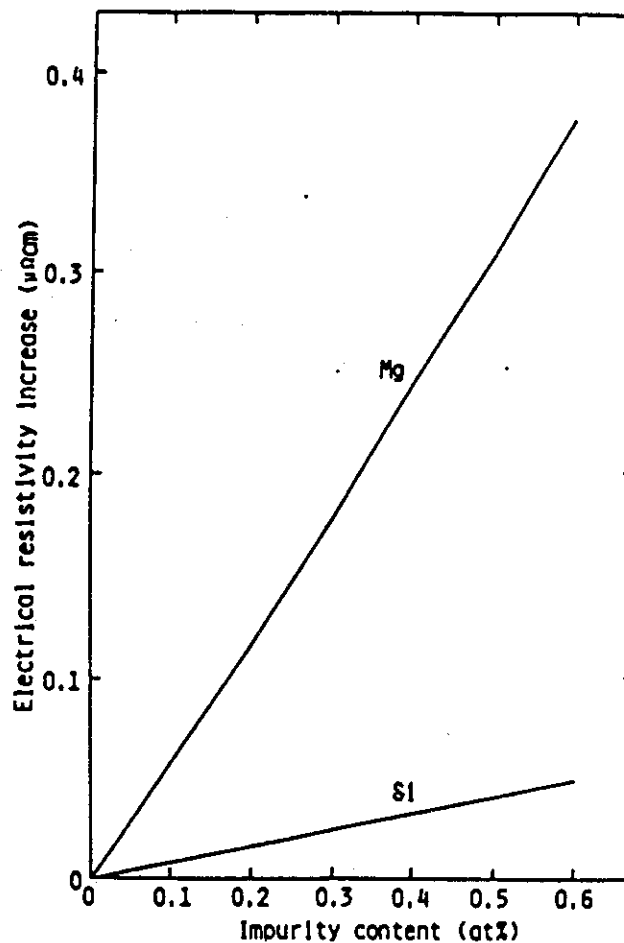


Fig.2.1.3. Electrical Resistivity Increase of Aluminum by Impurities.



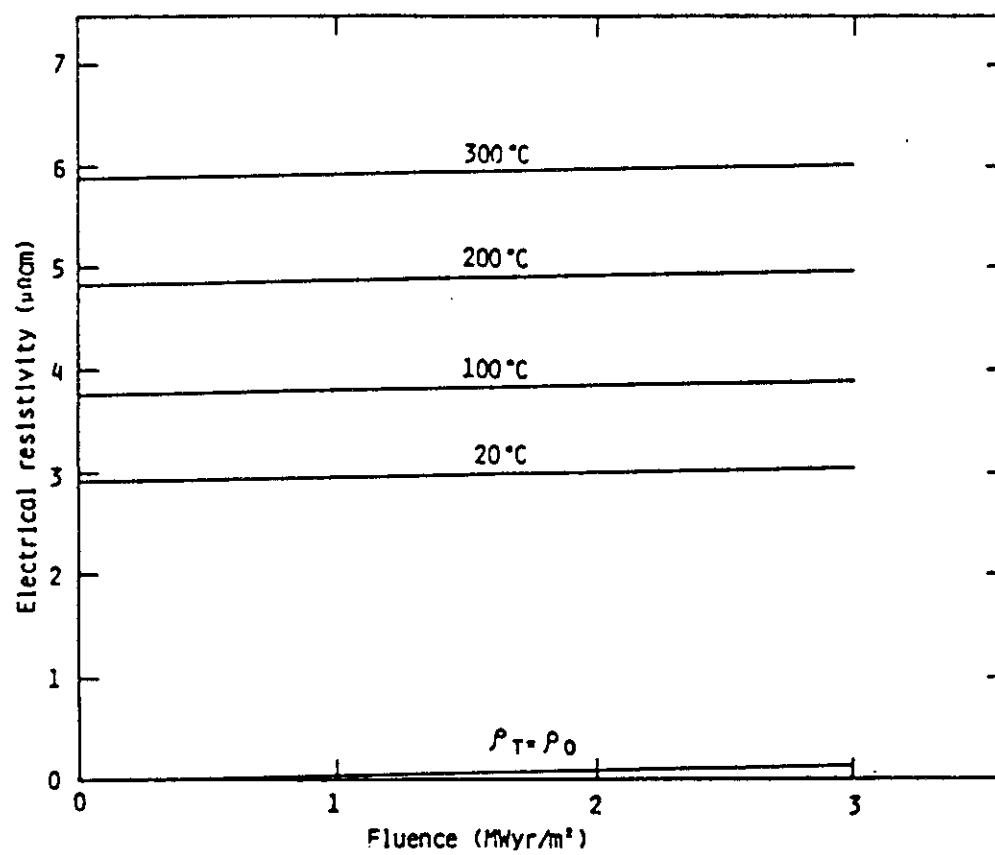


Fig.2.1.4. Effect of Neutron Irradiation on Electrical Resistivity of Aluminum.

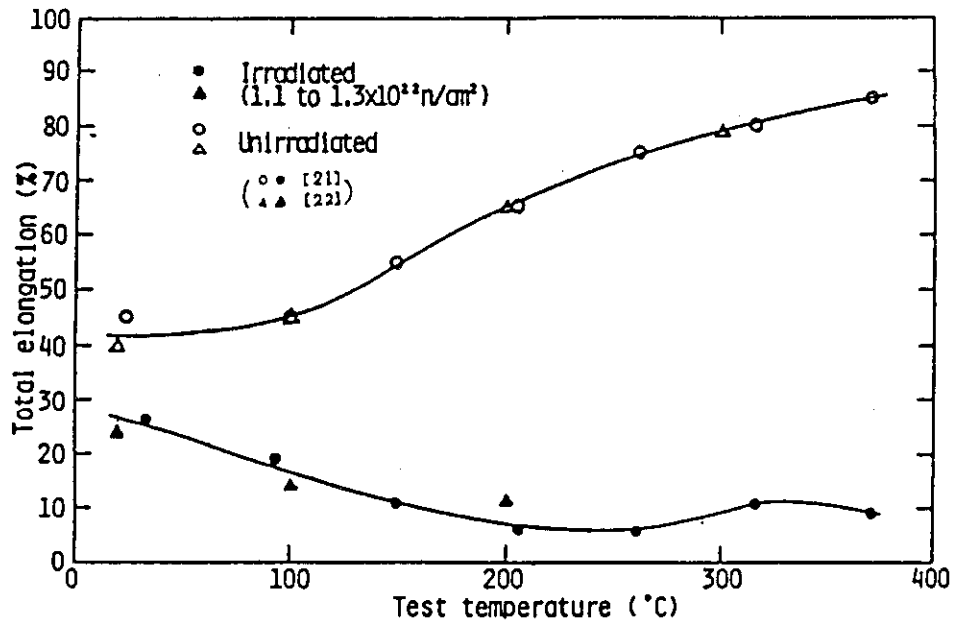


Fig.2.2.1. Effect of Neutron Irradiation on Temperature Dependence of Elongation of 1100 Aluminum.

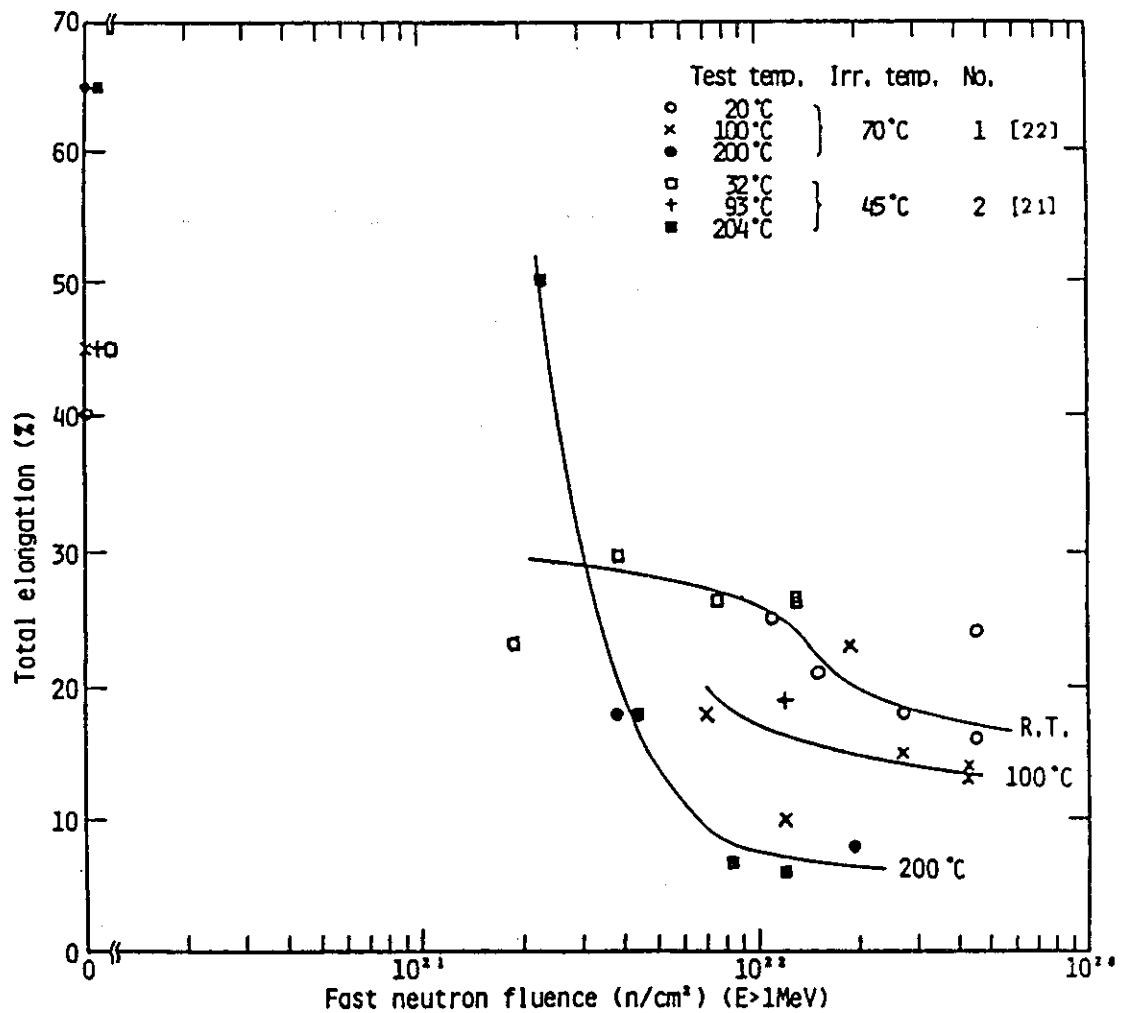


Fig.2.2.2. Effect of Neutron Irradiation on Elongation of 1100 Aluminum.

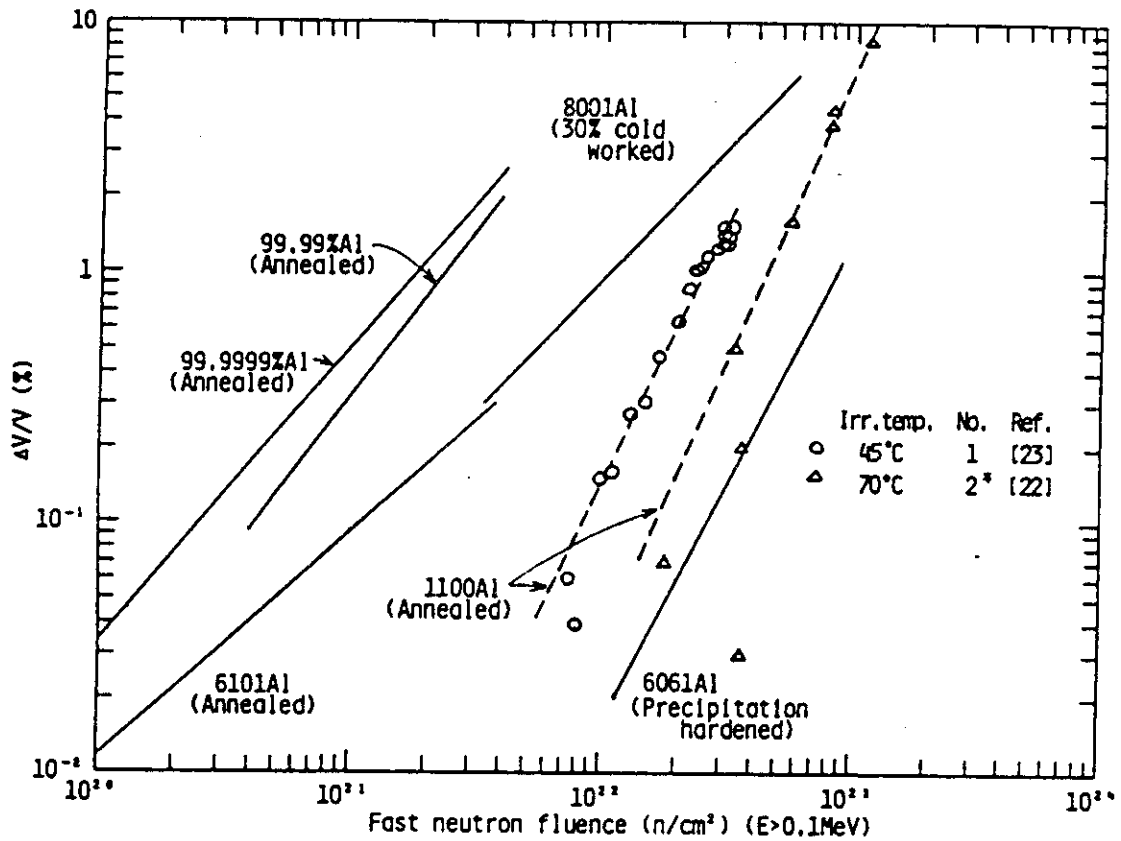


Fig.2.3.1 Irradiation-Induced Swelling in Aluminum and its Alloys.  
 \* Neutron fluence( $E>0.1MeV$ ) = 2 x Neutron fluence( $E>1MeV$ )

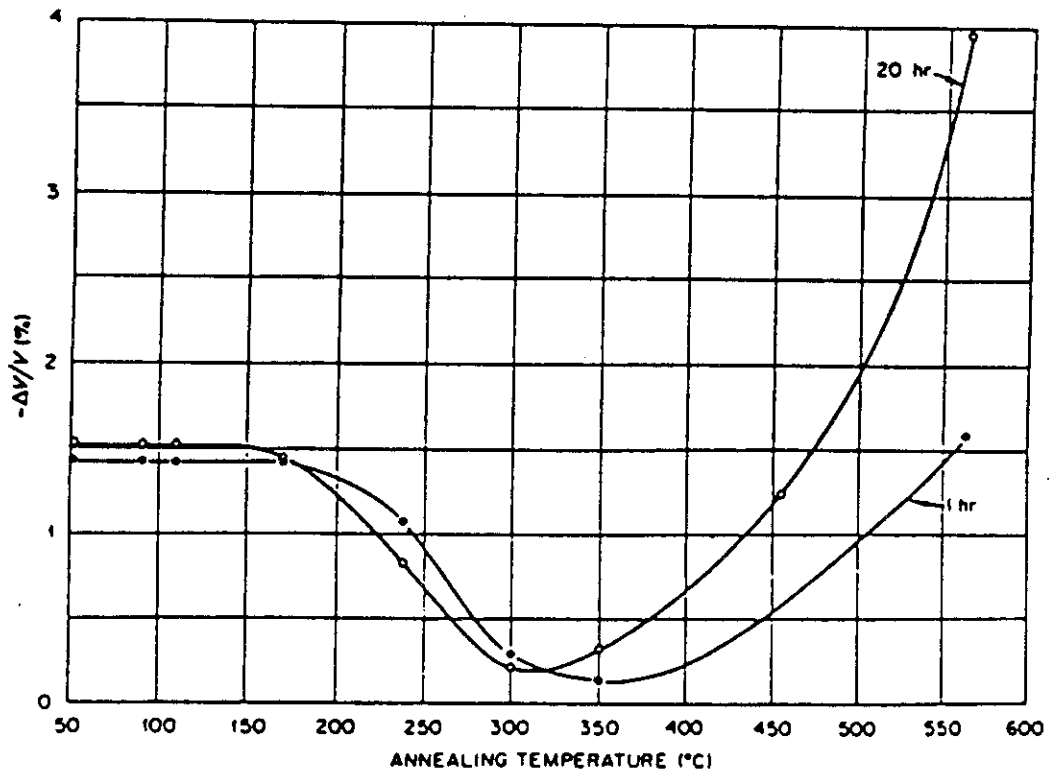


Fig.2.3.2. Effect of isochronal annealing on swelling of 1100 aluminum  
 irradiated to about  $3 \times 10^{22} n/cm^2$ .

### 3. Beryllium

#### 3.1 Electrical Resistivity

The electrical resistivity of beryllium is shown in Fig. 2.1.2[17].

The resistivity increase of beryllium irradiated at 90 K is shown in Fig. 3.1.1[25]. The isochronal recovery of resistivity of beryllium irradiated at 90 K to the fluence of  $1.05 \times 10^{19}$  n/cm<sup>2</sup> ( $E > 1.0$  MeV) is shown in Fig. 3.1.2[25]. Almost all of the irradiation-induced resistivity is recovered at 330 K. But it should be noted the fluence of these data is very low.

#### 3.2 Ductility

The elongation of beryllium is influenced by manufacturing process, distribution of impurities, and grain size. Beryllium has close packed hexagonal structure and the elongation of beryllium sheet exhibits anisotropy.

The effect of working on the elongation of beryllium is shown in Fig. 3.2.1[26]. The elongations of various form materials are shown in Table 3.2.1[17]. The maximum elongation of block material is about 4%. One of the sheet-type beryllium exhibits the elongation of 16% at room temperature. The temperature dependence of the elongation of hot-pressed beryllium is shown in Fig. 3.2.2[27]. The elongation of hot-pressed beryllium exhibits a minimum at room temperature. The elongation of other type beryllium exhibits a minimum between room temperature and 200°C.

The effect of irradiation on the elongations of sheet-type and block-type beryllium is shown in Fig. 3.2.3[28,29,30,31]. Neutron irradiation between room temperature to 350°C causes the striking loss of the longitudinal elongation of sheet-type beryllium. The longitudinal elongations of sheet-type beryllium irradiated to the fluence higher than  $10^{21}$  n/cm<sup>2</sup> ( $E > 1$  MeV) are estimated to be zero. The transverse elongation of unirradiated sheet-type beryllium is only about 0.2% at room temperature, and irradiation almost does not change it. The elongation of block-type beryllium irradiated to the fluence higher than  $3 \times 10^{21}$  n/cm<sup>2</sup> ( $E > 1$  MeV) is estimated to be almost zero.

Beryllium has been used as a reflector in some fission reactors, its life time fluence is about  $10^{22}$  n/cm<sup>2</sup> ( $E > 1$  MeV) and is limited by thermal and swelling stress.

#### 3.3 Swelling

The swelling of beryllium irradiated at room temperature and up to about 120°C is shown in Fig. 3.3.1[29,32,33,34,35].

The effect of postirradiation annealing on swelling of beryllium irradiated at 70°C to a fluence of about  $7.6 \times 10^{21}$  n/cm<sup>2</sup> ( $E > 1$  MeV) is shown in Fig. 3.3.2[35]. The large swelling occurs at higher temperature than 700°C.

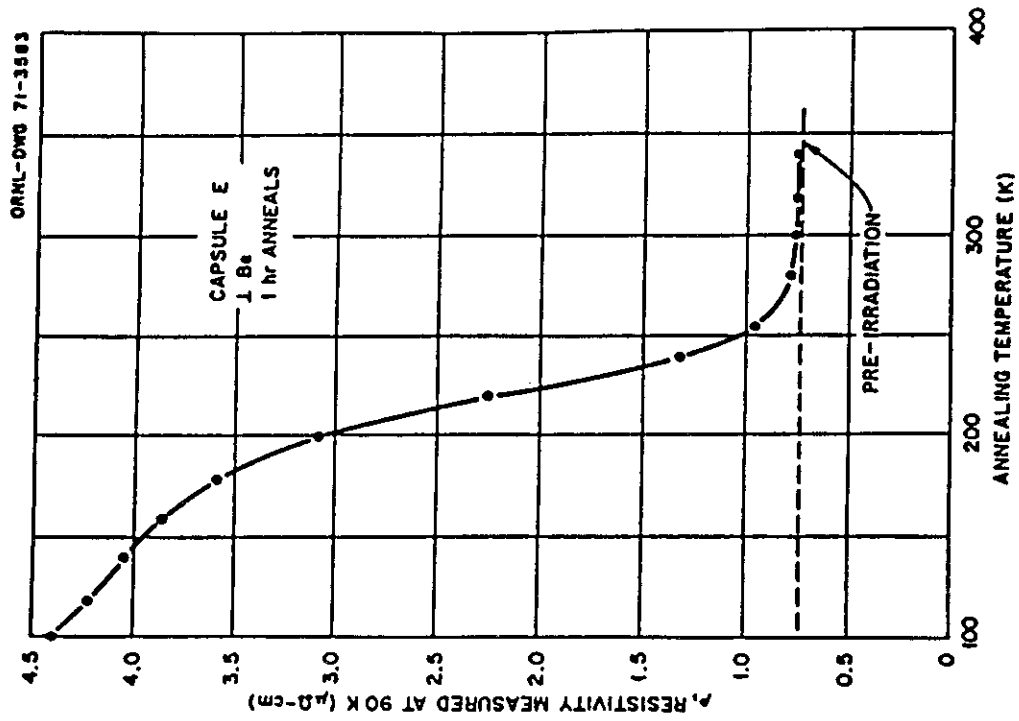


Fig.3.1.2. Isochronal recovery of electrical resistivity.

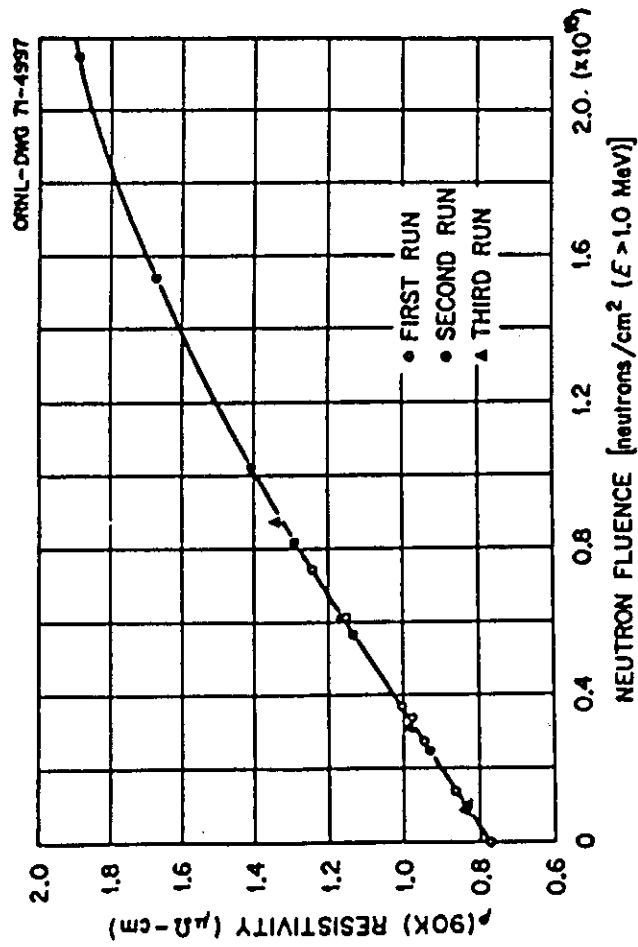


Fig.3.1.1. Electrical resistivity at 90K versus fluence.

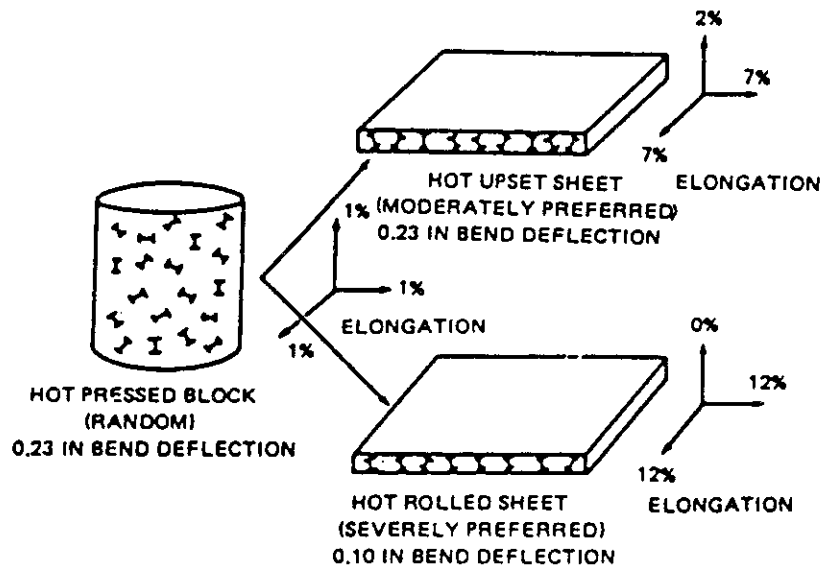


Fig.3.2.1. Effect of working and degree of preferred orientation on tensile ductility of beryllium hot-pressed block and sheet.

Table 3.2.1. Typical tensile properties of commercially available forms.

Material	Test direction	Ultimate tensile strength, $F_u$ [ksi (N/mm <sup>2</sup> )]	Yield strength 0.2% offset, $F_y$ [ksi (N/mm <sup>2</sup> )]	Elongation, $\epsilon$ (%) <sup>a</sup>
<b>Block</b>				
normal-purity (hot-pressed) <sup>a</sup> structural grade	L <sup>c</sup>	53(370)	38(266)	2.3
	T	56(390)	39(273)	3.6
thermal or brake grade	L <sup>c</sup>	42(294)	28(196)	2.7
	T	46(322)	28(196)	4.6
high-purity (isopressed)	L	65(455)	41(287)	3.9
	T	65(455)	41(287)	4.4
high-oxide instrument grade (hot pressed)	L	68(476)	58(406)	1.5
	T	73(511)	59(413)	2.7
fine grain size (isopressed)	L	84(580)	59(407)	3.7
	T	85(587)	59(407)	4.2
<b>Sheet (0.040–0.250 in. thick)</b>				
normal-purity powder		77(531)	54(372)	16
normal-purity ingot <sup>(m)</sup>		51(352)	25(172)	7
<b>Extrusions</b>				
normal-purity powder		95–100(655–690)	50–75(345–518)	8–13
high-purity powder		95–120(655–828)	50–75(345–518)	8–13
<b>Forgings</b>				
normal-purity <sup>a,711</sup>		70–87(483–600)	63–87(435–600)	0–4.5
<b>Wire (0.002–0.025 in. diameter)</b>				
high-purity ingot <sup>(m)</sup>		140(966)	115(793)	3

<sup>a</sup>Elongation in 2 in., except 10 in. for wire.

<sup>b</sup>Structural grade contains about 1.8% BeO and thermal or brake grade about 0.9%.

<sup>c</sup>L and T refer to longitudinal and transverse test directions.

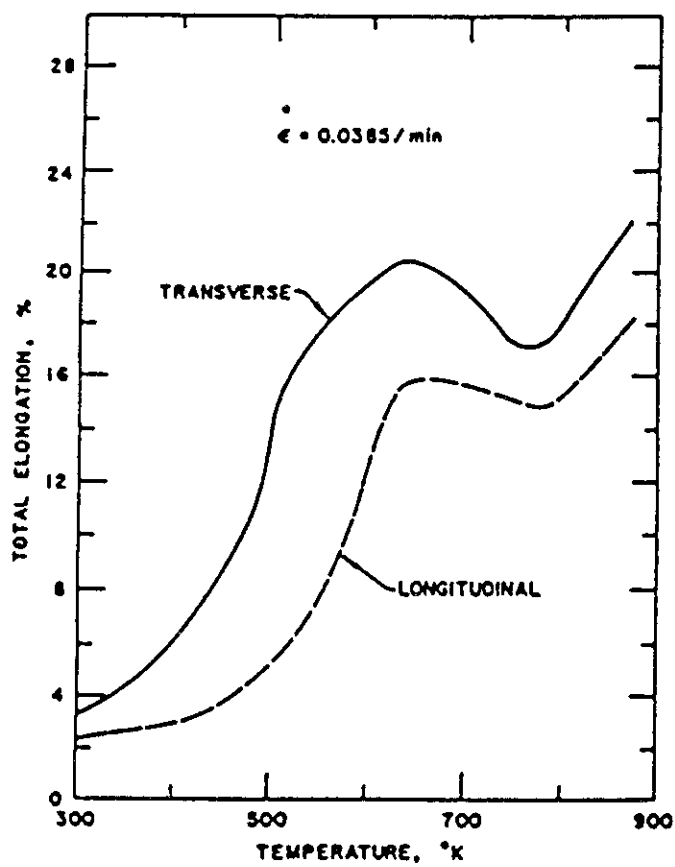


Fig.3.2.2. Temperature dependence of total elongation of HP21 beryllium.

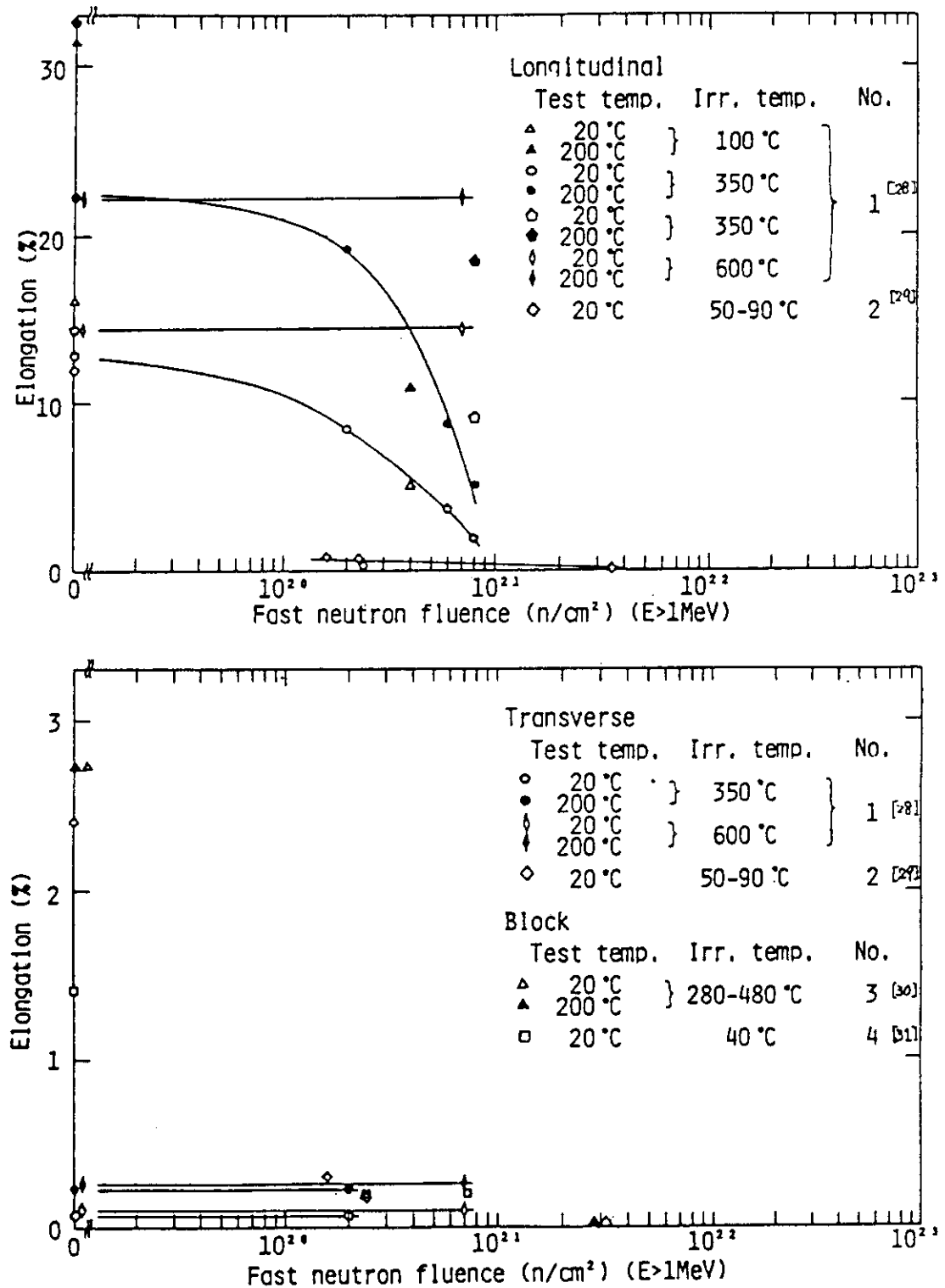


Fig.3.2.3. Effect of Neutron Irradiation on Elongation of Beryllium.



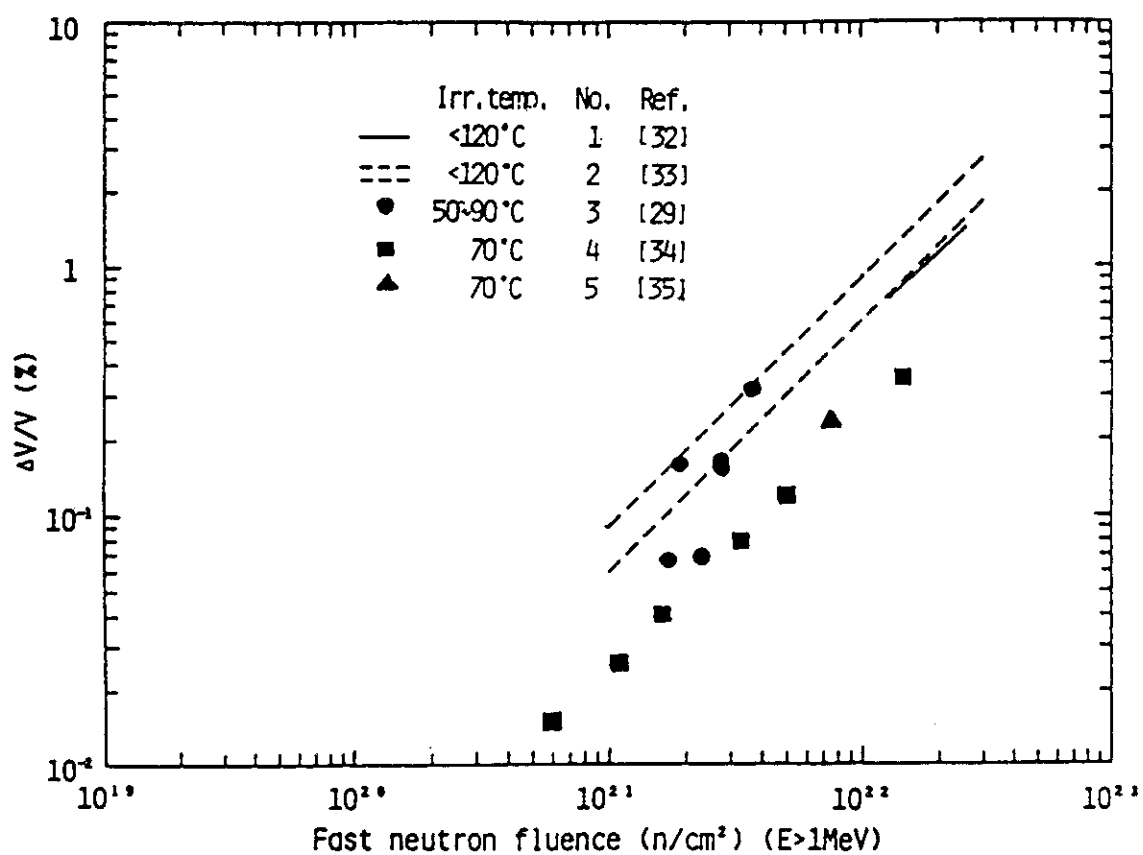


Fig.3.3.1. Neutron Irradiation Induced Swelling of Beryllium.

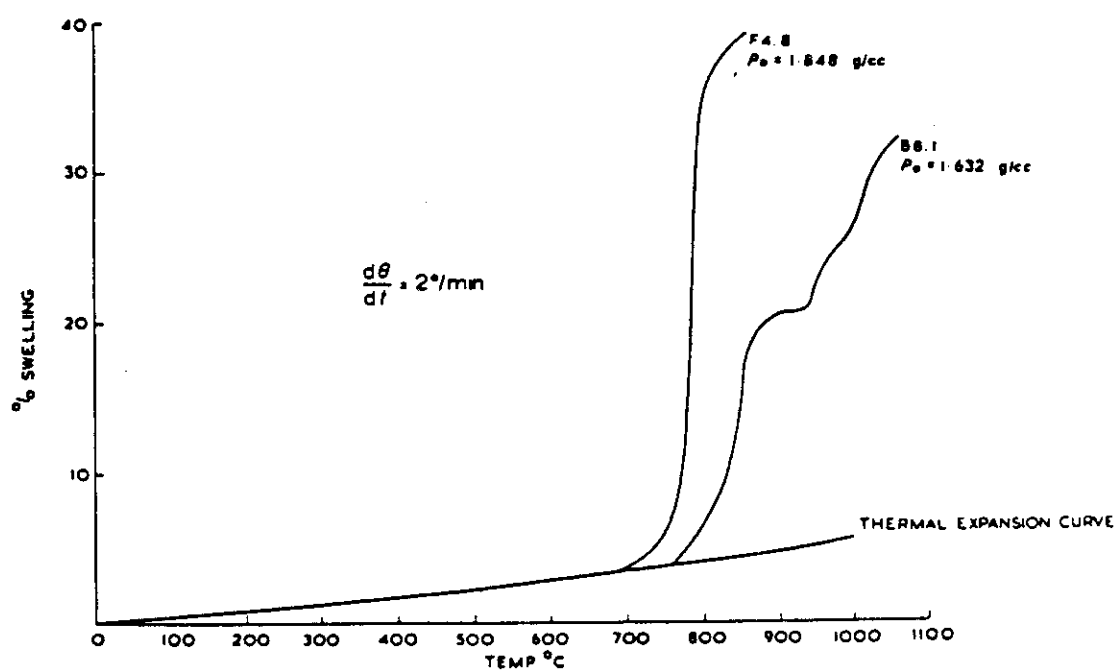


Fig.3.3.2. Linear temperature rise dilatometer curves.

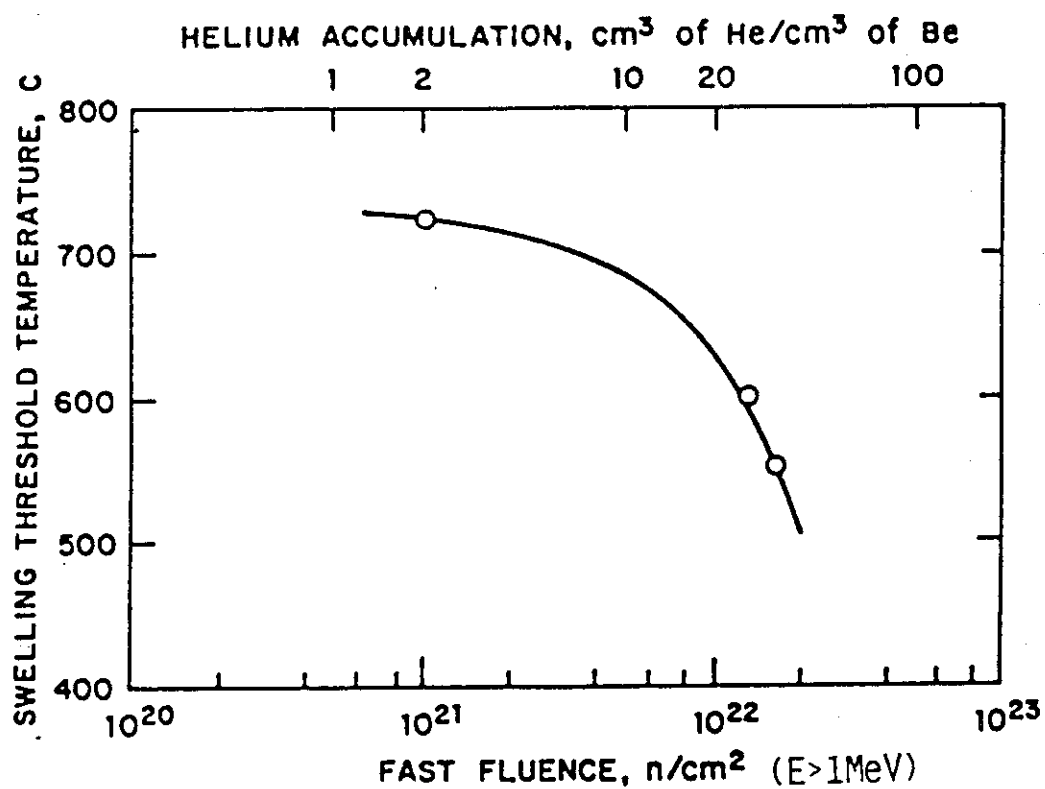


Fig.3.3.3. Swelling threshold of beryllium as a function of gas content and fluence.

At low temperatures, the helium atoms remain in solution in the matrix. However, at higher temperatures, the diffusion of helium atoms becomes significant and they agglomerate into bubbles and consequently the swelling grows larger.

The swelling threshold in irradiated beryllium is shown in Fig. 3.3.3 [27]. The swelling threshold temperature falls as irradiation fluence increases.

The swelling threshold temperature is estimated to be about 400°C for a fluence of  $3.0 \times 10^{22}$  n/cm<sup>2</sup> (~6 MWyr/m<sup>2</sup>). When beryllium is used between room temperature and 200°C, the swelling reaches to 1~3% for a fluence of  $3.0 \times 10^{22}$  n/cm<sup>2</sup>.

#### 4. Lead

The electrical resistivity of lead is shown in Fig. 2.1.2[18].

The elongation of lead at room temperature is about 55%.

There are no data of irradiation effect on the electrical resistivity and the elongation of lead and no data of neutron-irradiated swelling. Lead has large atomic mass and low melting point. Neutrons produce little transmutations in lead. As a result, it seems that the material properties of lead change little.

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## 5. Neutron absorption

We calculated the tritium breeding ratios by ANISN code in order to compare the effects of passive shell materials.

The calculation model is shown in Fig. 5.1 for the copper shell case. In case of aluminum, the composition of shell is 0.932 Al, 0.033 Cu and 0.035 SUS. In case of lead, the sixth zone is removed. In case of beryllium, the sixth zone is removed and lead in neutron multiplier is replaced by beryllium.

The results is shown in Fig. 5.2. The neutron absorption of copper is much larger than that of aluminum. The tritium breeding ratio in case of 3 cm aluminum shell is as large as that in case of 1 cm copper shell.

Zone No.	Radial distance (cm)	Volume fraction	Component
1	0.	He	Plasma
2	110.	He	Vacuum
3	125.	SUS	First wall
4	125.5	SUS(0.3295), H <sub>2</sub> O(0.6705)	
5	126.	SUS	
6	127.	Al or Cu(0.932), Cu(0.033), SUS(0.035)	
7	127. +X	Pb(0.612425), SUS(0.2317), H <sub>2</sub> O(0.1279), Cu(0.03289)	Neutron multiplier
8	127.9 +X	Pb(0.9301), SUS(0.03714), Cu(0.03277)	
9	132.1 +X	Pb(0.61043), SUS(0.22997), H <sub>2</sub> O(0.12695), Cu(0.03265)	
10	133. +X	SUS(0.96740), Cu(0.0326)	
11	134. +X	Li <sub>2</sub> O(0.15018), Be(0.41227), SUS(0.0472), H <sub>2</sub> O(0.02377), Cu(0.03242), Macor(0.05524)	Breeder
12	140.605+X	Li <sub>2</sub> O(0.14815), Be(0.4067), SUS(0.04616), H <sub>2</sub> O(0.02215), Cu(0.03212), Macor(0.06955)	
13	146.49 +X	Li <sub>2</sub> O(0.15272), Be(0.41924), SUS(0.04190), H <sub>2</sub> O(0.01329), Cu(0.03181), Macor(0.0574)	
14	154.325+X	Li <sub>2</sub> O(0.15748), Be(0.4323), SUS(0.038289), H <sub>2</sub> O(6.214-3), Cu(0.031372), Macor(0.04188)	
15	166.22 +X	Li <sub>2</sub> O(0.15262), Be(0.41894), SUS(0.03788), H <sub>2</sub> O(6.45-3), Cu(0.03092), Macor(0.07108)	
16	175.	SUS(0.98474), H <sub>2</sub> O(0.01526)	End wall
	185.		

Fig.5.1. Blanket model for nuclear analysis.

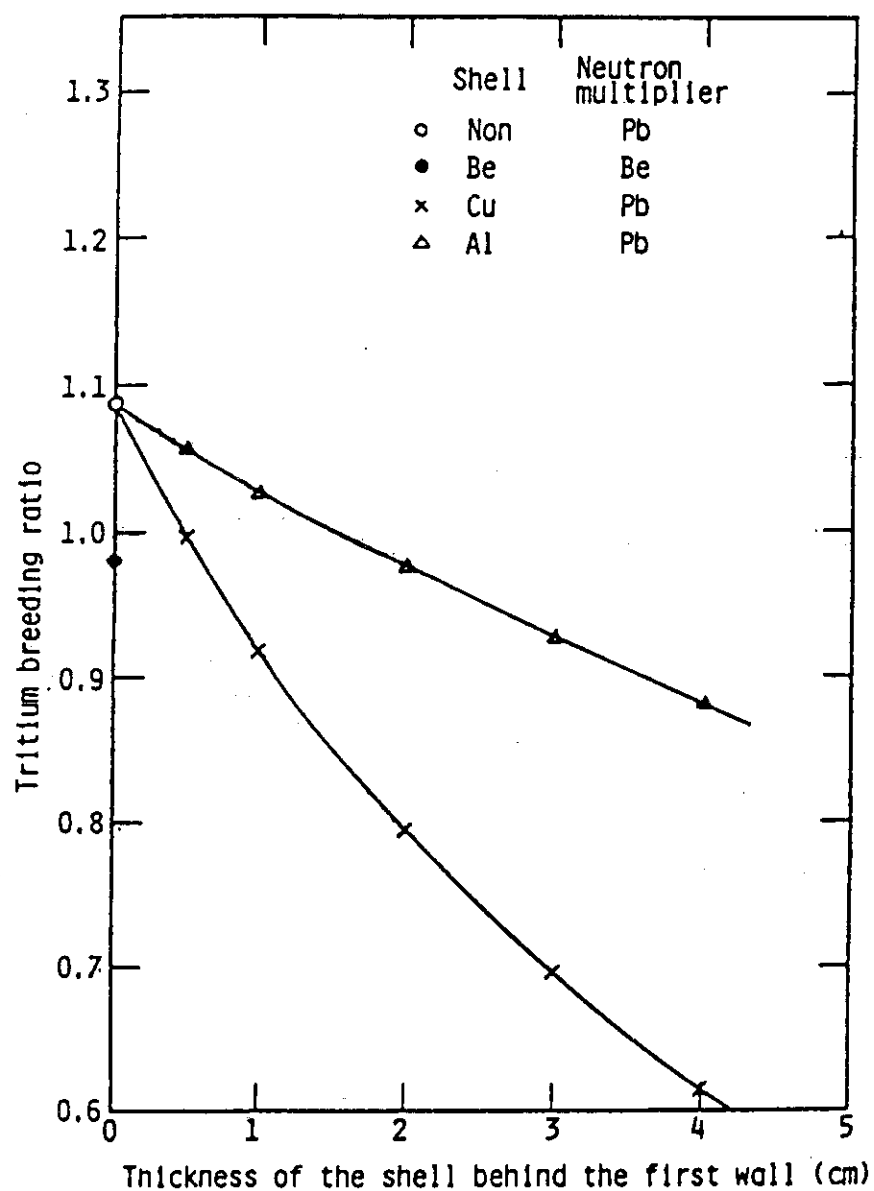


Fig.5.2. Effect of shell material and shell thickness on the tritium breeding ratio.

## 6. Summary

The summary of irradiated properties of candidate passive shell materials is shown in Table 6.1.

Copper has excellent electrical resistivity. However, the resistivity for a fluence of  $6 \text{ MWyr/m}^2$  is 2.7 times as large as the initial value at  $150^\circ\text{C}$ . It seems that the elongation of copper irradiated up to fluence of  $6 \text{ MWyr/m}^2$  at  $200^\circ\text{C}$  remains larger than 5%. The swelling of copper for a fluence of  $6 \text{ MWyr/m}^2$  is estimated to be larger than 10% and this becomes a serious problem. Besides the neutron absorption of copper is large. It seems that the swelling of copper can be suppressed by alloying with appropriate elements. But it is necessary to pay special attention to the electrical resistivity increase due to alloying. It seems that copper is alloyed with Au, Ag, Cd or Zr from the standpoint of electrical resistivity and that about 1 atom% of solute is required.

The electrical resistivity of aluminum at  $150^\circ\text{C}$  is 1.7 times as large as that of copper. The electrical resistivity increase due to neutron irradiation is very small. The elongation of aluminum remains larger than 5% at least up to a fluence of  $6 \text{ MWyr/m}^2$ . The swelling of aluminum to a fluence of  $6 \text{ MWyr/m}^2$  is estimated to be about 1~5%. The neutron absorption of aluminum is smaller than that of copper and it has an advantage over copper from this point of view.

Beryllium is desirable from the standpoint of tritium breeding. The swelling of beryllium irradiated up to a fluence of  $3 \text{ MWyr/m}^2$  is smaller than 1%. The electrical resistivity of beryllium are larger than those of copper or aluminum. However beryllium can serve as neutron multipliers as well, and than the bigger allowable thickness of shell can compensate sufficiency for these large resistivity. The problem of beryllium is its low ductility.

The electrical resistivity of lead is too large. And the low melting point is a problem from the design standpoint.

As mentioned above copper and aluminum have advantage as passive shell material over beryllium and lead.

Admittedly, the data compiled in this report are gotten by fission neutron irradiation and are insufficient to make a final assessment on the behavior of passive shell materials under fusion operating condition. It is necessary to reappraise the effect of 14 MeV neutron irradiation on shell materials.



Table 6.1. Summary of Irradiated Properties of Passive Shell Materials

Service	Material	Electrical resistivity ( $\mu\Omega\text{cm}$ )		Ductility (%)		Swelling (%)	Neutron absorption	Problems
		Initial	6MWyr/m <sup>2</sup>	Initial	6MWyr/m <sup>2</sup>			
Passive shell	Cu (OFHC)	2.6	7.0	46	>5	20-30	Large	Neutron absorption & Swelling
	Al (1100)	4.5	4.9	45	5	1-5	Small	Neutron absorption
Passive shell & Neutron multiplier	Be	6.3	?	4	<0.1	1-3	-	Ductility
	Pb	32	?	55	?	?	-	No Irradiation data & Electrical resistivity
Notice		150 °C		Minimum at 20-200 °C		Maximum at 20-200 °C		

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## **7.2 Benchmark Tests of the Model for Transient Electromagnetics**

## Benchmark Tests of the Model for Transient Electromagnetics

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### 1. Introduction

Models of transient electromagnetics provide eddy current analyses, dynamic analyses of the horizontal and vertical position, analyses of electromagnetic forces caused by interaction between the eddy current and the toroidal magnetic field, etc. In general, the models are composed of three sub-models of an eddy current evaluation model, a suppression and control model of the positional instability, and an electromagnetic force evaluation model.

Therefore, benchmark tests should include some checks for each sub-model described on the above.

In the previous workshop (January, 1984), <sup>(1)</sup> the benchmark models were set in consideration of the following points:

- (1) The benchmark models need to provide the way to check if the models of transient electromagnetics can be endurable for analyses required on the design of a tokamak reactor.
- (2) The structure of the benchmark models should not be so complex that the results obtained from the model of transient electromagnetics developed in each country can not be compared with each other and the differences among countries can not be analyzed.

Finally, three models were set: i.e. Models, 1, 2 and 3. The larger the model number is, the more complex the structure of model gets, as follows.

The Model-1 consists of a pair of ring shells, a plasma column (with the INTOR parameters... Major Radius: 5.3 m, Minor Radius: 1.2 m, Current: 6.4 MA) and a pair of control coils. In Model-2, the ring shells in Model-1 are replaced by 24 rectangular shells (shown in Figure X-26 in the green book<sup>(6)</sup>). In Model-3, radiation shields toroidally divided into 24 divisions are added to the structure of Model-2.

In this report, the following items are set on analyses of three Benchmark Models.

(1) The plasma column is treated as a filamentary current loop located along the major radius.

(2) As for the differential coefficient,  $M'_{ps}$  (with respect to the axial coordinate) of the mutual inductance between the plasma column and the structure (shell, control coils and radiation shield), the results obtained by three kinds of methods are used. Their methods are composed of the method using a filament current loop in place of the plasma column, the dipole approximation for a circular plasma, and the method using a distributed current plasma from plasma equilibrium calculations.

(3) In the case which the shell effect can be expected, the plasma inertial term can be neglected on the equation of plasma column motion.<sup>(7)</sup>

(4) The thin plate approximation method <sup>(8)</sup> is used on estimating the eddy currents flowing in the structure.

In the case of high- $\beta$  INTOR plasma, it has been found that the magnetic axis lies considerably outside the major radius, and the above mentioned differential coefficient,  $M'_{ps}$  can be expected to have larger value than one obtained from setting the radial coordinate of the filament or the dipole on the major radius. On this meaning, the dependence of the shell effects upon their radial coordinates is estimated in this report.

## 2. Requirements and Constraints

### 2.1 Initial conditions

The structures of three benchmark models, which were set in the workshop held in January 1984, are shown in Figures 2.1(a), (b) and (c), respectively. Model-1 consists of a pair of ring shells, the plasma column and a pair of coils. In Model-2, the ring shells in Model-1 is replaced by 24 rectangular shells. In Model-3, radiation shields toroidally divided into 24 divisions are added to the structure of Model-2.

Among the constants used in Models, 1, 2 and 3, those which have been already determined are shown in Table 2.1. Additionally others than figures in the table are used for some constants.

For analyses of the vertical position instability, it is necessary to determine initial conditions since obtained results are anticipated to be dependent upon those conditions.

They can be determined in three ways as follows:

$$(A) \quad Z_p|_{t=0} \neq 0, \quad \dot{Z}_p|_{t=0} = 0 \quad (1)$$

$$(B) \quad B_d = B_0 \cdot [1 - \exp(-t/\tau_d)] \quad (2)$$

$$(C) \quad Z_p|_{t=0} \neq 0, \quad \dot{Z}_p|_{t=0} = \gamma \cdot Z_p|_{t=0} \quad (3)$$

where  $Z_p$  is the displacement in the plasma vertical direction, and denotes the time derivative.  $B_d$  is the horizontal magnetic field due to an external disturbance, and  $B_0$  and  $\tau_d$  are the value of  $B_d$  at  $t \rightarrow \infty$  and the time constant of the external disturbance, respectively.  $\gamma$  is the instability growth time.

In the initial condition (A), the plasma is instantaneously balanced at the position of  $Z_p|_{t=0} \neq 0$  at  $t=0$ , and then the plasma position begins to diverge, because the eddy current balancing the plasma position attenuates according to its time constant. In the case there is a single eddy current mode in the object to be analyzed, the condition (A) are meaningful because the initial values for the eddy current can be definitely determined. In the case there are a number of eddy current modes, however, the condition (A) cannot be used.

In the condition (B), the initial values can be determined even if there are a number of eddy current modes. Moreover, if a value of  $\tau_d$  is chosen properly, the condition  $\dot{Z}_p \approx 0$  can be obtained at a certain point of time  $t > 0$ , which is similar to the condition in (A), thus providing a way to compare the results with those obtained from the condition (A) for a single mode.

As seen from Table 2-1,  $Z_p$  at  $\dot{Z}_p \approx 0$  is taken to be an initial displacement of 1 cm or a maximum displacement of 5 cm. In the case the condition (B) is taken, a position control is set to start at 1 ms after occurrence of the external disturbance. If a stable control can be obtained, the maximum displacement lies in the neighborhood of the displacement at the start of control. Consequently, in this case the initial displacement is found to be equal to the maximum displacement.

The condition (C) can be obtained from the initial condition (B), because the plasma displacement without active control, in the case which the shell effect can be expected, can be represented by the form of  $Ae^{\gamma t}$  after an influence of the disturbance of type (B) will become negligibly small. In this report, very small  $B_d$  ( $\approx 10^{-3}$  Gauss) is selected and the origin of new time is set at  $(t-t')=0$  so that  $Ae^{\gamma t'}$  is equal to  $Z_0$  (for example, 1 cm).

## 2.2 Resistance of control coils

As seen from Table 2-1, the control coils are made of copper. Taking its size into consideration, the self-inductance and the resistance are known to be approximately 54.4  $\mu H$  and 5.6  $\mu\Omega$ , respectively. Since the resistance value of 5.6  $\mu\Omega$  is too small to be realistic, it is important to study if variation of this value largely affects the results given in this report.

In Figure 2.2, variations of the vertical displacement without active control are shown for two cases of 5.6  $\mu\Omega$  and 0.56  $m\Omega$ . Although slight difference can be observed between two cases, the small value of 5.6  $\mu\Omega$  is not supposed to largely affect the results.

In Figure 2.3, the maximum control current and maximum power required for the vertical position control are given. Moderate decrease on both parameters is observed in a range of more value of control coil resistance than 0.5  $m\Omega$ . On the other hand, time to return to 0.1  $Z_0$  is found to abruptly increase in the corresponding range and have  $\infty$  on about 1  $m\Omega$  in its resistance under a control condition: Gain =  $1 \times 10^4$  V/m, Derivative time = 10 msec. In conclusion, active control characteristics for Model-1 are found to change whatever the coil resistance is more than 0.5  $m\Omega$  or not.

## 2.3 Plasma conductor models for evaluating $M'_{ps}$

Three kinds of methods have been proposed on evaluating the differential coefficient,  $M'_{ps}$  (': derivative with respect to the axial coordinate) of the mutual inductance between the plasma column and the structure (shell, control coils, radiation shield, etc).

One of them is the method using a filamentary current loop in place of the plasma column (Here, this method is called the filamentary model), and being under an obligation to use on the benchmark calculation as shown in Table 2.1.

One of the rest methods in the dipole approximation method on which the plasma is assumed to have a circular cross-section with a constant current density, and additionally approximated as a dipole current. Recently, a new dipole method has been proposed.<sup>(10)</sup> In the method both of the dipole position and current are adjusted so as to have a flux function almost equal to the flux function from a plasma equilibrium calculation. This method is called the modified dipole model.

The final method, which is called the distributed current model, evaluates  $M'_{ps}$  using the distributed current distribution obtained by a plasma equilibrium calculation.



Others than the first method using a filamentary current loop are shown in Figure 2.4 where used parameters are defined for each method.

Figure 2.5 shows  $n_s$  ( $=\sum n_i$ )<sup>(3)</sup> which is one of the measures for the shell effects, on comparing some results obtained by each method. In Fig. 2.5 the abscissa is selected so as to be equal to the radial coordinate of the location of the filament in case of the filamentary model, the dipole in case of the dipole and modified dipole approximation or the magnetic axis in case of the distributed current model.

In case of 5.3 m in the radial coordinate, comparison between the filamentary and dipole models is given on Table 2.2. The results from the filamentary model are found to be larger than them in case of the dipole model by 0.2~0.23 in  $n_s$  and by about 53 ms (Model-1), 16 ms (Model-2) and 12 ms (Model-3) in the growth time.

In conclusion, difference between the filamentary and dipole model is considerably large in the growth time, but selection of the radial position, at which the filament, the dipole or the magnetic axis is located, is found to be important, too.

#### 2.4 Comparison between circular coil and finite element models

In case of the eddy current calculation of the Benchmark Model-1, there are proposed three methods.<sup>(2)~(5)</sup> Here, both of the circular coil and finite element models are examined and compared. Used circular coil models are shown in Figure 2.6. Figs. 2.6(a), (b) and (c) show the Benchmark Model-1, a circular coil which the Model-1 is represented by one coil, and two circular coils into which it is divided, respectively. Table 2.3 shows some results obtained by both methods of the circular coil and finite element models. Numerical difference between both methods is within several per cent in  $n_s$ , but is found to attain to a few tens per cent in the growth time,  $\gamma^{-1}$ .

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### 3. Structure of the Benchmark Models

In Figures 3.1 and 3.2, the structure of conductor and a typical eddy current flow ( $1/48$  of a torus) in the Benchmark Model-1 are shown, respectively. In Figures 3.3 and 3.4 are shown the corresponding figures in the Benchmark Model-2. In Figures 3.5 and 3.6 the structure of the upper half of a  $1/24$  sector (a portion of 7.5 degrees in the torus direction) and its typical eddy current flow are shown, respectively.

Although the thickness of side walls in Models, 2 and 3 is selected to be 2 cm as shown in Table 2.1, the case of 1 cm in its thickness is calculated in this report, too. The gap between adjacent side walls is taken to be 1 cm, as the distance between the currents is set to be 3 cm, and then they are supposed to flow along the center of side wall thickness.

## 4. Results

### 4-1 Model-1

#### 4.1.1 Initial condition of type (B)\*

In Figure 4.1 is shown time evolution of the vertical displacement of plasma column,  $Z_p$ , which is obtained when external disturbance of the form of Equation (2) is applied to the Benchmark Model-1. Time evolutions of the horizontal magnetic field due to the external disturbance, the magnetic field due to the eddy currents, and the current in the control coils which are regarded as passive conductors, are also shown in the figure. The time constant of the control coils,  $\tau_c$ , is taken to be approximately 9.73 s ( $= 54.5 \mu\text{H}/5.6 \mu\Omega$ ). The magnetic field due to the external disturbance at  $t \rightarrow \infty$ ,  $B_o$ , is taken to be 3.29 Gauss, which is determined so that  $Z_p$  is equal to 1 cm at  $t=1$  ms.  $\tau_d$  is selected to be equal to 0.1 ms.

The initial condition for position control is given at  $t=1$  ms, and are used in the following discussion for the Benchmark Model-1. This initial condition will be found to be approximately equivalent to the initial condition of type (A) as known in Figure 4.1.

In Figure 4.2, time evolutions of  $Z_p$  are shown for the cases with and without control. In case with control, the gain is fixed at  $10^4$  V/m, and the derivative action time,  $T_d$ , is taken at 1 ms or 10 ms. In the case without control, the instability growth time is approximately 72 ms, and at  $s \rightarrow \infty$ ,  $n_s$  is approximately 1.72.

In case of the control with  $T_d = 10$  ms,  $Z_p$  tends to slightly fluctuate after external disturbance is applied. In Figure 4.3 is shown Bode diagram with  $T_d$  as a parameter. Since less frequency than  $10^3$  is used, there is no unstable factors in phases. Consequently, it is seen that instability occurs when  $T_d = 10$  ms, because the gain is set to be too large.

In Figure 4.4, the control characteristics of  $Z_p$  are shown for  $T_d = 1$  ms, and  $Z_p$  can be seen to stabilize by about 60 ms. Time evolution of the magnetic field due to the eddy current, the magnetic field produced by a pair of control coils, and the current and voltage of the control coils are also shown in the figure. The time in which  $Z_p$  is stabilized is almost the same as the time by which the voltage of control coils reaches a stationary value. The magnetic field due to the eddy current tends to decrease after  $Z_p$  is stabilized. Since the time dependence of the current of control coils is similar to that of the eddy current, it can be seen that the field by the control coils after  $Z_p$  is stabilized plays a role of cancelling the magnetic field due to the eddy current.

In Figure 4.5, time evolutions of the same parameters as in Figure 4.4 are shown for the case of  $T_d = 10$  ms. Although both figures give similar curves, the time in which  $Z_p$  is stabilized is longer, and the time, at which the eddy current and the current of control coils turn to decrease, is different from then in case of  $T_d = 1$  ms. This is caused by the fluctuation of  $Z_p$  on the initial stage of control and shows that selection of a suitable gain is important.

\* When the initial condition of this type is used, the dipole model is used for evaluation of  $M'_{ps}$  all over this report.

#### 4.1.2 Initial condition of type (C)\*\*

Here, some characteristics of the vertical position control under the initial conditions of type (C) are studied. These initial conditions are set on the INTOR Workshop, May, 1984 and ought to be used on the benchmark tests.

Figure 4.6 shows time evolutions of the vertical position,  $Z_p$ , the control current,  $I_c$  and the required power,  $P_c$  under the initial condition of type (C). Figs. 4.6 (a) and (b) correspond to 1 ms and 10 ms in derivative time,  $T_d$  with a fixed gain,  $10^4$  V/m, respectively. The time evolution of each parameter is found to show the almost same feature as one in Figure 4.4 or Figure 4.5.

### 4.2 Model-2

#### 4.2.1 Initial condition of type (B)

In Figure 4.7, time evolutions of  $Z_p$  in the Benchmark Model-2 are shown for both cases with and without active control. The gain with active control is fixed at  $10^4$  V/m, and the derivative action time,  $T_d$ , is set at 1 ms or 10 ms as used in Figure 4.2. The instability growth time is approximately 20 ms, and  $n_s$  at  $s \rightarrow \infty$  is approximately 1.62 in this case. Although there is not significant difference in the values of  $n_s$  in Models, 1 and 2, both values of instability growth time differ largely.

The control characteristics vary largely as the value of  $T_d$  changes.  $Z_p$  at  $T_d = 1$  ms is controlled so as to be stable, but diverging oscillation occurs at  $T_d = 10$  ms. In Figure 4.8, Bode diagram corresponding to Figure 4.7 is shown. Figure 4.8 suggests that instability occurs when  $T_d = 10$  ms, because the gain is set to be too larger.

Figure 4.9 shows time evolutions of the same parameters of Model-2 as in Figure 4.4. It can be seen that the time in which  $Z_p$  is stabilized is approximately 40 ms, and is a little less than that for Model-1, as shown in Figure 4.4. Although time evolution of the magnetic field due to the eddy current is seen to be similar to that of the magnetic field due to the control coil current in Figure 4.4, the curves for these quantities show some differences in the starting part in this figure. The current of control coils rises rapidly because it is necessary for them to play a role of the eddy current, and moreover to produce larger magnetic field than one produced by the eddy current.

The time in which  $Z_p$  is stabilized gets shorter because the time constant of shell itself, or the shell effect, is decreased on comparison with one of Model-1.

#### 4.2.2 Initial condition of type (C)

Figure 4.10 shows time evolution of  $Z_p$ ,  $I_c$  and  $P_c$  under the condition (C). Fig. 4.10 (a) corresponds to 1 ms in  $T_d$  and the time evolution of each parameters shows the almost same feature as one in Fig. 4.9. Fig. 4.10 (b) corresponds to 10 ms in  $T_d$ . Some signs of the unstable tendency, which the case of 10 ms in  $T_d$  in case of the condition of type (B) has

\*\* When this condition is used, the filamentary model is used for evaluation of  $M'_{ps}$  all over this report.

shown, are observed from Fig. 4.10(b), but the disturbance is stabilized little by little.

### 4.3 Model-3

#### 4.3.1 Initial condition of type (B)

In Figure 4.11, Bode diagram is shown for the Benchmark Model-3. It can be expected that a stable control may be obtained when the derivative action time is 10 ms.

Figure 4.12 shows time evolutions of the same parameters of Model-3 as them in Fig. 4.4.  $Z_p$  is seen to fluctuate slightly on the initial stage, but the fluctuation attenuates in a short period.

For this model, the instability growth time is about 26 ms, and  $n_s$  is approximately 1.85. Additionally the instability growth time in case of the side walls with 1 cm in thickness decreases about 18 ms.

#### 4.3.2 Initial condition of type (C)

Figure 4.13 shows time evolution of  $Z_p$ ,  $I_c$  and  $P_c$  under the condition (C). Figs 4.13 (a) and (b) correspond to 1 ms and 10 ms in  $T_d$ , respectively.

### 4.4 Required Capacity of Power Supply

#### 4.4.1 Initial condition of type (B)

The required capacity of power supply was evaluated for the Benchmark Models, 1, 2 and 3 and the results are given on Table 4.1. There are shown the required maximum voltage and current together.

As for Model-1, calculations are made for three values of  $B_0$ . These values of  $B_0$  give the maximum vertical displacements of 1, 3 and 5 cm (at 1 ms after imposition of each external disturbance), respectively. Using the calculated results, the required capacity of power supply ( $V_{\max}$ : maximum voltage,  $I_{\max}$ : maximum current) is given as follows:

$$P = 0.2072 \cdot B_0^2 \text{ (MVA, } B_0: \text{ Gauss)} \quad (3)$$

$$= 2.248 \cdot Z_{\max}^2 \text{ (MVA, } Z_{\max}: \text{ cm)} \quad (4)$$

where  $Z_{\max}$  is the maximum vertical displacement. These equations can be used under the conditions that  $G = 10^4$  V/m,  $T_d = 1$  ms and  $\tau_d = 0.1$  ms.

In case of Model-1,  $\tau_c$  is calculated for a case set at 0.0973 s, too. (taking the resistance of control coils up two orders) and the required capacity of power supply is evaluated about 5% smaller than one in the case of 9.73 s in  $\tau_c$ .

In cases of Models, 2 and 3, a value of  $B_0$  is set so that an initial displacement of about 1 cm can be given. There is only a little difference with respect to the shell effect between Models, 2 and 3. On Table 4.2, the major parameters are shown. It can be seen that the figures for two

models are nearly equal. However, the required capacity for Models, 2 and 3 is about 2.0 and 6.4 MVA, respectively. This is caused by the structural difference of two models, that is, with and without radiation shield and also means that the magnetic shield effect of the radiation shield is significant.

Table 4.2 shows number of the modes, taken into consideration in the analysis, whose contribution to  $n_s$  is greater than  $5 \times 10^{-3}$ . Whether the radiation shield exists or not is very different in the number of the modes taken for Models, 2 and 3.

#### 4.4.2 Initial condition of type (C)

On Table 4.3, main results for three benchmark models in case of the initial condition of type (C) are listed.

It is found from the table that the required capacity of power supply is only a little dependent on the instability growth time, except Model-1b. Such a result seems to mean that, in case of the initial condition of type (C), only the vertical displacement,  $Z_0$  influences mainly one the required capacity of power supply.

#### 4.5 Disruption Analysis

The eddy current and the electromagnetic force on a plasma disruption are analyzed using Model-2.

The disruption condition to be assumed is that the plasma current decreases linearly at the stationary position and at 20 ms in time constant.

Figure 4.14 shows variations of the eddy current on the rectangular shell for the case which the plasma disappears in the way described above. (a), (b), (c) and (d) show the results of observation at the time of 5, 10, 20 and 30 ms, respectively. The direction of the eddy current in the upper and lower parts of the toroidal bar is reversed, and a loop current is formed so as to flow through the side boards. The loop current reaches the maximum of about 260 KA at the time of 20 ms. The toroidal field (4.6T at  $R = 6.34$  m) is coupled with the current flowing through the side walls.

The triangle, which will have the maximum line current density (A/m), is found to be the dark one of the side wall in Fig. 4.14. The electromagnetic force due to coupling between the toroidal magnetic field and the eddy current on the disruption has the maximum value all over the period of disruption on the above triangle of side wall, too, and the maximum is found to attain to about 50 in  $\text{kgf/cm}^2$  at 20 ms after the disruption. Fig. 4.15 shows time evolutions of the plasma current, the looped eddy current on the toroidal bar, and the electromagnetic force on the dark triangle of side wall.

Neighboring side walls have an equal value but mutually inverse direction with the induced eddy currents at every time of the disruption, and therefore their currents result in a repulsive force between them. On evaluating this repulsive force under an assumption of infinitely extended parallel-planes for the neighbouring side walls, the force

becomes about  $7.2 \text{ in kgf/cm}^2$  at 20 ms and is found to be negligibly small compared with the above electromagnetic force between the toroidal magnetic field and the eddy current.

#### 4.6 Current Flow on the Radiation Shield

Figures 4.16 and 4.17 show variations of current integral ( $\oint_a^s J_\theta \cdot t \, ds$ )<sup>(9)</sup> on the radiation shield in cases of  $T_d = 1 \text{ ms}$  and  $10 \text{ ms}$  for a fixed gain,  $10^4 \text{ V/m}$ , respectively. The current flow between a and c is found to have the largest one for both cases of Figs. 4.16 and 4.17.

The minimum current on time evolution is at  $t = 0 \text{ ms}$  and the maximum occurs at about several tens ms on both cases, but decreasing speed (for example,  $\text{KA/ms}$ ) from the maximum is found to be faster in the case of  $T_d = 1 \text{ ms}$  than one of  $T_d = 10 \text{ ms}$ .



## 5. Conclusions

From some studies with three kinds of benchmark models, which were proposed on the INTOR Workshop, January ('84) and modified partially on the Workshop, May ('84), are obtained the following results.

### (1) Instability growth time, $\gamma^{-1}$

The results (about 125, 36 and 38 ms for Models, 1, 2 and 3, respectively) from the filamentary model are larger than them in case of the dipole model by 0.20~0.23 in  $n_s$  and by about 53 ms (Benchmark Model-1), 16 ms (Model-2) and 12 ms (Model-3) in the growth time, respectively. Moreover, the selection of the radial position, at which the filament (of the filamentary model), the dipole (of the dipole approximation) or the magnetic axis (of the distributed current model) is located, is found to be important, too.

### (2) Required capacity of power supply

Required capacity of power supply was studied under two kinds of initial conditions shown in the following conditions.

$$(1) \quad Z_p|_{t=0} \neq 0, \quad \dot{Z}_p|_{t=0} \neq 0$$

$$(2) \quad Z_p|_{t=0} = 1 \text{ cm}, \quad \dot{Z}_p|_{t=0} = \gamma \cdot Z_p|_{t=0}$$

Under the initial condition, (1), based on a disturbance field ( $B_0 = 4.18$  Gauss and  $\tau_d = 0.1$  ms), the required capacities for Models, 1, 2 and 3 are about 3.6, 6.5 and 6.4 in MVA, respectively, and difference between Model-1 and the others on the required capacity is found to be smaller than one in the instability growth time. The maximum vertical displacement under this disturbance field for each of Models, 1, 2 and 3 is not kept constant and is about 1.27 cm, 1.8 cm and 1 cm (standard), respectively.

On the other hand, under the initial condition (2), the required capacity is only a little dependent on the instability growth time. Such a result seems to mean that, in case of this initial condition, only the initial vertical displacement,  $Z_0$  influence mainly on the required capacity of power supply. The capacity for each Model-1, 2 or 3 in case of  $T_d = 1$  ms is about 3.22, 2.82 or 3.55 MVA.

### (3) Electromagnetic force

The maximum electromagnetic force by the toroidal magnetic field and the induced eddy current on the disruption is produced on the top and bottom of side walls, directs normal to and out of the toroidal bar and parallel to the side wall, and attains to about 50 in  $\text{kgf/cm}^2$  at 20 ms.

Table 2.1 - 1 Parameter Values for Benchmark Models

Model	1	1 bis	2	3
Shell Geometry				
no. of sectors	0	0	24	24
shell shape	toroidal	toroidal	saddle	saddle
toroidal shell thickness, $t_1$ , m	0.01	0.01	0.01	0.01
sidewall thickness, m	0	0	0.02	0.02
sidewall width, m	0	0	0.5	0.5
gap between adjacent sectors, m	-	-	0.03	0.03
Radiation Shield				
no. of sectors	-	-	-	24
thickness, $t_2$ , m	-	-	-	0.1
resistivity, $\Omega\text{m}$	-	-	-	$10^{-6}$
gap between adjacent sectors, m	-	-	-	0.01
Plasma Current, MA	6.4	6.4	6.4	6.4
Plasma Radius, m	5.3	-	5.3	5.3
Plasma Model	1 filament	distributed	1 filament	1 filament
Plasma Mass, kg	0	0	0	0
Vertical Field, T	-0.5	-0.5	-0.5	-0.5
Field Index	-1.3	-1.5517	-1.3	-1.3

Table 2.1 - 2 Additional Parameters for Model 1 bis only

Between plasma and;	M ( $10^{-6}\text{H}$ )	$dM/dz$ ( $10^{-6}\text{H/m}$ )
passive upper loop	13.4	2.91
passive lower loop	8.77	-2.83
active upper loop	3.09	0.747
active lower loop	2.25	-0.505

Table 2.1 - 3     Control Parameters for All Models

control coil location:      $6.5 \text{ m} \pm 6.5 \text{ m}$   
 cross-section per coil:      $0.5 \times 0.5 \text{ m}$   
 resistivity (solid Cu):      $1.72 \times 10^{-8} \Omega\text{m}$   
 control law:      $V_c = -G(Z_p + T_d Z_p)$

$V_c$  = voltage across two coils connected series opposing

$G = 10^4 \text{ V/m}$

$T_d = 1 \text{ ms}$

Table 2.1 - 4     Additional Parameters for Model 2 Only

For Model 2 Plasma Disruption:

- 1) Plasma stationary at (5.3, 0)
- 2) Time scale: 20 ms linear decay to 0 current
- 3) Plasma initial current = 6.4 MA
- 4) Toroidal field =  $5.5 \times 5.3/R, \text{ T}$ .

Table 2.2 Comparison between Filamentary and Dipole Models

Parameters	Plasma Models Benchmark Models			Filamentary Model			Dipole Model		
	1*	2	3	1*	2	3	1*	2	3
$n_g$	1.88	1.82	2.08	1.72	1.62	1.85			
$\gamma^{-1}$ (ms)	125	36	38	72	20	26			

\* This is in case of Ne = 12 [Ne: Number of the triangle elements (per 7.5 degrees in the toroidal angle) of which the Benchmark Model 1 is composed].

Table 2.3 Comparison between Circular Coil and Finite Element Models\*\*\*

Shell Models Parameters	Circular Coil Model			Finite Element Model	
	$N_c^*=1$	$N_c^*=2$	$N_c^*=10$	$N_e^{**}=12$	$N_e^{**}=24$
$n_s$	2.11	-	-	1.88	2.00
$\gamma^{-1}$ (ms)	163	139	111	.125	141

\*  $N_c$ : Number of the circular coils of which the Benchmark Model 1 is composed.

\*\*  $N_e$ : Number of the triangle elements (per 7.5 degrees in the toroidal angle) of which the Benchmark Model 1 is composed.

\*\*\* These results were obtained from the filamentary model.

Table 4.1 Required maximum voltage and current for Model-1, 2 and 3

Model No.	B <sub>0</sub> (Gauss)	τ <sub>c</sub> (sec.)	T <sub>d</sub> (ms)	V <sub>max</sub> (Volt)	I <sub>max</sub> (10 KA)	V <sub>max</sub> × I <sub>max</sub> (MVA)
1**	3.29	9.73	10	116.7	2.225	2.597
		9.73	1	102.4	2.196	2.248
		0.0973	10	117.1	2.121	2.484
	9.88	9.73	1	307.4	6.592	20.26
		9.73	10	350.4	6.681	23.41
	16.47	9.73	1	512.3	10.986	56.28
2	2.32	9.73	1	109.0	1.848	2.014
		9.73	10	V <sub>max</sub> > 5kV required	-	-
3	4.18	9.73	10	180.6	3.522	6.361

\* (1) Gain, G is fixed at 10<sup>4</sup> V/m.

(2) Dipole current model for a circular cross-sectional plasma is used.

\*\* P(≡V<sub>max</sub>·I<sub>max</sub>) can be approximated by the following equation,

$$P = 0.2072 \cdot B_0^2 = 2.248 \times Z_{\max}^2, \text{ (MVA, } B_0: \text{ Gauss, } Z_{\max}: \text{ cm)}$$

where Z<sub>max</sub> is the vertical displacement in cm.

Table 4.2 Major parameters for evaluating shell effects

Model No. Contents	1	2	3
• Number of total eddy current modes	3	6	32
• $\sum \eta_i$	1.72 (1.65)	1.62 (1.56)	1.85 (1.78)
• The maximum time constant of modes (s)	0.283	0.097	0.101
• Instability growth time (ms)	72 (59)	20 (16)	26 (22)

Note: Numerical values with ( ) are in case of τ<sub>c</sub> = 97.3 ms.

Table 4.3 Main Results for Three Benchmark Models (1)

Model $T_d$ (ms)	1		2(2)		3(2)	
	1a(2)					
	1	10	1	10	1	10
Parameters						
$I_c^{MAX}$ (KA)	-32.2	-31.0	13.5	-27.1	-25.8	-31.3
$V_c^{MAX}$ (VOLT)	-100.0	-108.0	-107.0	-104.0	-128.0	-125.0
$P_c^{MAX}$ (MW)	1.1	0.72	0.40	1.0	0.75	1.1
$T_R$ (ms) <sup>(3)</sup>	40	57	11	28	44	46
$T_I$ (ms) <sup>(3)</sup>	48	60	23	31	38	38
$T_p$ (ms) <sup>(3)</sup>	16	16	5	12	12	14
$\eta_s$	1.88	1.88	1.80	1.82	1.82	2.08
$\gamma^{-1}$ (ms)	125	125	39	36	38	38

(1) Grain, G is fixed at  $10^4$  v/m. (2) The filamentary model is used.  
 (3)  $T_R$ : Time to 0.1  $Z_0$ ,  $T_I$ : Time to Max. current,  $T_p$ : Time to Max. Power.

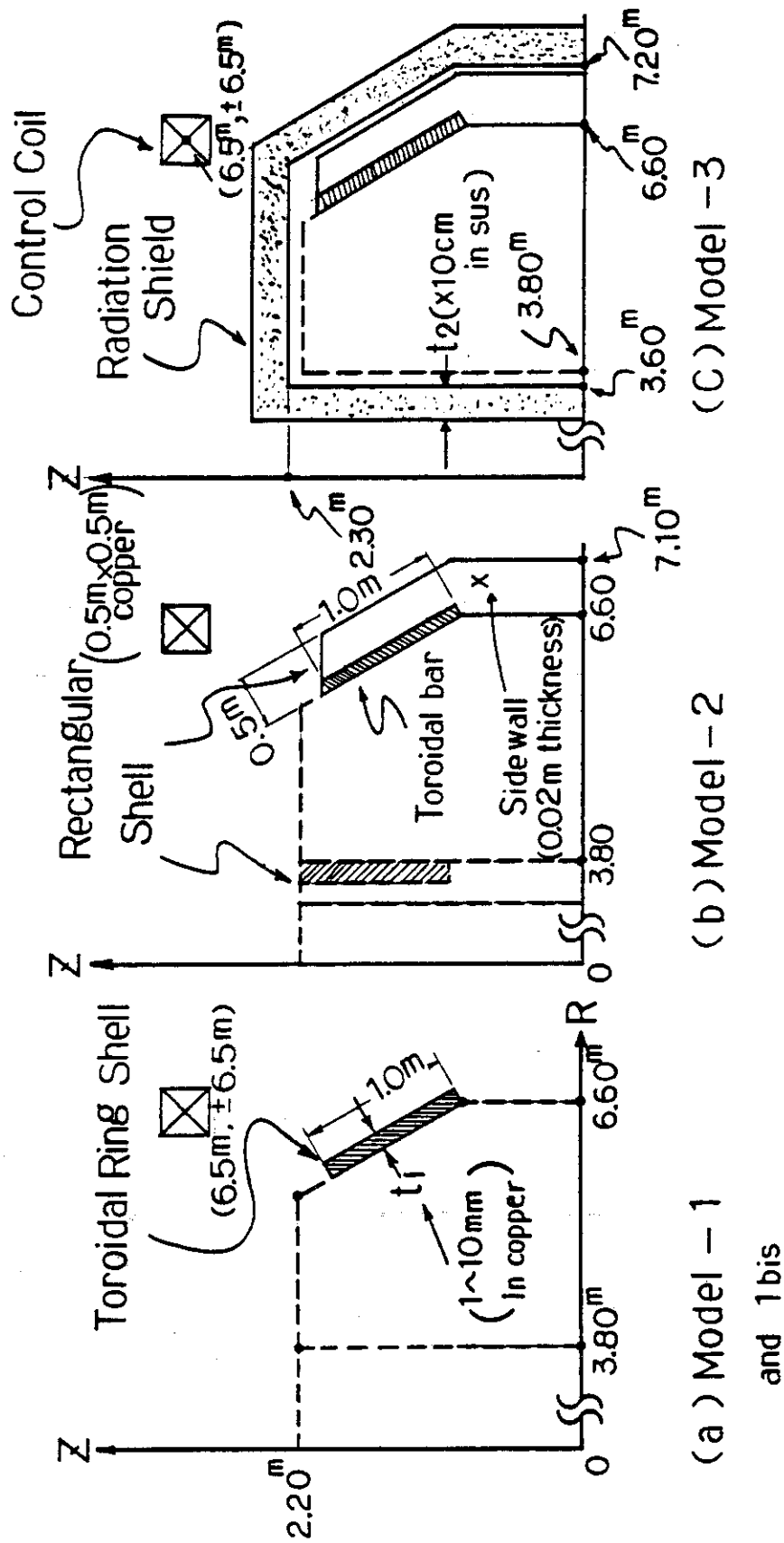


Fig. 2.1 Three benchmark models



# INTOR BENCHMARK MODEL-1

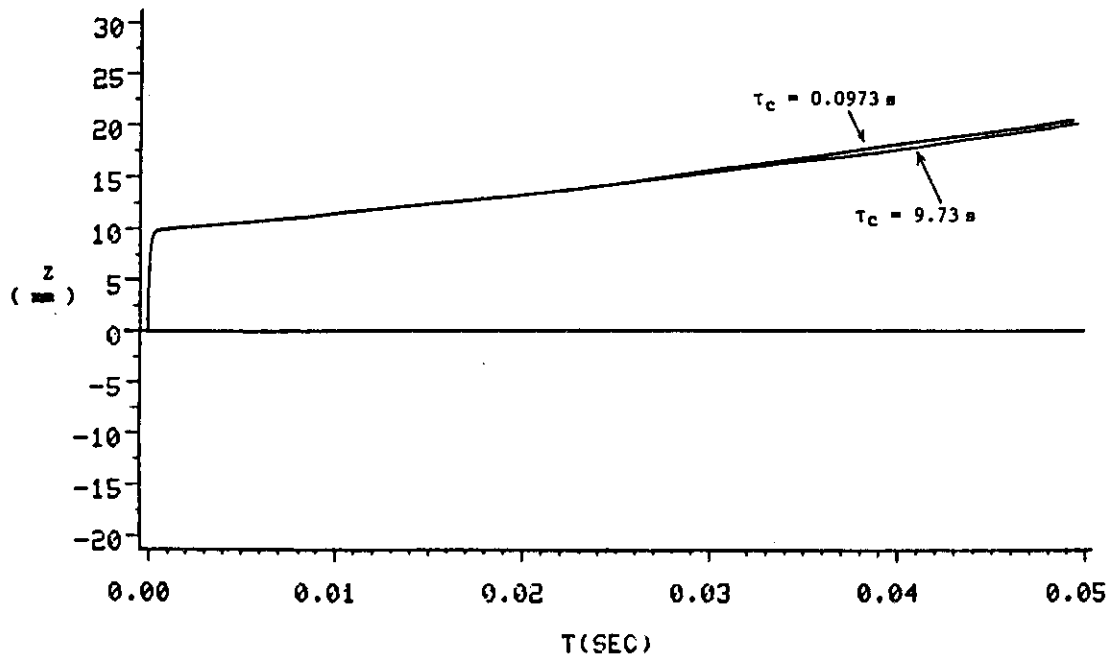


Fig. 2.2 Time evolution of the vertical displacement without PD controller for two cases of 5.6 and 560  $\mu\Omega$  in the resistance of control coil

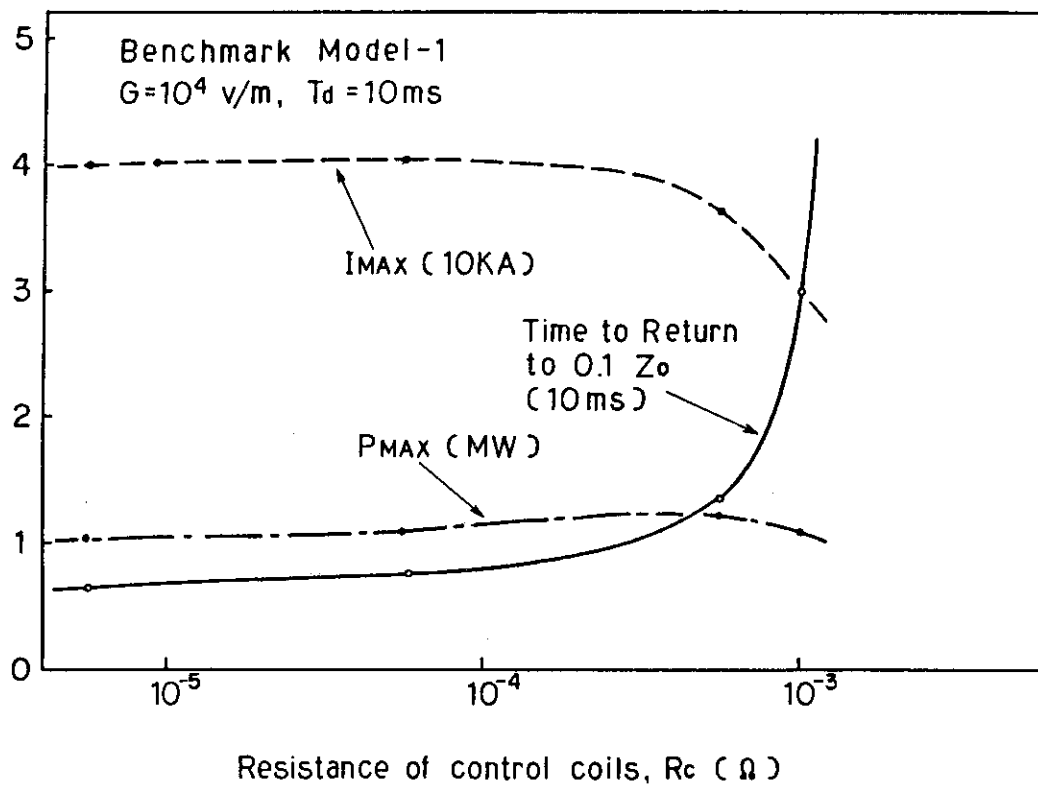
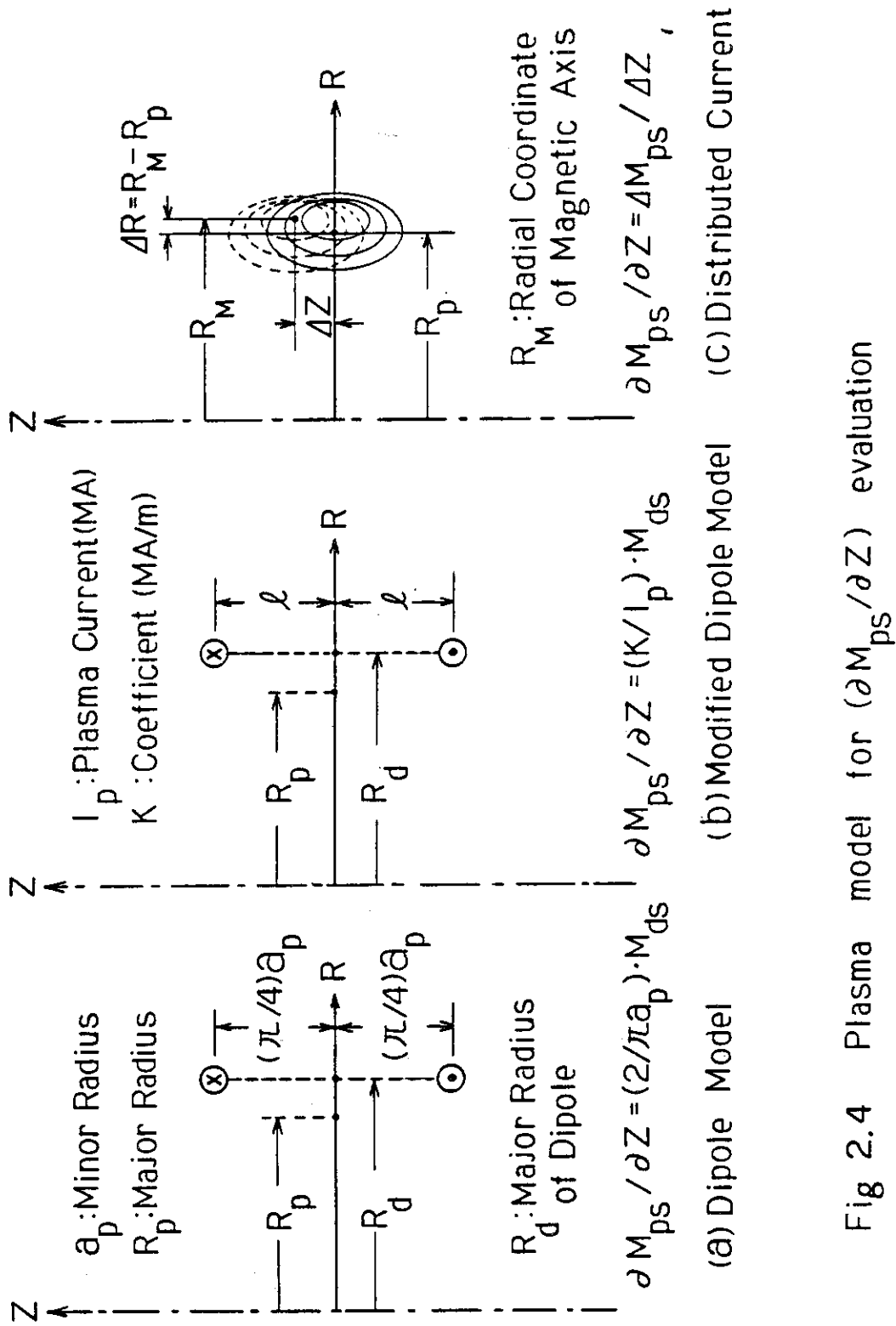


Fig. 2.3 Dependence of time to return to 0.1  $Z_0$  upon control coil resistance,  $R_c$ .

Fig 2.4 Plasma model for  $(\partial M_{ps} / \partial Z)$  evaluation

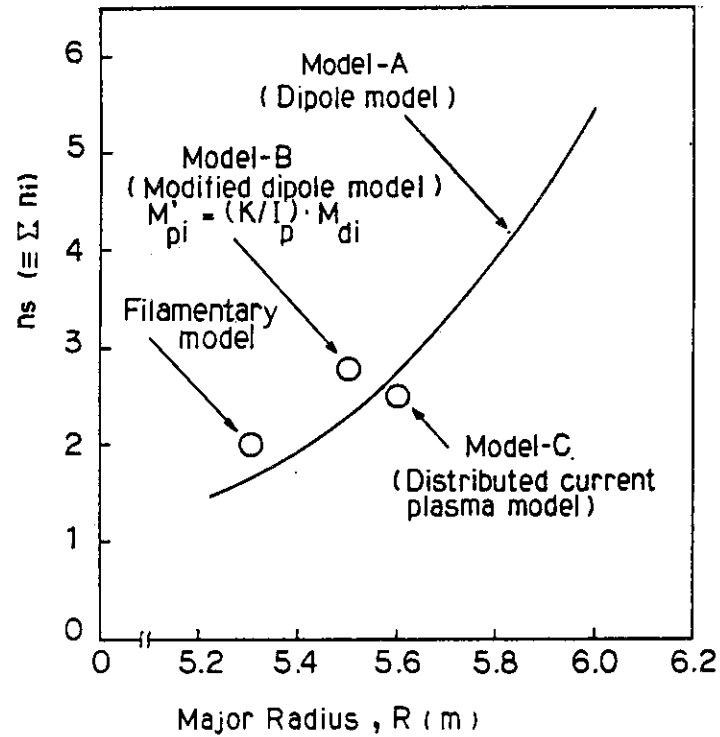


Fig. 2.5 Results of  $\Sigma n_i$  evaluation obtained by the filamentary, dipole and distributed current models.

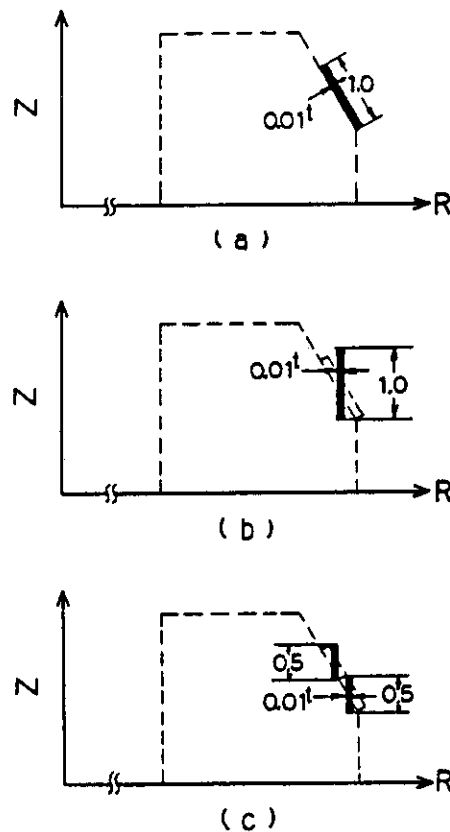


Fig. 2.6 Shell cross-sections used for test calculations.

★BENCHMARK MODEL-1★ TOROIDAL RING SHELL

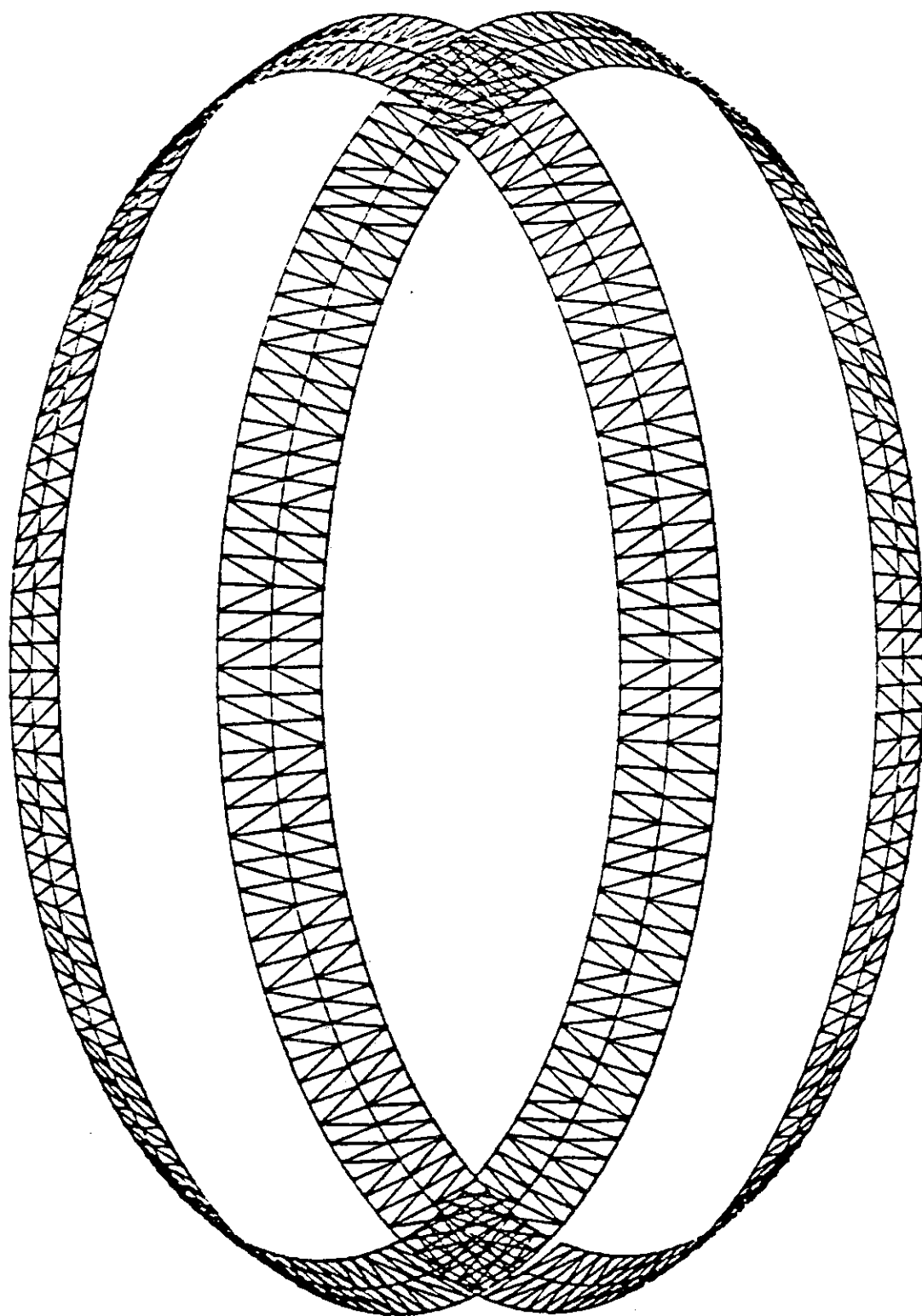


Fig. 3.1 A structure of finite elements for Benchmark Model-1

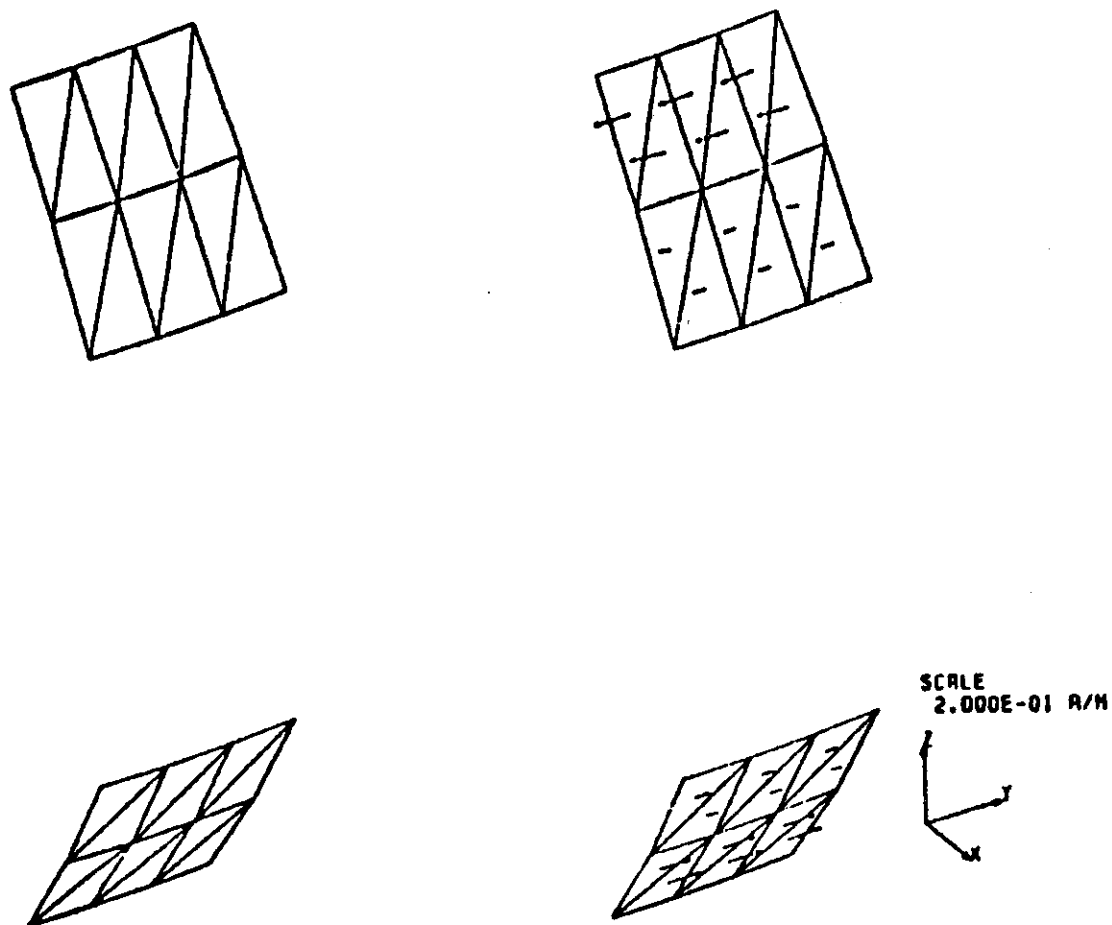


Fig. 3.2 An example of eddy current distributions for Model-1

✕BENCHMARK MODEL-2✕ RECTANGULAR SHELL

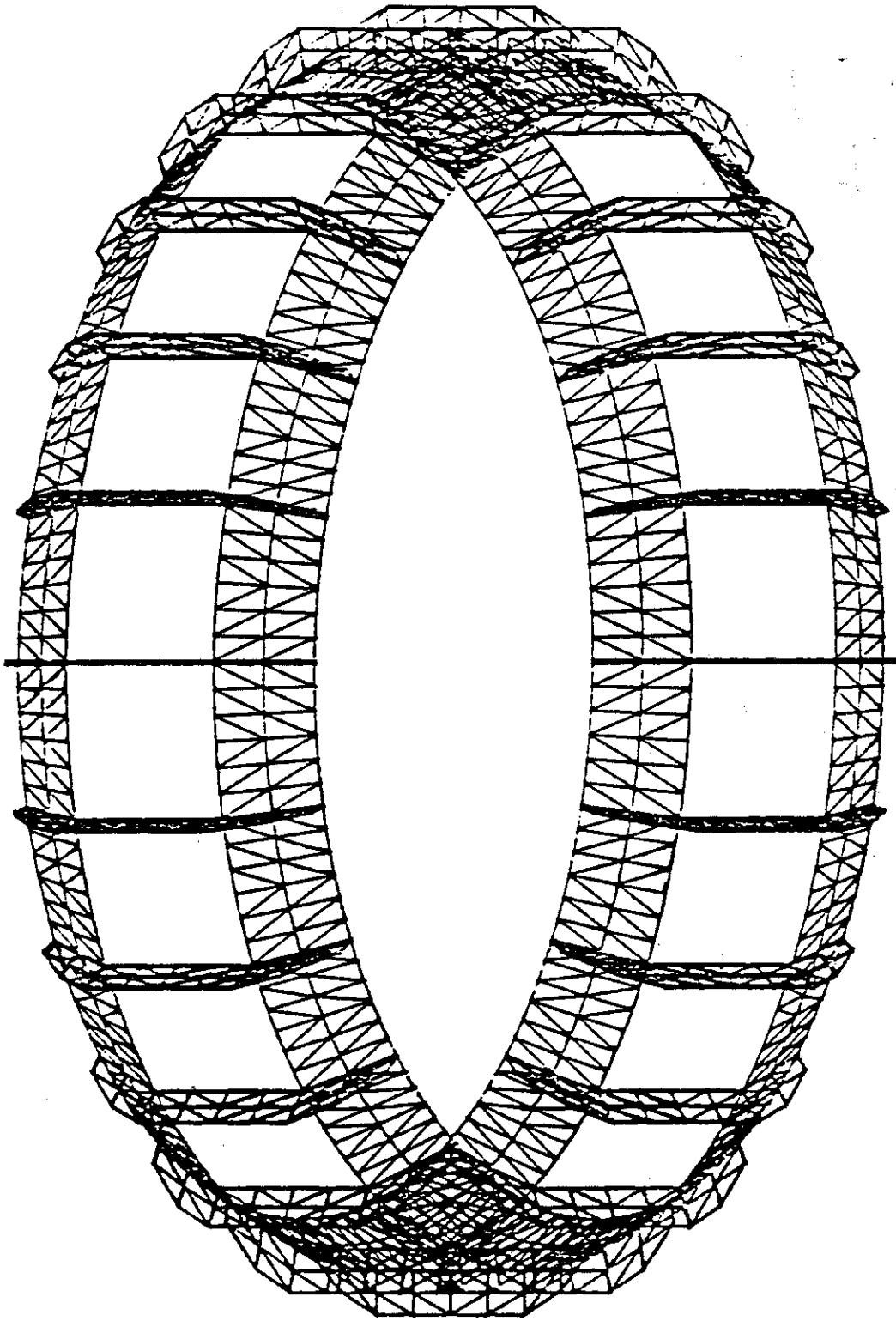
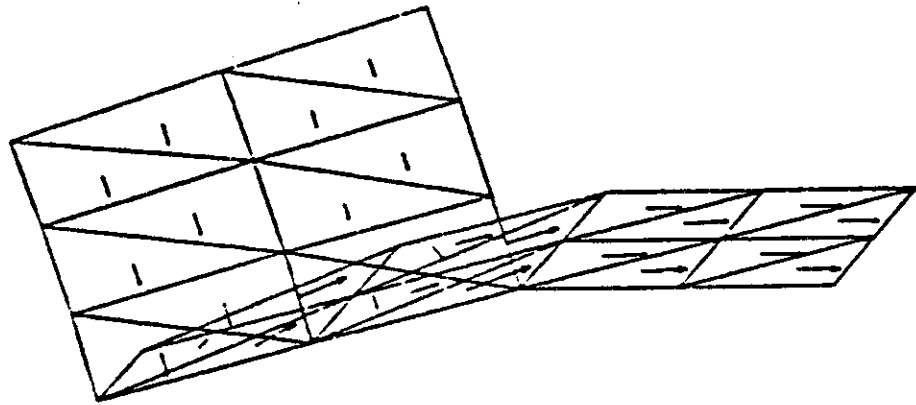
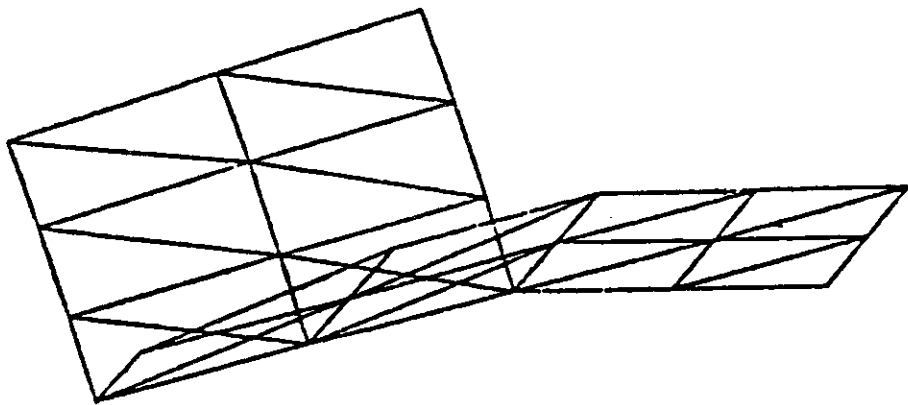


Fig. 3.3 A structure of finite elements for Benchmark Model-2

BENCHMARK MODEL-2# RECTANGULAR#COIL#  
 TIME (SEC) = 1.0000E-02  
 JOULE LOSS (W) = 3.20986E-06



SCALE  
 2.000E-01 A/M

Fig. 3.4 An example of eddy current distributions for Model-2

×BENCHMARK MODEL-3×

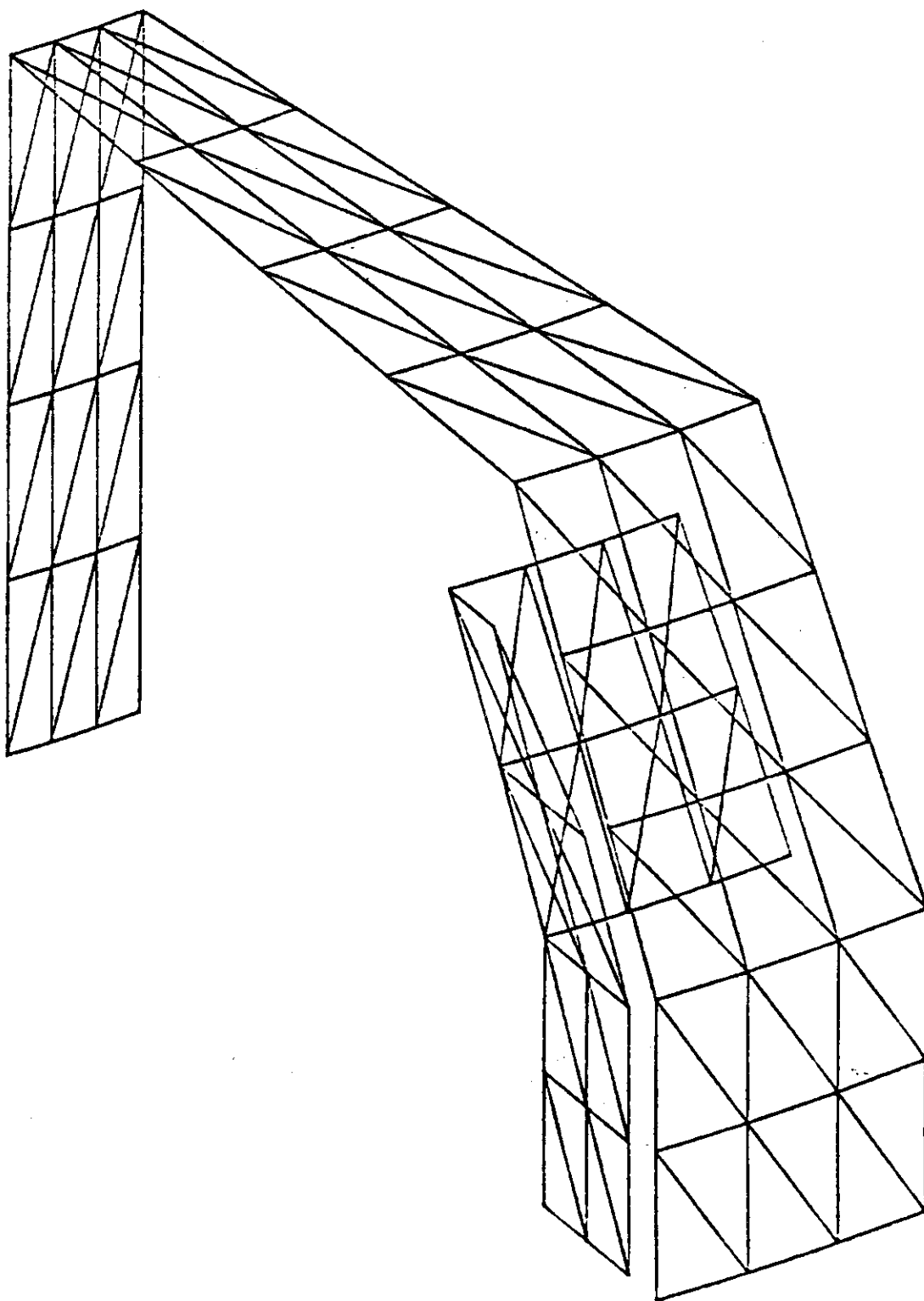


Fig. 3.5 A structure of finite elements for Benchmark Model-3



\*BENCHMARK MODEL-3\*  
 TIME(SEC)= 1.00000E-02  
 JOULE LOSS(W)= 4.41795E-06

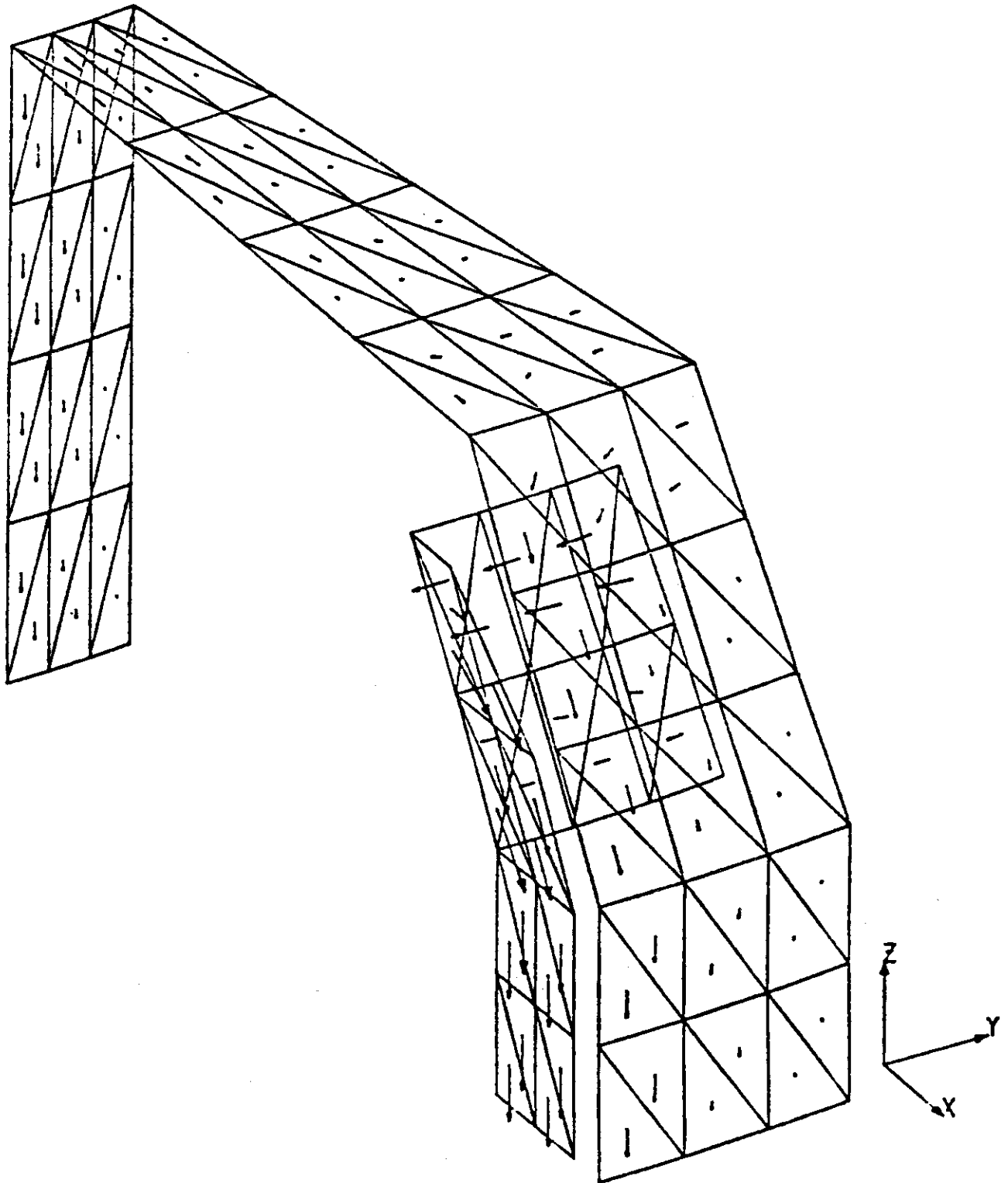


Fig. 3.6 An example of eddy current distributions for Model-3

## INTOR BENCHMARK MODEL-1

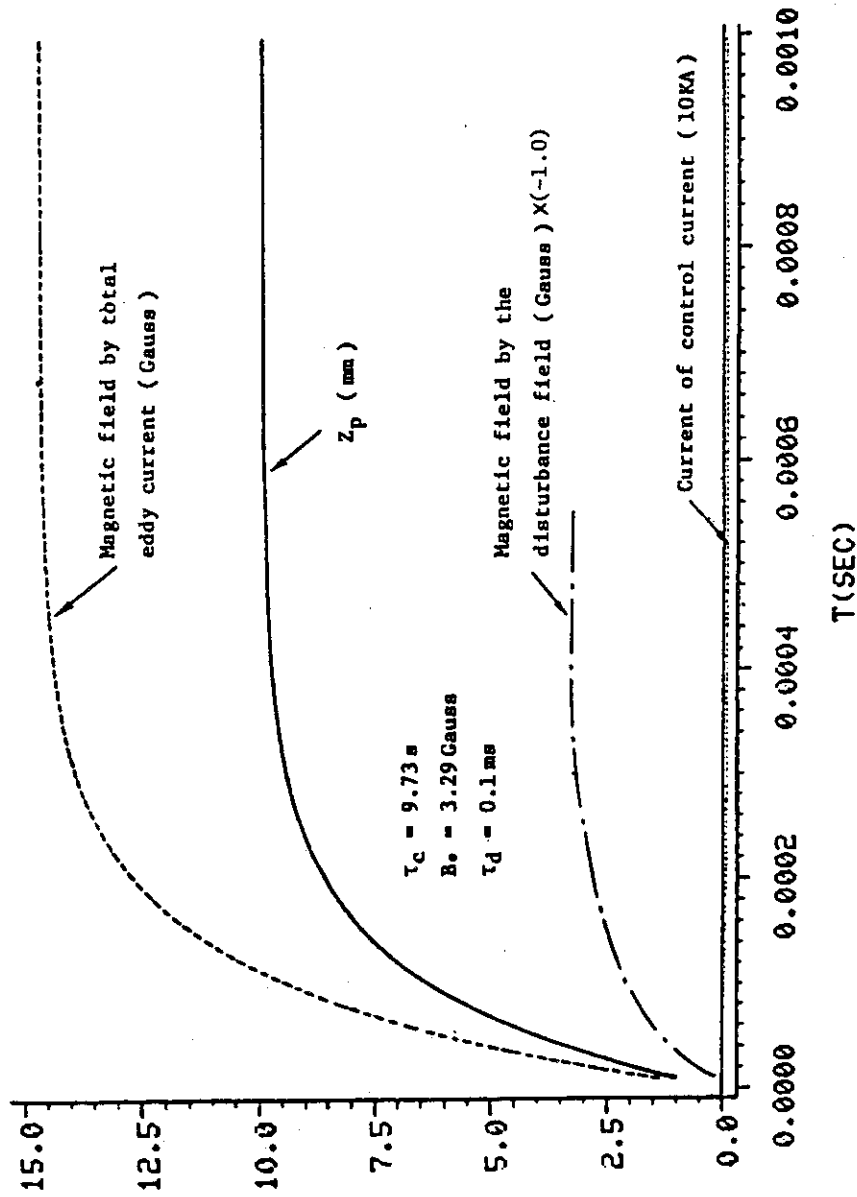


Fig. 4.1 Time evolution of the vertical displacement of plasma column,  $Z_p$ , under an external disturbance field,  $B_d = B_0 \cdot [1 - \exp(-t/\tau_d)]$  for Benchmark Model-1, and time evolution of the horizontal magnetic field due to the external disturbance, the magnetic field due to the eddy currents, and the current in the control coils when being regarded as passive conductors.

# INTOR BENCHMARK MODEL-1

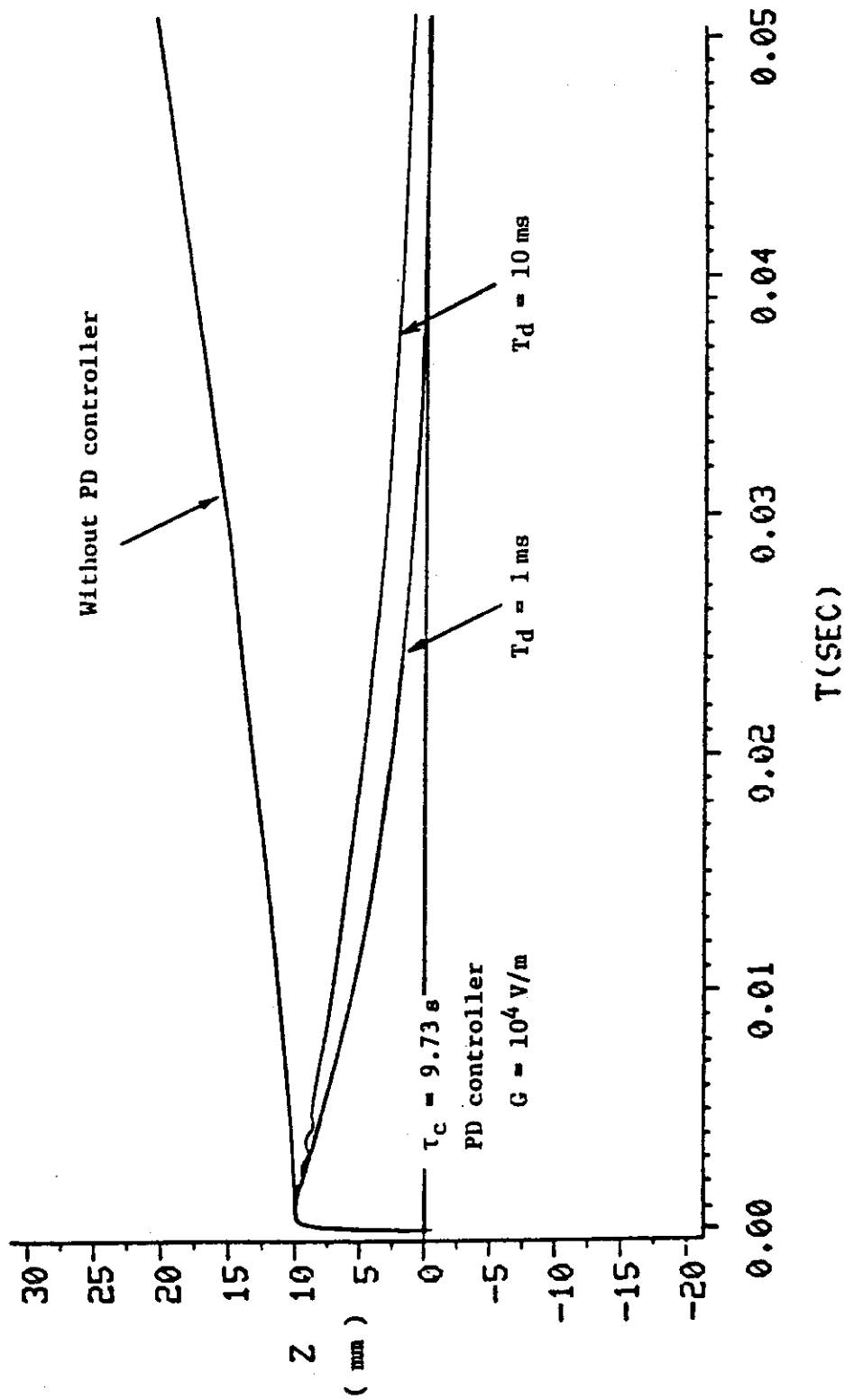


Fig. 4.2 Time evolutions of  $Z_p$  with and without active control for Benchmark Model-1

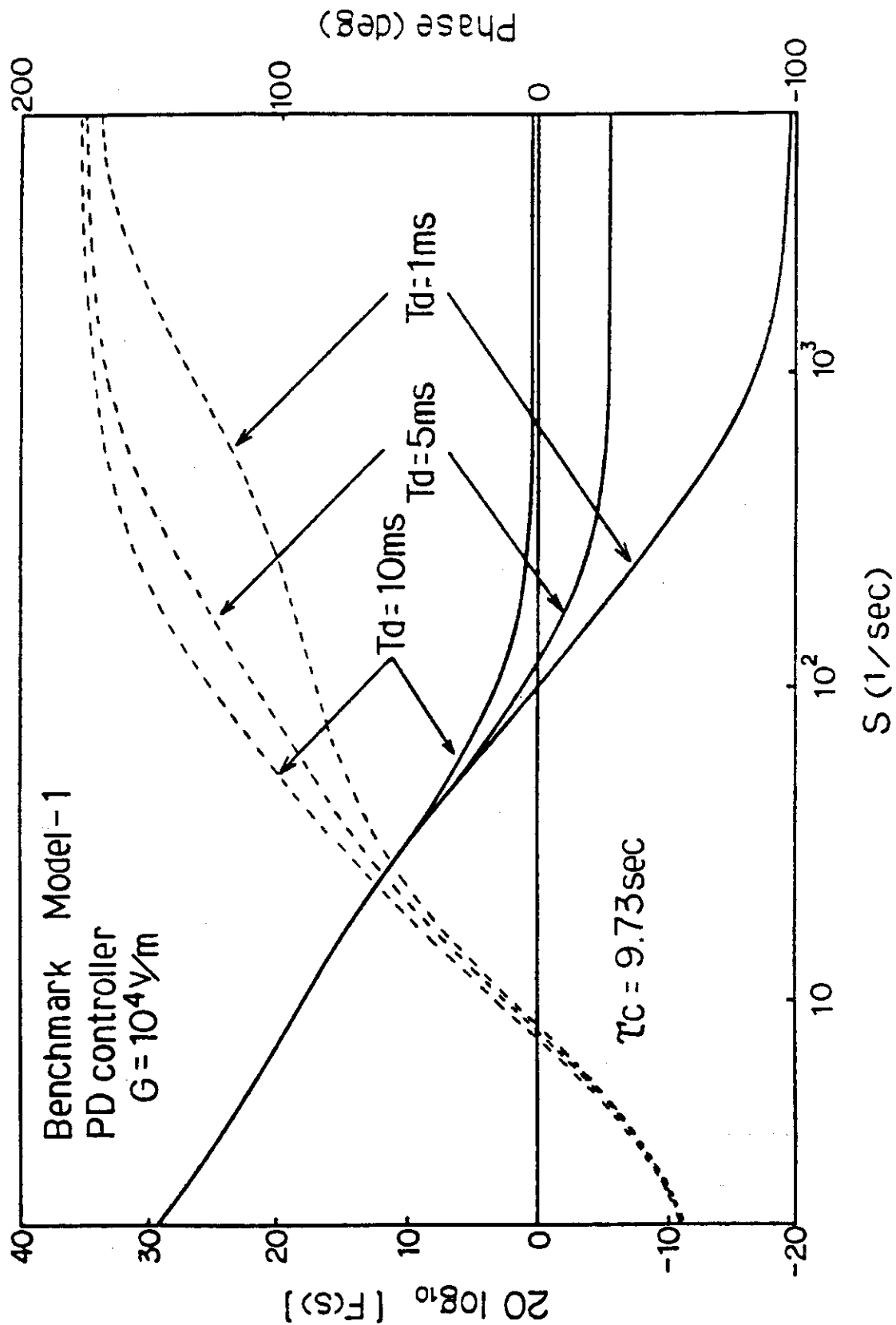


Fig. 4.3 Bode diagram with the derivative action time,  $T_d$  as a parameter for Benchmark Model-1

# INTOR BENCHMARK MODEL-1

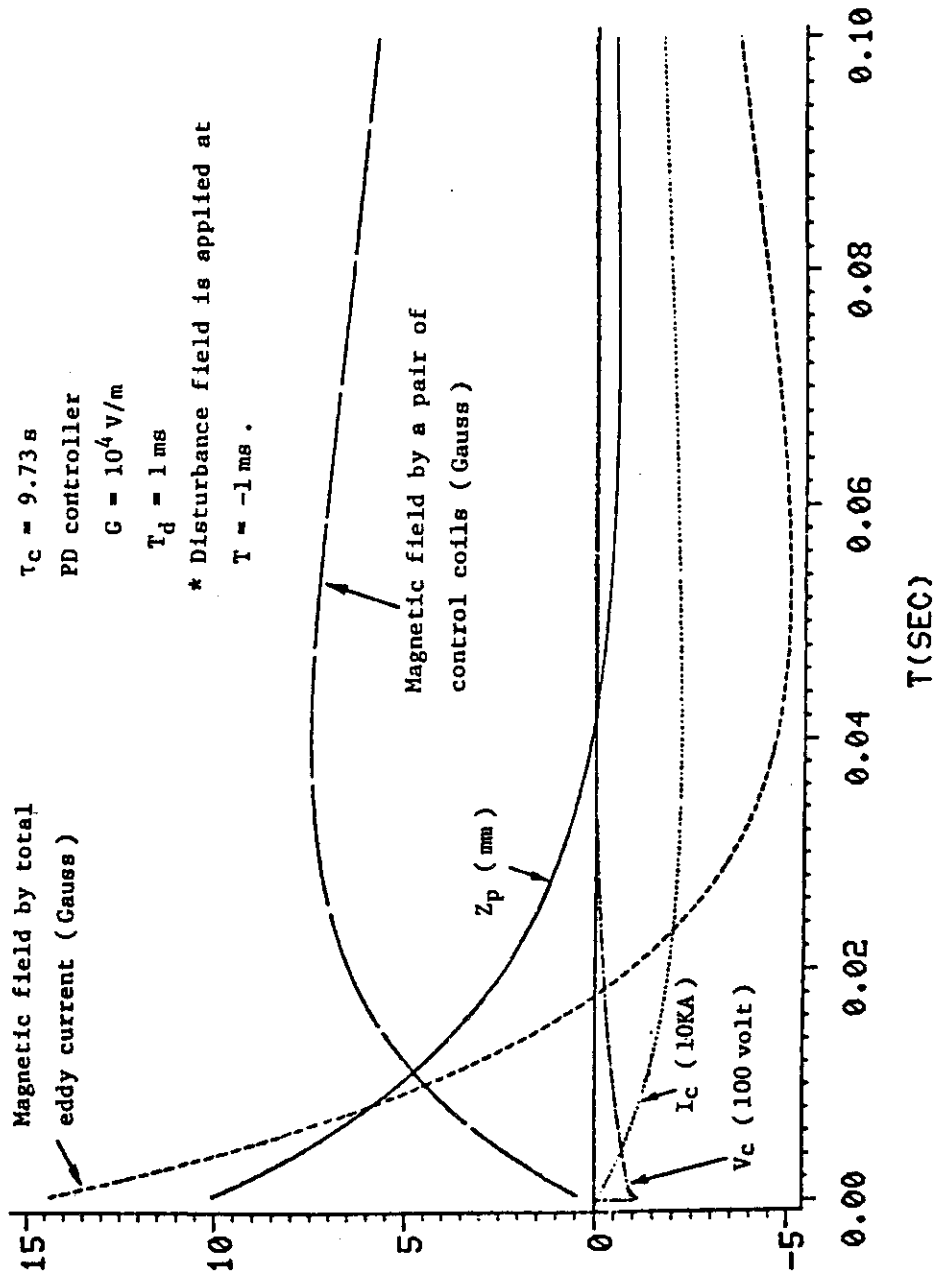


Fig. 4.4 Position control characteristics for Benchmark Model-1 on the derivative action time,  $T_d=1 \text{ ms}$ , and time evolution of the magnetic field due to the eddy current, the magnetic field produced by a pair of control coils, the magnetic field composed from these two fields and the current and voltage of the control coils.

# INTOR BENCHMARK MODEL-1

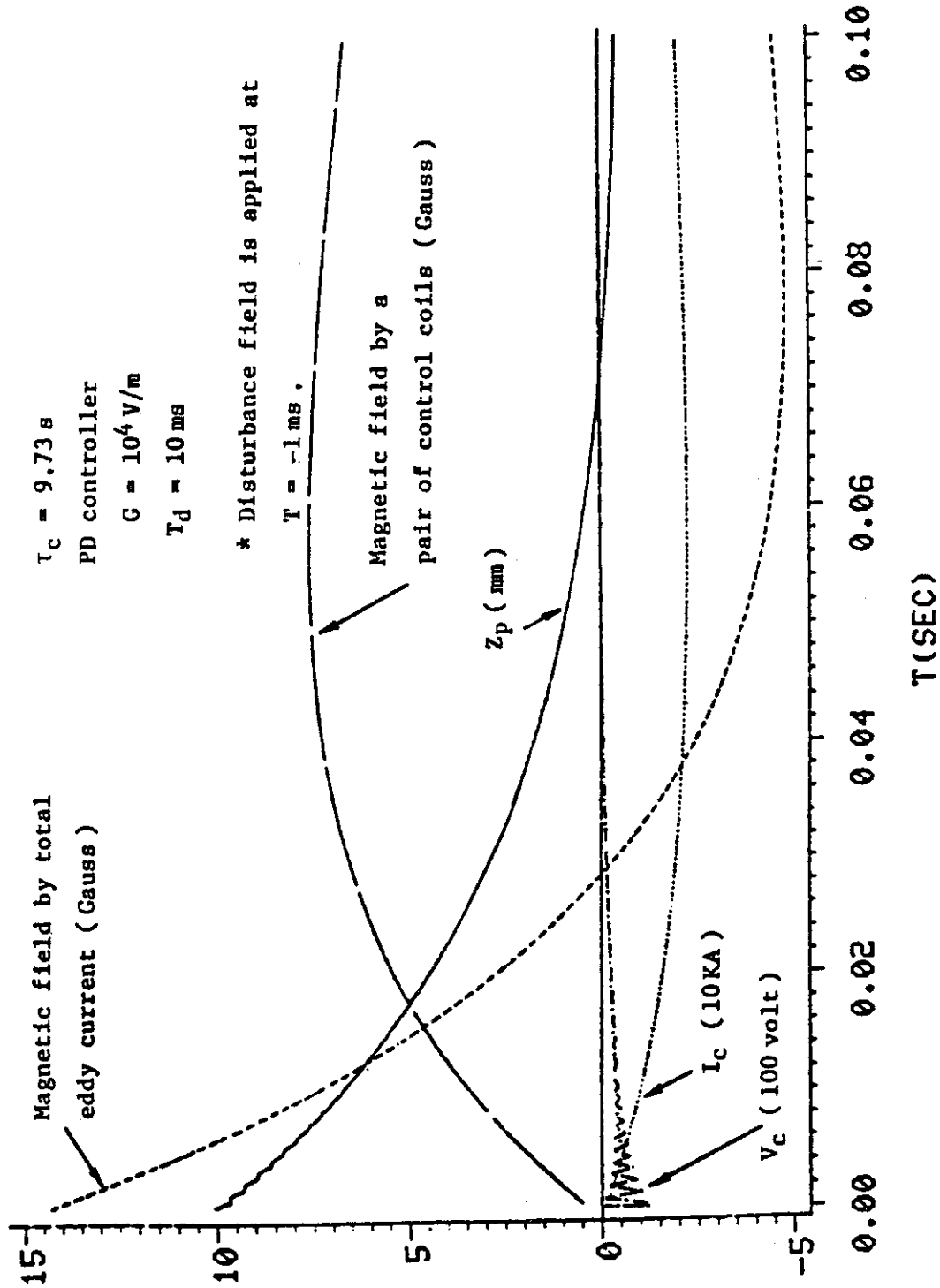


Fig. 4.5 Position control characteristics for Benchmark Model-1 on the derivative action time,  $T_d=10 \text{ ms}$ , and time evolutions of the same parameters as in Fig. 4.4.

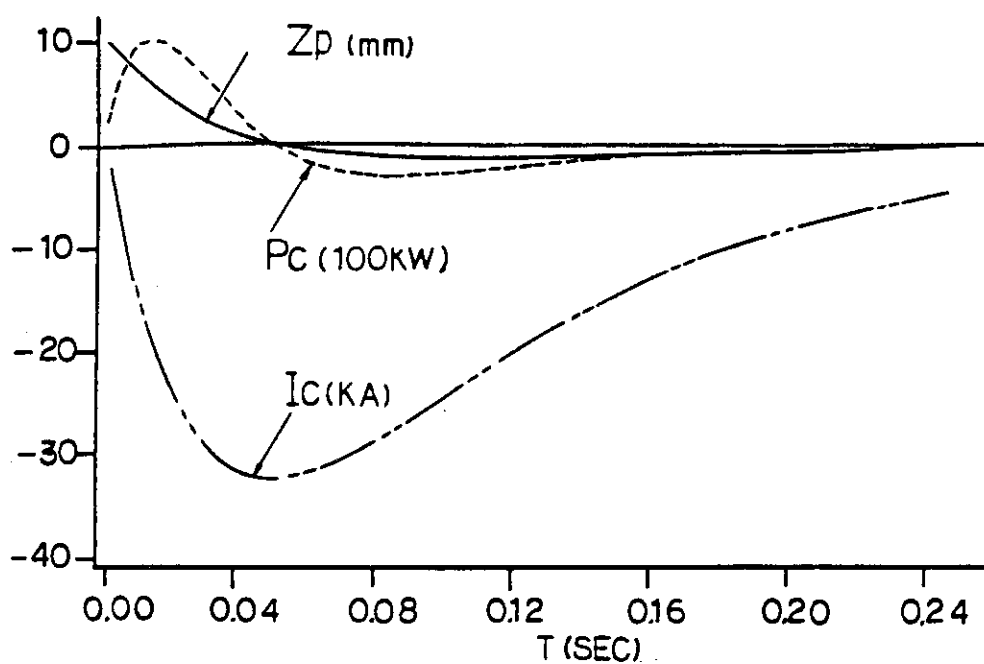
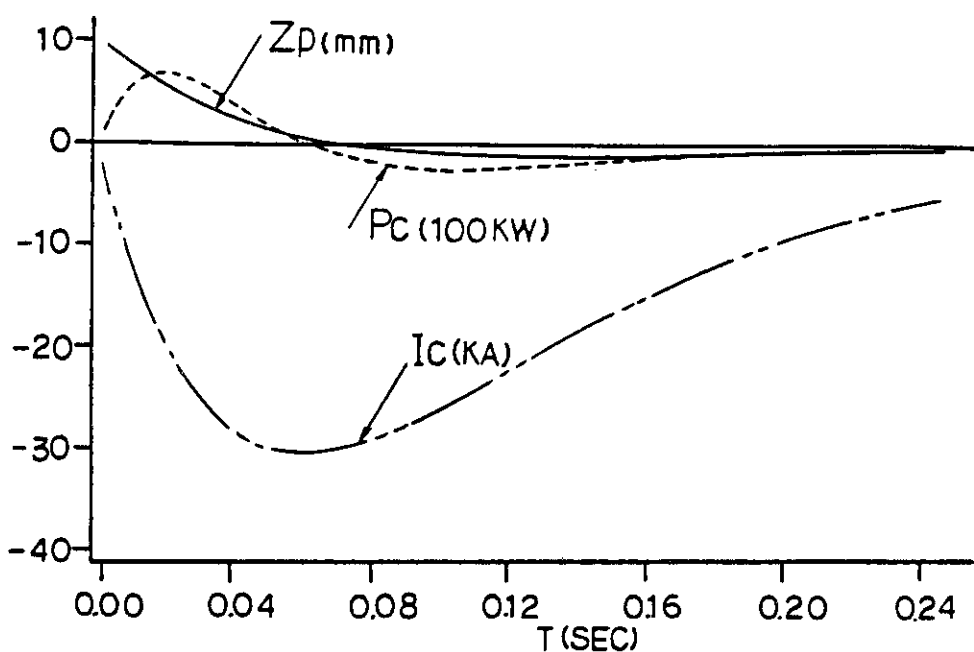
(a)  $T_d = 1\text{ms}$  and  $G = 10^4 \text{ v/m}$ (b)  $T_d = 10\text{ms}$  and  $G = 10^4 \text{ v/m}$ 

Fig. 4.6 Time evolution of  $Z_p$ , control coil current,  $I_c$  and required power,  $P_c$  in case of Benchmark Model-1 (The filamentary model is used for evaluation of  $(\partial M_{ps} / \partial Z) (\equiv M_{ps}')$ )

## INTOR BENCHMARK MODEL-2

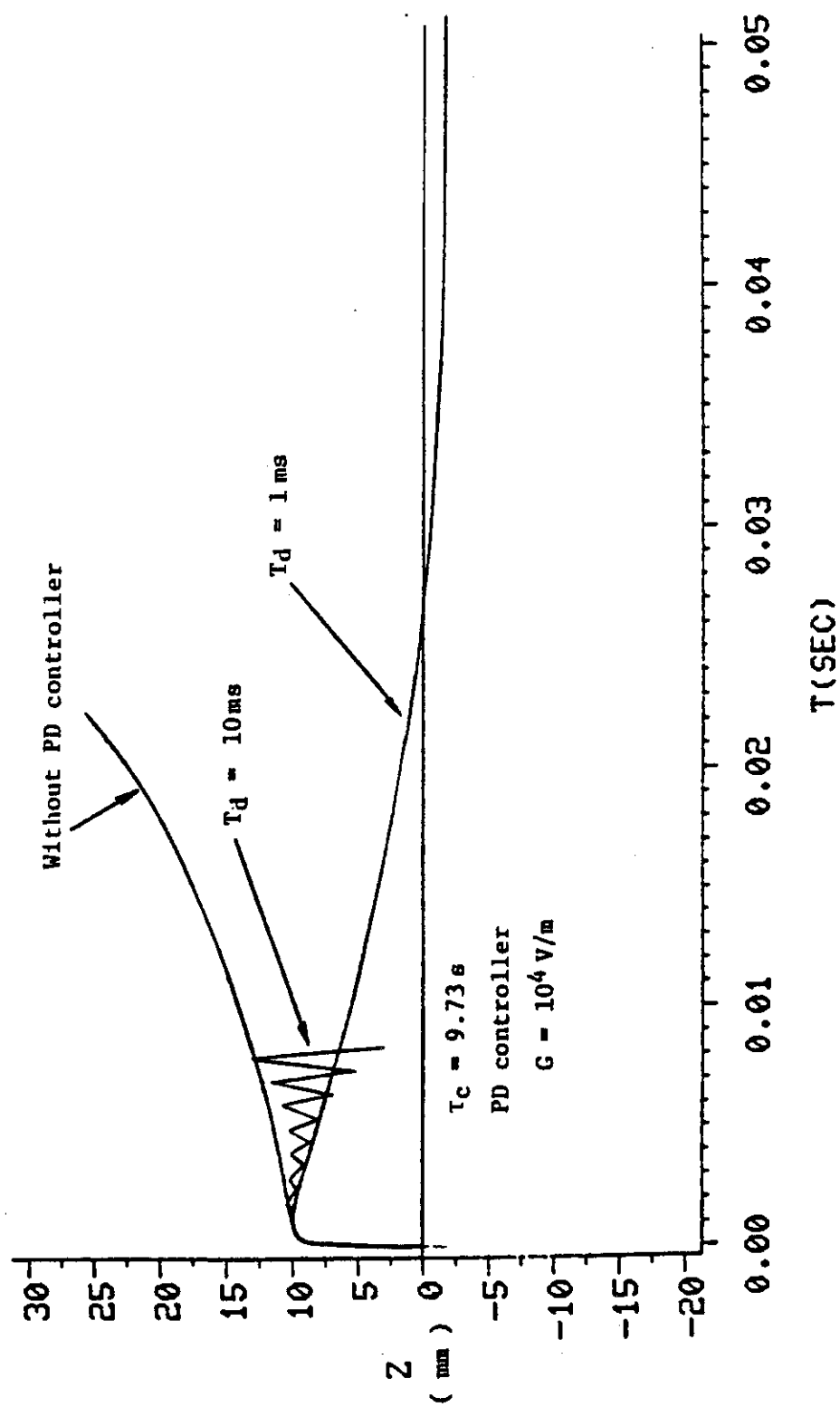


Fig. 4.7 Time evolutions of  $Z_p$  in Benchmark Model-2 with and without active control ( $T_d=1$  or  $10\text{ ms}$ )



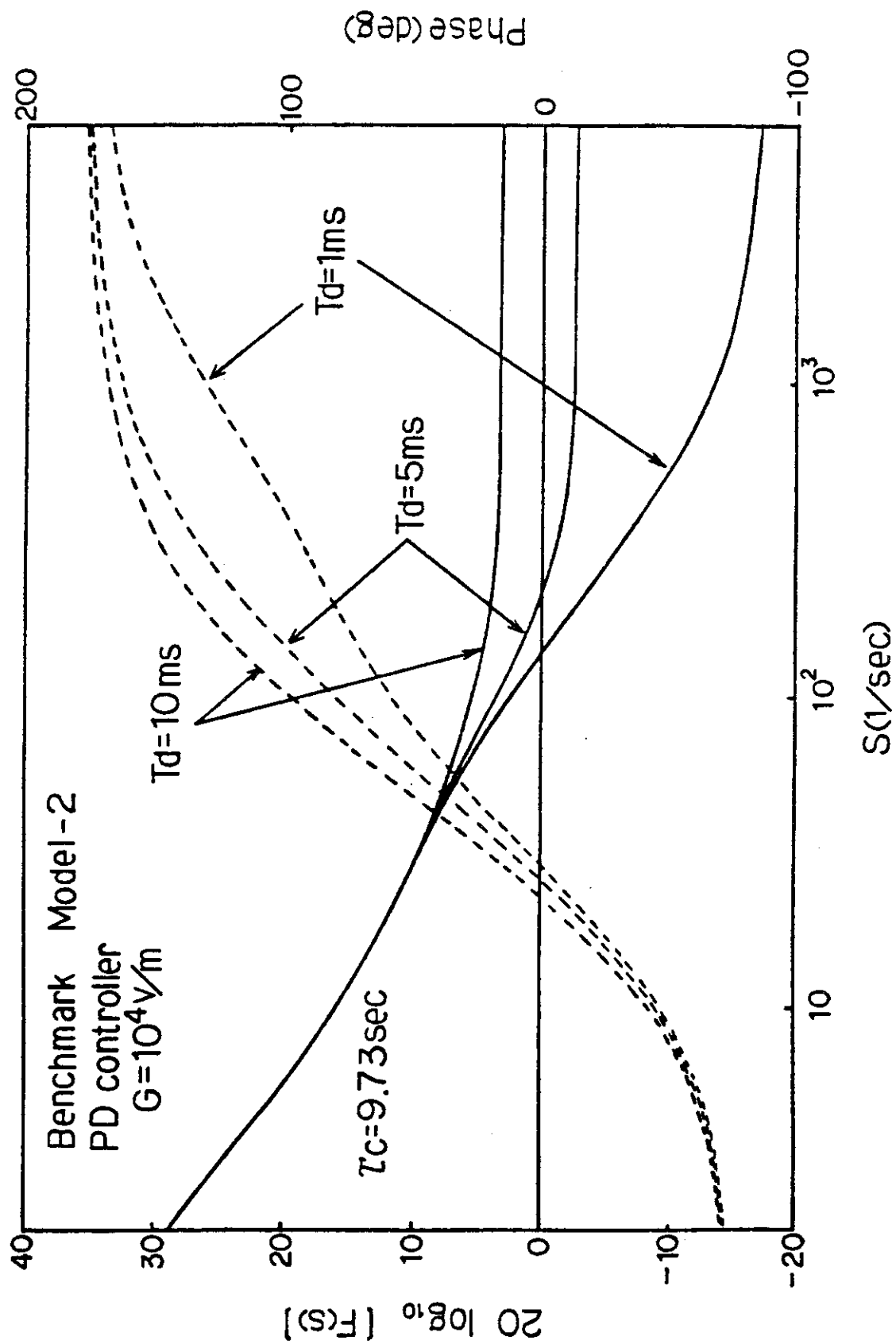


Fig. 4.8 Bode diagram with the derivative action time,  $T_d$  as a parameter

# INTOR BENCHMARK MODEL-2

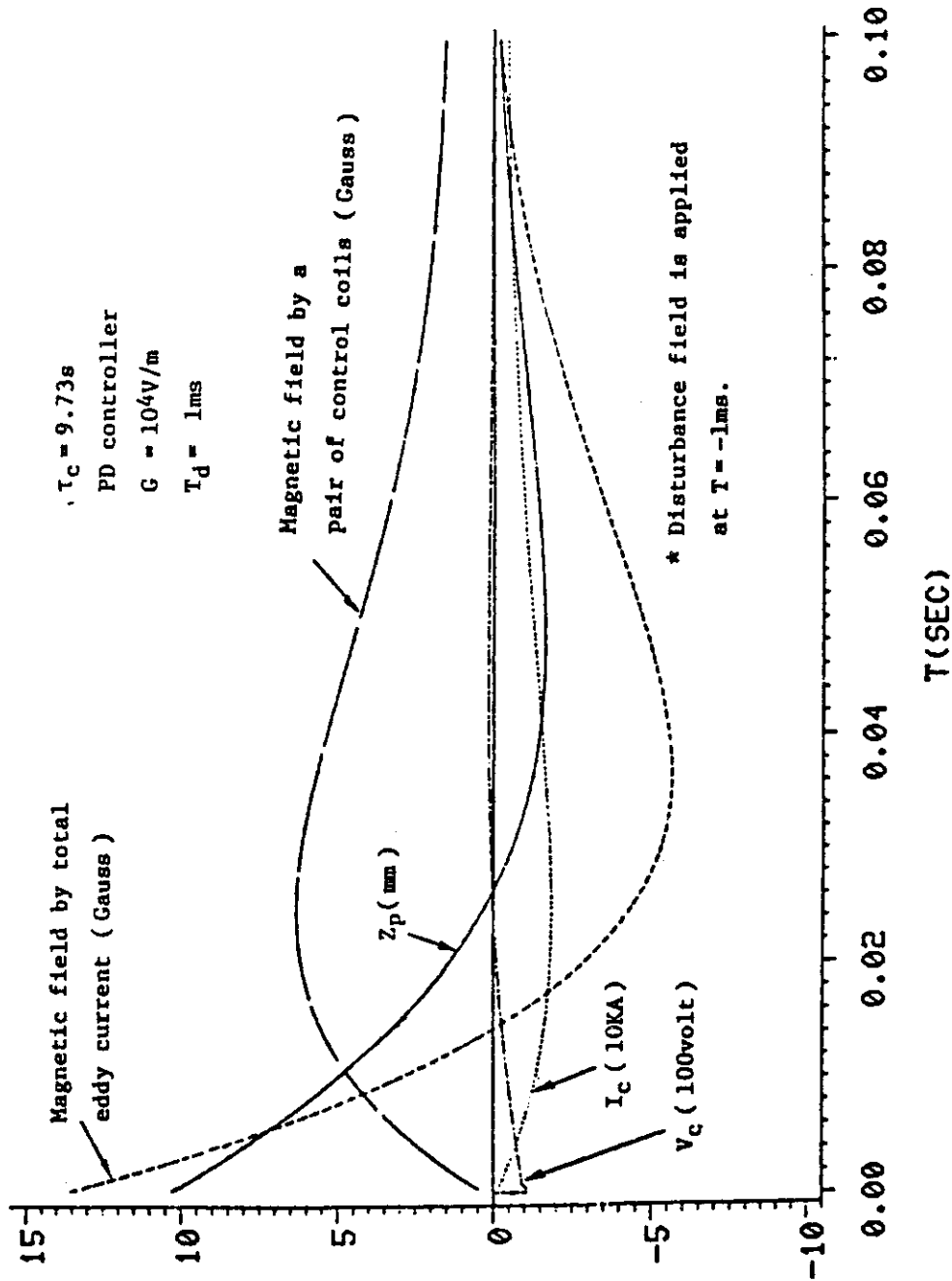


Fig. 4.9 Position control characteristics for Benchmark Model-2 on the derivative action time,  $T_d = 1ms$ , and time evolutions of the same parameters as in Fig. 4.4.

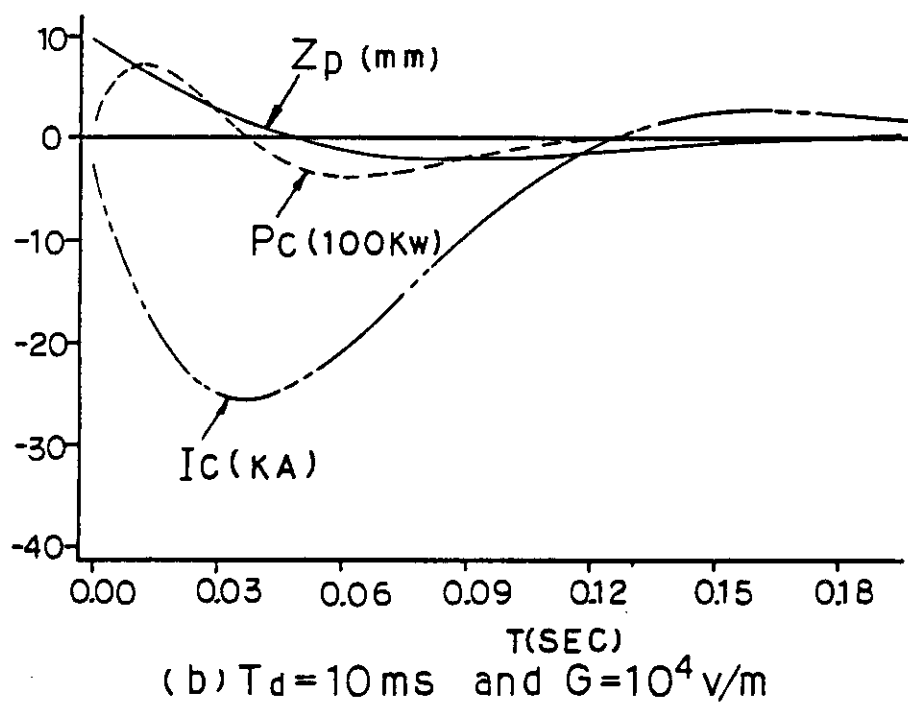
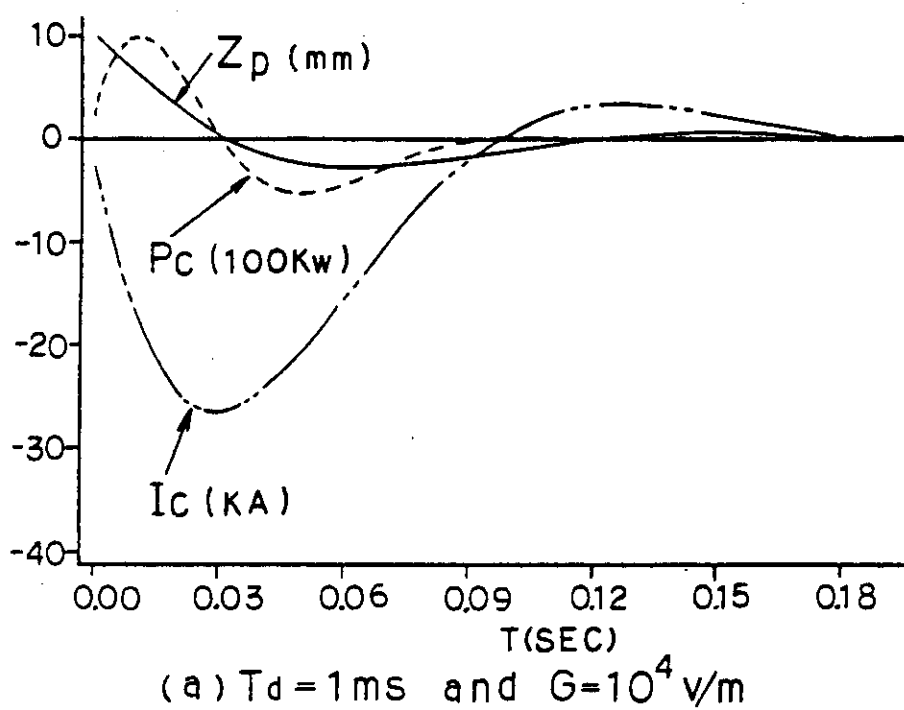


Fig. 4.10 Time evolution of  $Z_p$ ,  $I_c$ , and  $P_c$  in case of Benchmark Model-2. (The filamentary model is used for evaluation of  $M'_{ps}$ )

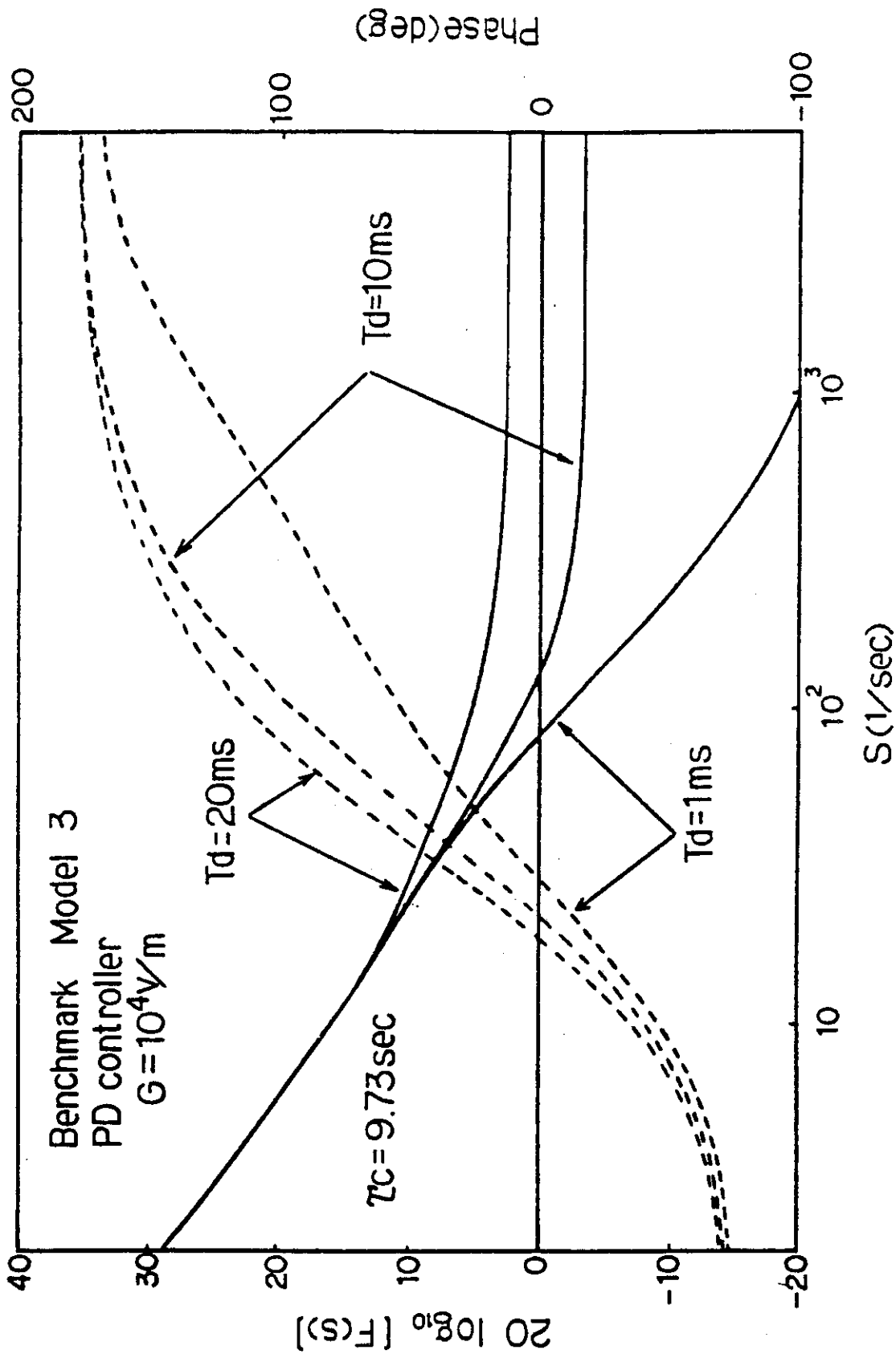


Fig. 4.11 Bode diagram with the derivative action time,  $T_d$  as a parameter

## INTOR BENCHMARK MODEL-3

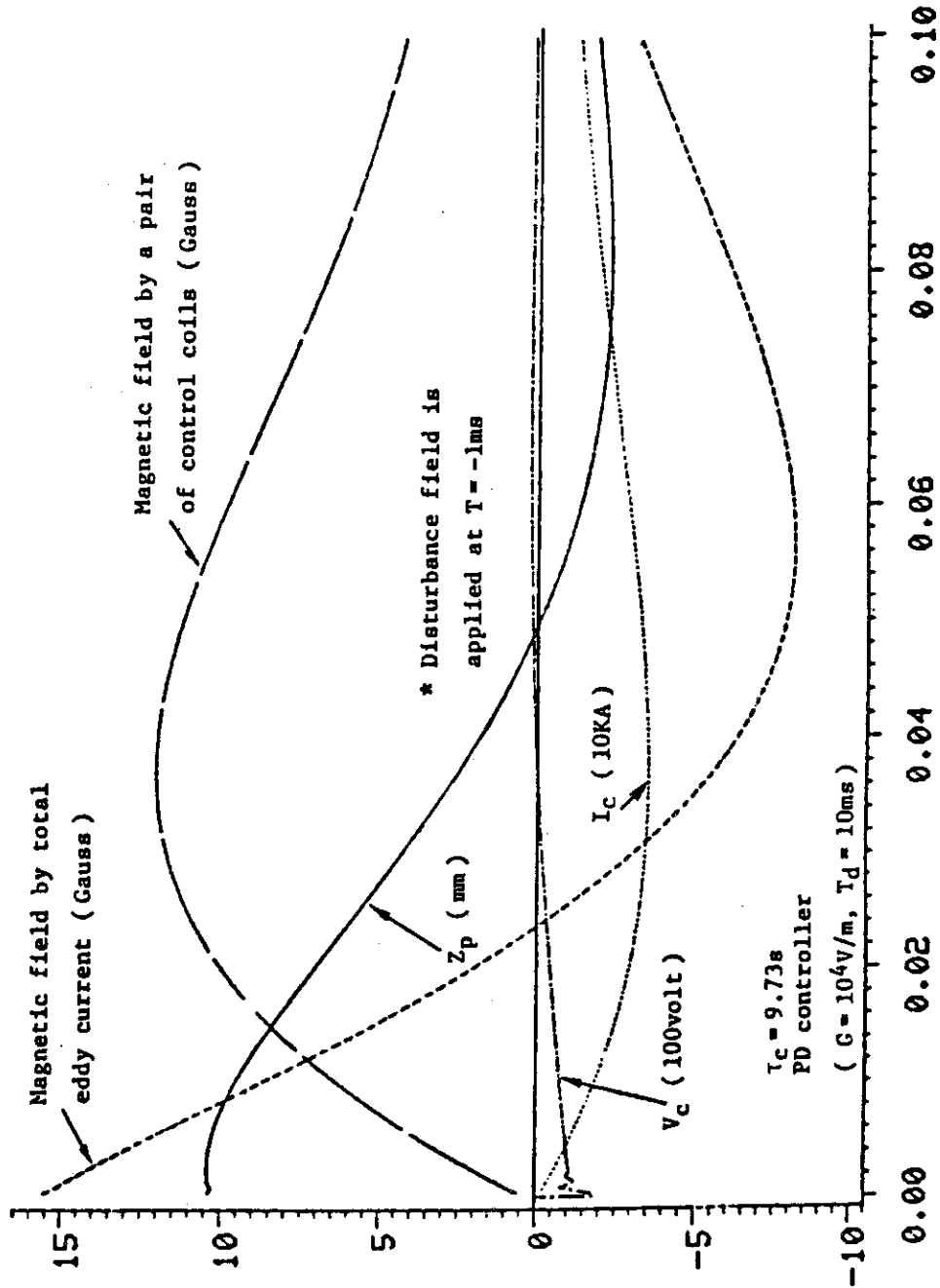
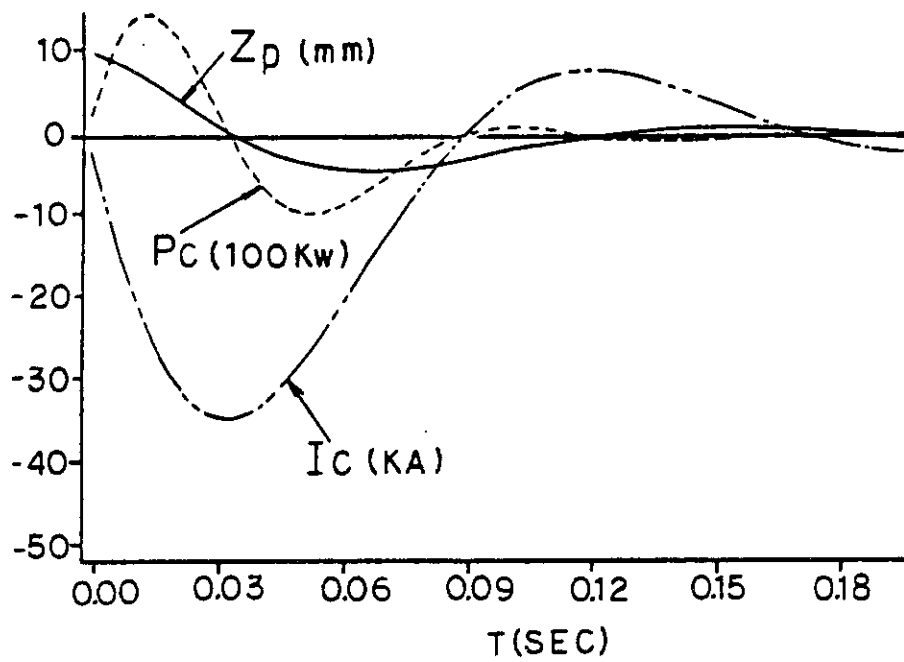


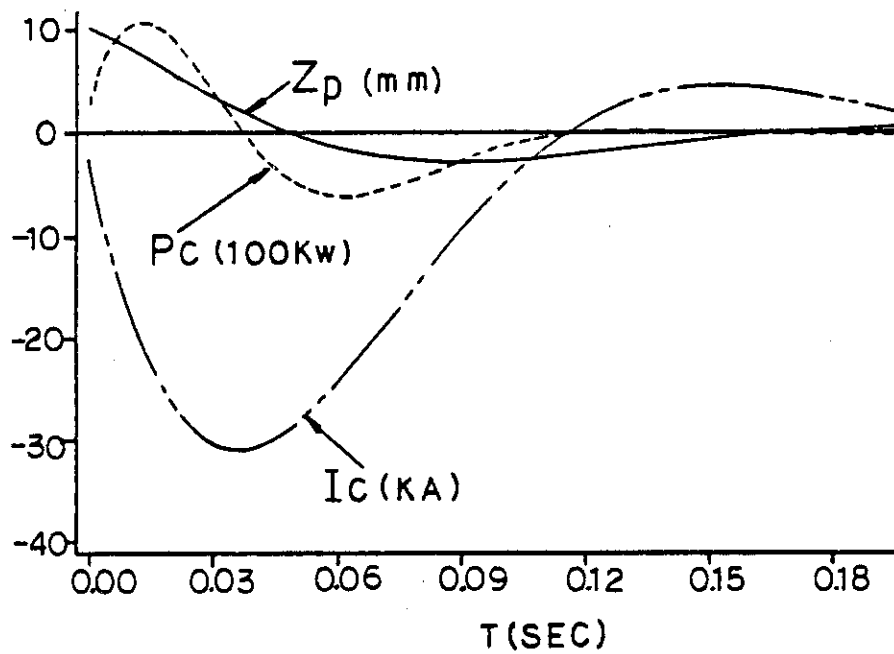
Fig. 4.12 Position control characteristics for Benchmark Model-3 on

the derivative action time,  $T_d = 10\text{ms}$ , and time evolutions

of the same parameters as in Fig. 4.4.



(a)  $T_d = 1\text{ms}$  and  $G = 10^4\text{v/m}$



(b)  $T_d = 10\text{ms}$  and  $G = 10^4\text{v/m}$

Fig. 4.13 Time evolution of  $Z_p$ ,  $I_c$ , and  $P_c$  in case of Benchmark Model-3. (The filamentary model is used for evaluation of  $M'_{ps}$ )

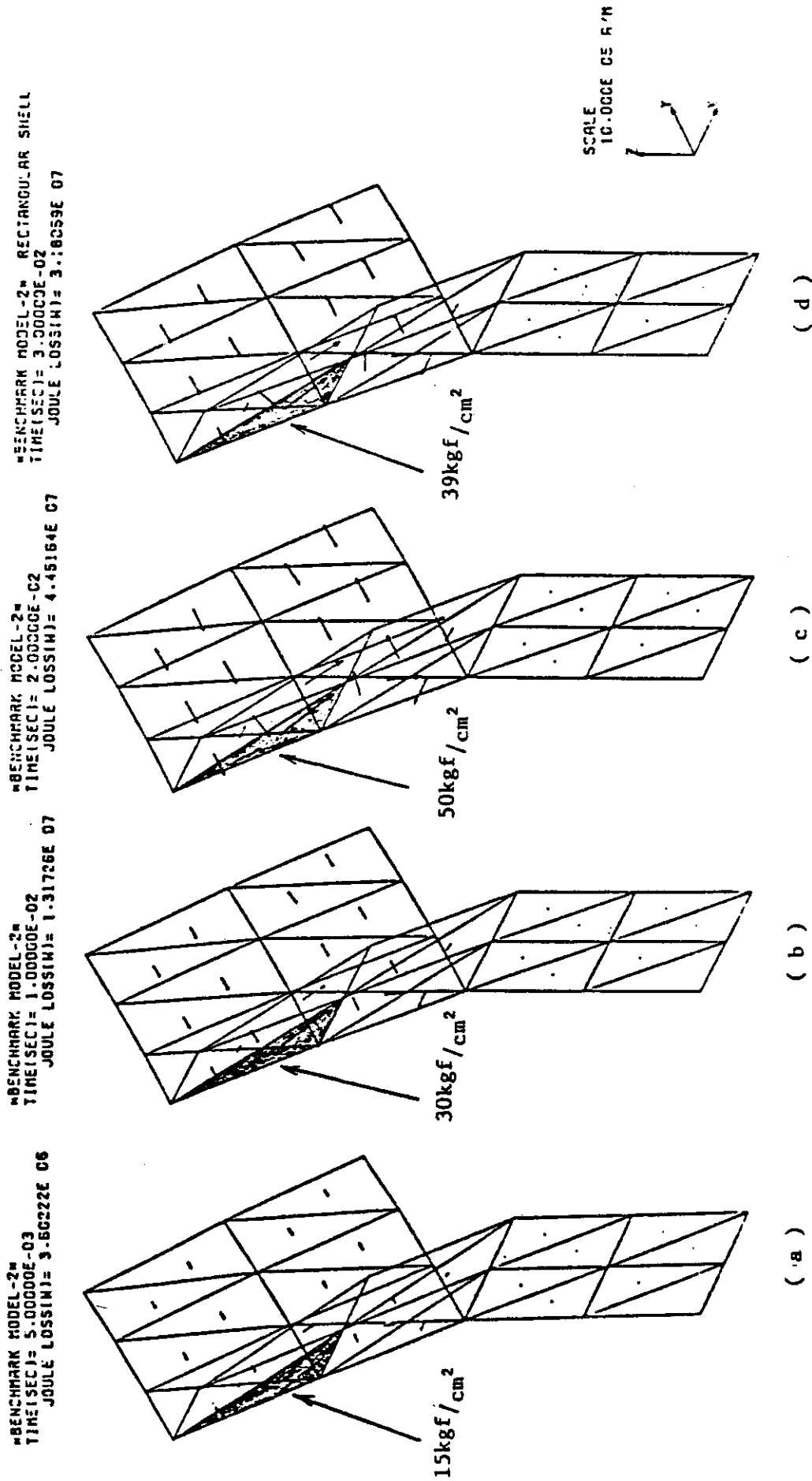


Fig. 4.14 Variation of the eddy current with the rectangular shell on plasma disruption in case of Benchmark Model-2.

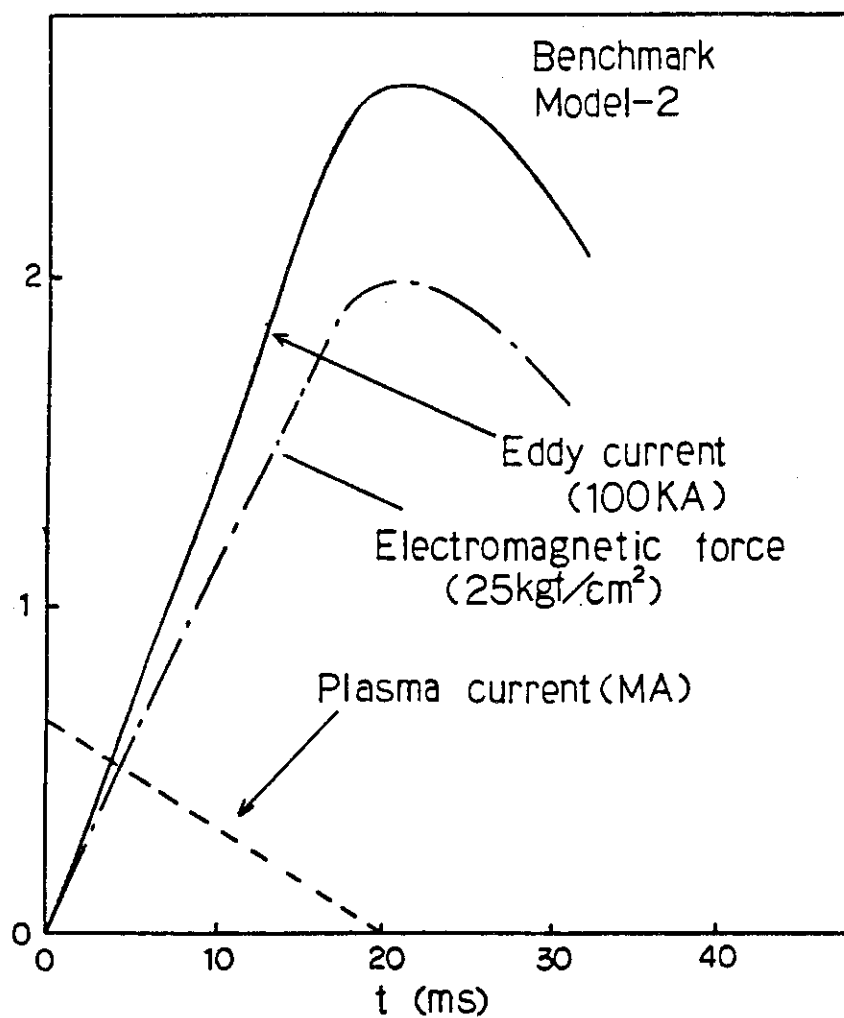


Fig. 4.15 Time evolutions of the plasma current, the eddy current and the electromagnetic force on plasma disruption in case of Benchmark Model-2.



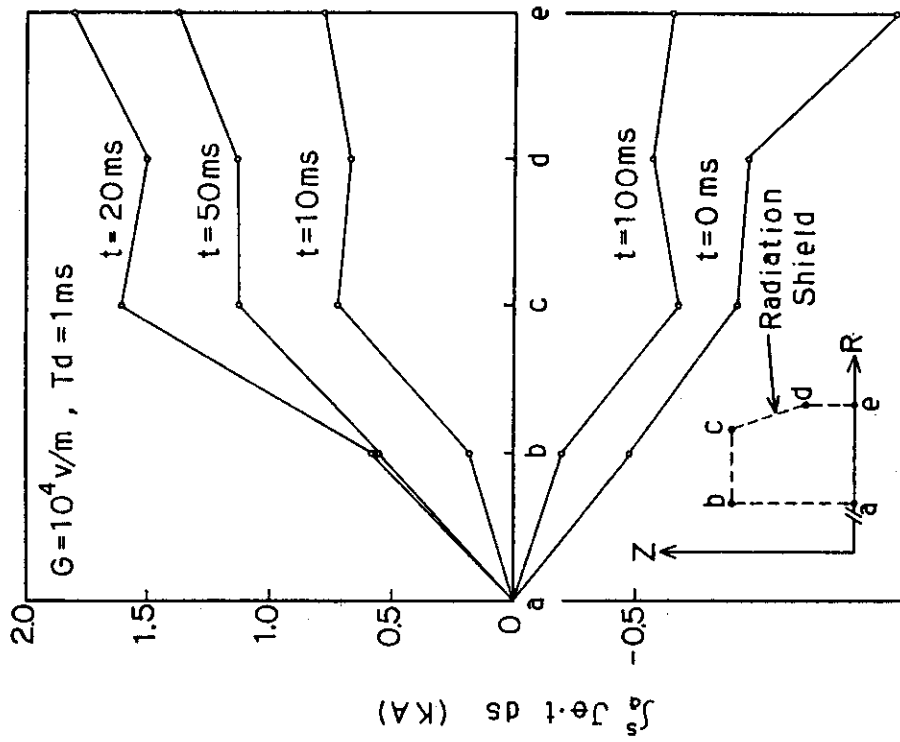


Fig. 4.16 Variations of current integral on the radiation shield in case of  $T_d = 1$  ms and  $G = 10^4$  V/m.

[This figure corresponds to Fig. 4.13 (a)]

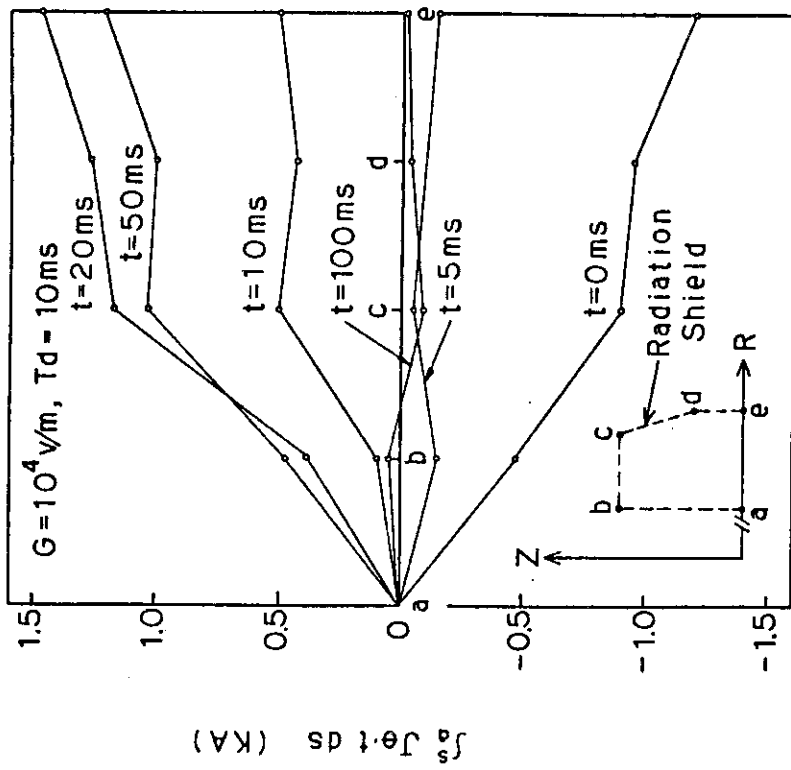


Fig. 4.17 Variations of current integral on the radiation shield in case of  $T_d = 10$  ms and  $G = 10^3$  V/m.

[This figure corresponds to Fig. 4.13 (b)]

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